



DAC7612

Dual, 12-Bit Serial Input DIGITAL-TO-ANALOG CONVERTER

FEATURES

- LOW POWER: 3.7mW
- FAST SETTLING: 7µs to 1 LSB
- 1mV LSB WITH 4.095V FULL-SCALE RANGE
- COMPLETE WITH REFERENCE
- 12-BIT LINEARITY AND MONOTONICITY OVER INDUSTRIAL TEMP RANGE
- 3-WIRE INTERFACE: Up to 20MHz Clock
- SMALL PACKAGE: 8-Lead SOIC

APPLICATIONS

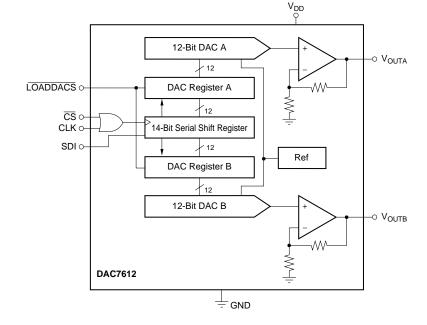
- PROCESS CONTROL
- DATA ACQUISITION SYSTEMS
- CLOSED-LOOP SERVO-CONTROL
- PC PERIPHERALS
- PORTABLE INSTRUMENTATION

DESCRIPTION

The DAC7612 is a dual, 12-bit digital-to-analog converter (DAC) with guaranteed 12-bit monotonicity performance over the industrial temperature range. It requires a single +5V supply and contains an input shift register, latch, 2.435V reference, a dual DAC, and high speed rail-to-rail output amplifiers. For a full-scale step, each output will settle to 1 LSB within 7µs while only consuming 3.7mW.

The synchronous serial interface is compatible with a wide variety of DSPs and microcontrollers. Clock (CLK), Serial Data In (SDI), Chip Select (\overline{CS}) and Load DACs ($\overline{LOADDACS}$) comprise the serial interface.

The DAC7612 is available in an 8-lead SOIC package and is fully specified over the industrial temperature range of -40° C to $+85^{\circ}$ C.



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SPECIFICATIONS

At $T_A = -40^{\circ}C$ to +85°C, and $V_{DD} = +5V$, unless otherwise noted.

			DAC7612U		I	DAC7612U	В	
PARAMETER	CONDITIONS	MIN	ТҮР	MAX	MIN	TYP	MAX	UNITS
RESOLUTION		12			*			Bits
ACCURACY Relative Accuracy ⁽¹⁾ Differential Nonlinearity Zero-Scale Error Zero Scale Match Full-Scale Voltage Full-Scale Match	Guaranteed Monotonic Code 000 _H Code 000 _H Code FFF _H Code FFF _H	-2 -1 -1 4.079	±1/2 ±1/2 +1 1/2 4.095 1/2	+2 +1 +3 4.111	-1 -1 * 4.087	±1/4 ±1/4 * 1/2 4.095 1/2	+1 +1 * 2 4.103 2	LSB LSB LSB V LSB
ANALOG OUTPUT Output Current Load Regulation Capacitive Load Short-Circuit Current Short-Circuit Duration	$\label{eq:code_sol} \begin{array}{c} \mbox{Code $800_H} \\ \mbox{R}_{\mbox{LOAD}} \geq 402\Omega, \mbox{ Code $800_H} \\ \mbox{ No Oscillation} \\ \mbox{ GND or } \mbox{V}_{\mbox{DD}} \end{array}$	±5	±7 1 500 ±15 Indefinite	3	*	* * * * *	*	mA LSB pF mA
DIGITAL INPUT Data Format Data Coding Logic Family Logic Levels		S	Serial traight Bina CMOS	ry		* * *		
V _{IH} V _{IL} I _{IH}		0.7 • V _{DD}		0.3 • V _{DD} ±10 ±10	*		* * *	V V μΑ μΑ
DYNAMIC PERFORMANCE Settling Time ⁽²⁾ (t _S) DAC Glitch Digital Feedthrough	To ±1 LSB of Final Value		7 2.5 0.5			* * *		μs nV-s nV-s
POWER SUPPLY V _{DD} I _{DD} Power Dissipation Power Supply Sensitivity	$\label{eq:V_IH} \begin{split} V_{IH} = 5V, V_{IL} = 0V, \text{No Load, at Code } 000_{\text{H}} \\ V_{IH} = 5V, V_{IL} = 0V, \text{No Load} \\ \Delta V_{DD} = \pm 5\% \end{split}$	+4.75	+5.0 0.75 3.5 0.0025	+5.25 1.5 7.5 0.002	*	* * * *	* * * * *	V mA mW %/%
TEMPERATURE RANGE Specified Performance		-40		+85	*		*	°C

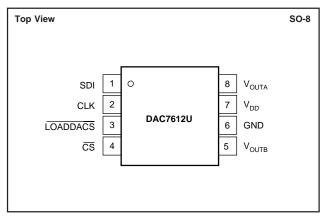
* Same specification as for DAC7612U.

NOTES: (1) This term is sometimes referred to as Linearity Error or Integral Nonlinearity (INL). (2) Specification does not apply to negative-going transitions where the final output voltage will be within 3 LSBs of ground. In this region, settling time may be double the value indicated.

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PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS(1)

V _{DD} to GND	–0.3V to 6V
Digital Inputs to GND	0.3V to V _{DD} + 0.3V
V _{OUT} to GND	0.3V to V _{DD} + 0.3V
Power Dissipation	325mW
Thermal Resistance, θ_{JA}	150°C/W
Maximum Junction Temperature	+150°C
Operating Temperature Range	–40°C to +85°C
Storage Temperature Range	–65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

NOTE: (1) Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

PACKAGE/ORDERING INFORMATION

PIN DESCRIPTIONS

PIN	LABEL	DESCRIPTION
1	SDI	Serial Data Input. Data is clocked into the internal serial register on the rising edge of CLK.
2	CLK	Synchronous Clock for the Serial Data Input.
3	LOADDACS	Loads the internal DAC registers. All DAC registers are transparent latches and are transparent when $\overline{\text{LOADDACS}}$ is LOW (regardless of the state of $\overline{\text{CS}}$ or CLK).
4	CS	Chip Select. Active LOW.
5	V _{OUTB}	DAC B Output Voltage
6	GND	Ground
7	V _{DD}	Positive Power Supply
8	V _{OUTA}	DAC A Output Voltage

ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

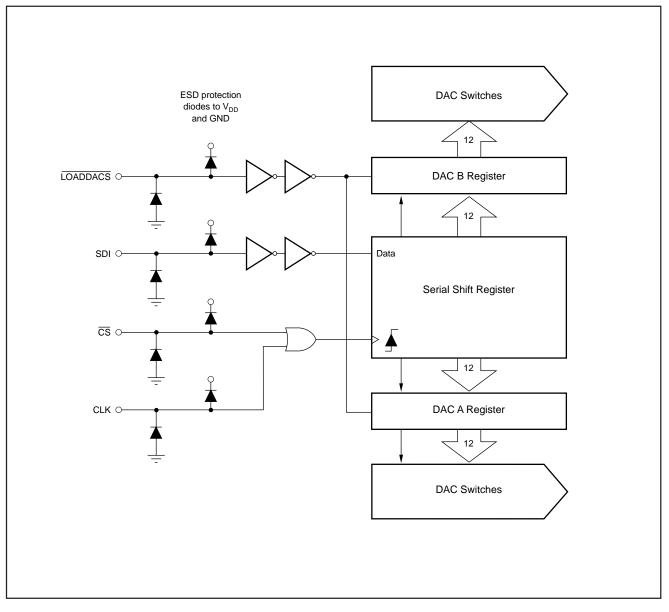
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PRODUCT	MINIMUM RELATIVE ACCURACY (LSB)	DIFFERENTIAL NONLINEARITY (LSB)	SPECIFICATION TEMPERATURE RANGE	PACKAGE	PACKAGE DRAWING NUMBER ⁽¹⁾	ORDERING NUMBER ⁽²⁾	TRANSPORT MEDIA
DAC7612U " DAC7612UB "	±2 " ±1 "	±1 " ±1"	-40°C to +85°C " -40°C to +85°C "	SO-8 " SO-8 "	182 " 182 "	DAC7612U DAC7612U/2K5 DAC7612UB DAC7612UB/2K5	Rails Tape and Reel Rails Tape and Reel

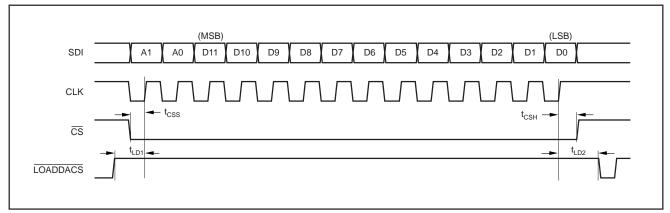
NOTES: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book. (2) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K5 indicates 2500 devices per reel). Ordering 2500 pieces of "DAC7612U/2K5" will get a single 2500-piece Tape and Reel. For detailed Tape and Reel mechanical information, refer to Appendix B of Burr-Brown IC Data Book.

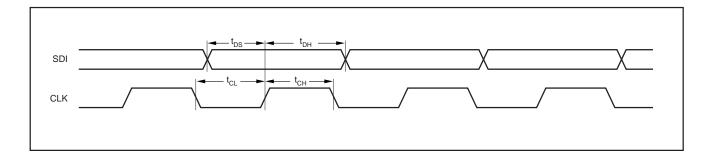


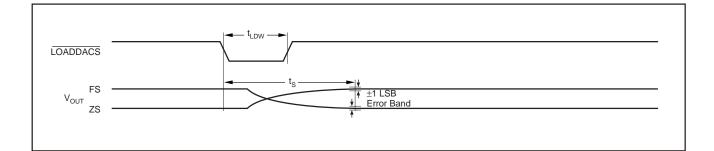
EQUIVALENT INPUT LOGIC



TIMING DIAGRAMS







LOGIC TRUTH TABLE

A1	A0	CLK	ĊS	LOADDACS	SERIAL SHIFT REGISTER	DAC REGISTER A	DAC REGISTER B
Х	Х	Х	Н	Н	No Change	No Change	No Change
Х	Х	Ŷ	L	Н	Shifts One Bit	No Change	No Change
L	Х	Х	H ⁽¹⁾	L	No Change	Loads Serial Data Word	Loads Serial Data Word
Н	L	Х	Н	L	No Change	Loads Serial Data Word	No Change
Н	Н	Х	Н	L	No Change	No Change	Loads Serial Data Word

NOTE: (1) A HIGH value is suggested in order to avoid to "false clock" from advancing the shift register and changing the DAC voltage.

DATA INPUT TABLE

l

B0	B1	B2	B3	B4	B5	B6	B7	B8	B9	B10	B11	B12	B13
A1	A0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0

TIMING SPECIFICATIONS

 $T_A = -40^{\circ}C$ to +85°C and $V_{DD} = +5V$.

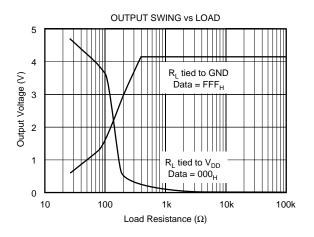
SYMBOL	DESCRIPTION	MIN	ТҮР	МАХ	UNITS
t _{CH}	Clock Width HIGH	30			ns
t _{CL}	Clock Width LOW	30			ns
t _{LDW}	Load Pulse Width				ns
t _{DS}	Data Setup	15			ns
t _{DH}	Data Hold	15			ns
t _{LD1}	Load Setup	15			ns
t _{LD2}	Load Hold	10			ns
t _{CSS}	Select	30			ns
t _{CSH}	Deselect	20			ns

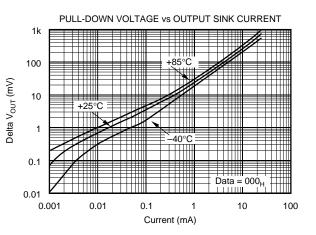
NOTE: All input control signals are specified with $t_R = t_F = 5$ ns (10% to 90% of +5V) and timed from a voltage level of 2.5V. These parameters are guaranteed by design and are not subject to production testing.



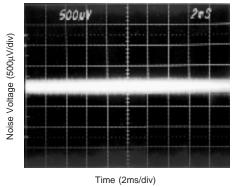
TYPICAL PERFORMANCE CURVES

At $T_A = +25^\circ$, and $V_{DD} = 5V$, unless otherwise specified.

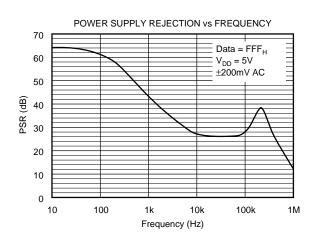


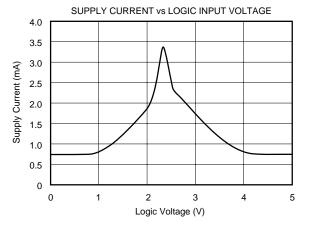


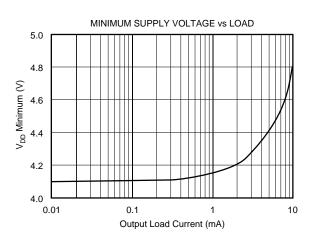
BROADBAND NOISE







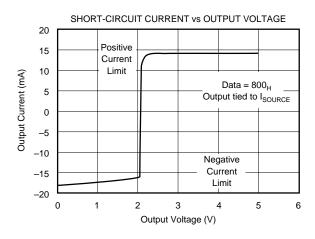


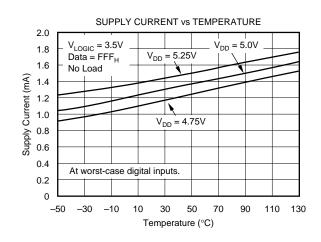




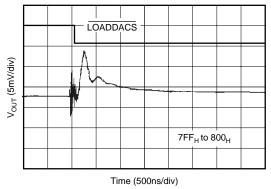
TYPICAL PERFORMANCE CURVES (CONT)

At $T_A = +25^\circ$, and $V_{DD} = 5V$, unless otherwise specified.





MIDSCALE GLITCH PERFORMANCE



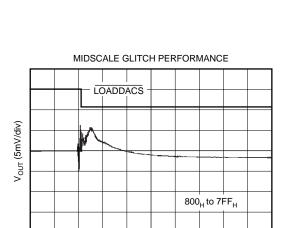
LARGE-SIGNAL SETTLING TIME

Time (20µs/div)

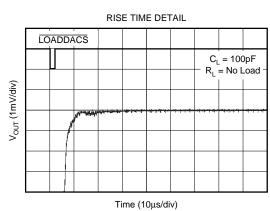
LOADDACS

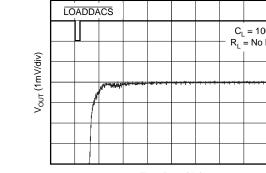
V_{OUT} (1V/div)

 $C_L = 100 pF$ R_L = No Load





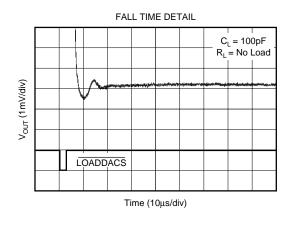


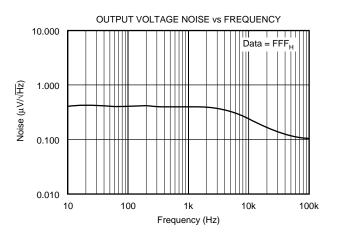




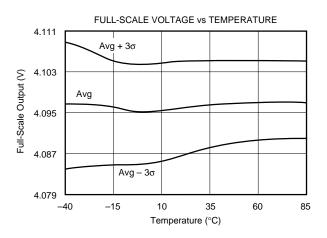
TYPICAL PERFORMANCE CURVES (CONT)

At T_{A} = +25°, and V_{DD} = 5V, unless otherwise specified.

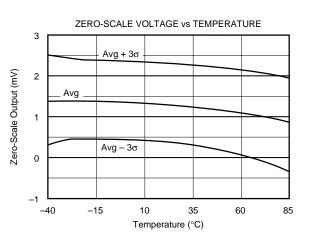




LONG-TERM DRIFT ACCELERATED BY BURN-IN 5 Output Voltage Change at FS (mV) 4 Max 3 2 1 Avg 0 -1 -2 Min -3 -4 -5 0 168 336 504 672 840 1008 Hours of Operation at +150°C

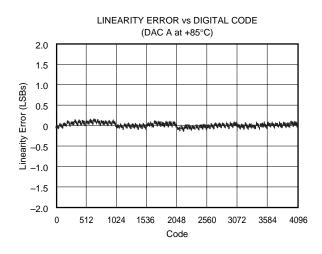


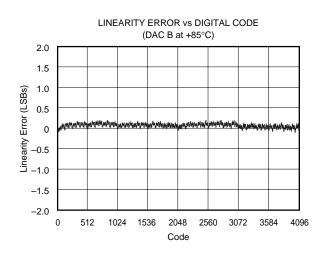
TOTAL UNADJUSTED ERROR HISTOGRAM 35 $T.U.E = \Sigma (INL + Z_{SE} + FSE)$ 30 Sample Size = 200 Units T_A = +25°C 25 Number of Units 20 15 10 5 0 -12 -10 -8 -6 -4 -2 0 2 6 8 10 12 4

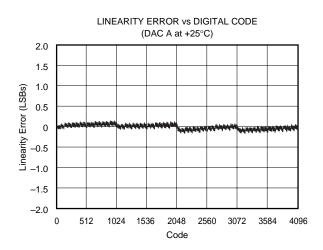


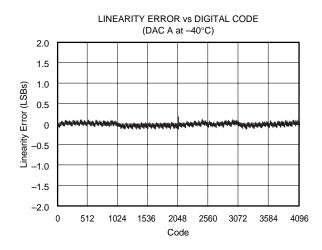
TYPICAL PERFORMANCE CURVES (CONT)

At $T_A = +25^\circ$, and $V_{DD} = 5V$, unless otherwise specified.

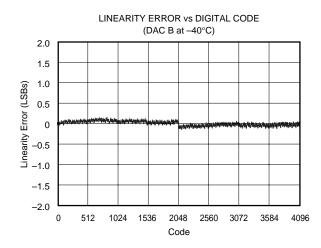








LINEARITY ERROR vs DIGITAL CODE (DAC B at +25°C) 2.0 1.5 1.0 Error (LSBs) 0.5 0 Linearity -0.5 -1.0 -1.5 -2.0 0 512 1024 1536 2048 2560 3072 3584 4096 Code





OPERATION

The DAC7612 is a dual, 12-bit digital-to-analog converter (DAC) complete with a serial-to-parallel shift register, DAC registers, laser-trimmed 12-bit DACs, on-board reference, and rail-to-rail output amplifiers. Figure 1 shows the basic operation of the DAC7612.

INTERFACE

Figure 1 shows the basic connection between a microcontroller and the DAC7612. The interface consists of a Serial Clock (CLK), Serial Data (SDI), and a Load DAC signal ($\overline{\text{LOADDACS}}$). In addition, a chip select ($\overline{\text{CS}}$) input is available to enable serial communication when there are multiple serial devices. Loading either DAC A or DAC B is done by shifting 14 serial bits in via the SDI input. The first 2 bits represent the address of the DAC to be updated and the

DAC7612 Full-Scale Range = 4.095V Least Significant Bit = 1mV							
DIGITAL INPUT CODE STRAIGHT OFFSETBINARY	ANALOG OUTPUT (V)	DESCRIPTION					
FFF _H	+4.095	Full Scale					
801 _H	+2.049	Midscale + 1 LSB					
800 _H	+2.048	Midscale					
7FF _H	+2.047	Midscale – 1 LSB					
000 _H	0	Zero Scale					

TABLE I. Digital Input Code and Corresponding Ideal Analog Output.

next 12 bits are the code (MSB-first) sent to the DAC. The data format is Straight Binary and is loaded MSB-first into the shift registers after loading the address bits. Table I shows the relationship between input code and output voltage.

The digital data into the DAC7612 is double-buffered. This means that new data can be entered into the chosen DAC without disturbing the old data and the analog output of the converter. At some point after the data has been entered into the serial shift register, this data can be transferred into the DAC registers. This transfer is accomplished with a HIGH to LOW transition of the LOADDACS pin. The LOADDACS pin makes the DAC registers transparent. If new data is shifted into the shift register while LOADDACS is LOW, the DAC output voltages will change as each new bit is entered. To prevent this, LOADDACS must be returned HIGH prior to shifting in new serial data.

DIGITAL-TO-ANALOG CONVERTER

The internal DAC section is a 12-bit voltage output device that swings between ground and the internal reference voltage. The DAC is realized by a laser-trimmed R-2R ladder network which is switched by N-channel MOSFETs. Each DAC output is internally connected to a rail-to-rail output operational amplifier.

OUTPUT AMPLIFIER

A precision, low-power amplifier buffers the output of each DAC section and provides additional gain to achieve a 0V to 4.095V range. Each amplifier has low offset voltage, low

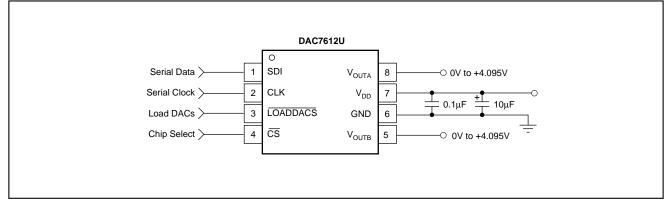


FIGURE 1. Basic Operation of the DAC7612.

noise, and a set gain of 1.682V/V (4.095/2.435). See Figure 2 for an equivalent circuit schematic of the analog portion of the DAC7612.

The output amplifier has a 7μ s typical settling time to ± 1 LSB of the final value. Note that there are differences in the settling time for negative-going signals versus positive-going signals.

The rail-to-rail output stage of the amplifier provides the fullscale range of 0V to 4.095V while operating on a supply voltage as low as 4.75V. In addition to its ability to drive resistive loads, the amplifier will remain stable while driving capacitive loads of up to 500pF. See Figure 3 for an equivalent circuit schematic of the amplifier's output driver and the Typical Performance Curves section for more information regarding settling time, load driving capability, and output noise.

POWER SUPPLY

A BiCMOS process and careful design of the bipolar and CMOS sections of the DAC7612 result in a very low power device. Bipolar transistors are used where tight matching and low noise are needed to achieve analog accuracy, and CMOS transistors are used for logic, switching functions and for other low power stages. If power consumption is critical, it is important to keep the logic levels on the digital inputs (SDI, CLK, \overline{CS} , $\overline{LOADDACS}$) as close as possible to either V_{DD} or ground. This will keep the CMOS inputs (see "Supply Current vs Logic Input Voltages" in the Typical Performance Curves) from shunting current between V_{DD} and ground.

The DAC7612 power supply should be bypassed as shown in Figure 1. The bypass capacitors should be placed as close to the device as possible, with the 0.1μ F capacitor taking priority in this regard. The "Power Supply Rejection vs Frequency" graph in the Typical Performance Curves section shows the PSRR performance of the DAC7612. This should be taken into account when using switching power supplies or DC/DC converters.

In addition to offering guaranteed performance with V_{DD} in the 4.75V to 5.25V range, the DAC7612 will operate with reduced performance down to 4.5V. Operation between 4.5V and 4.75V will result in longer settling time, reduced performance, and current sourcing capability. Consult the " V_{DD} vs Load Current" graph in the Typical Performance Curves section for more information.

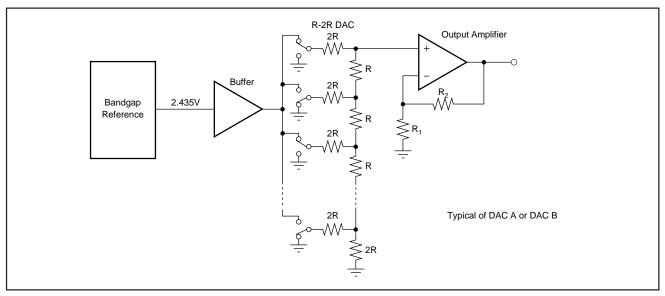


FIGURE 2. Simplified Schematic of Analog Portion.

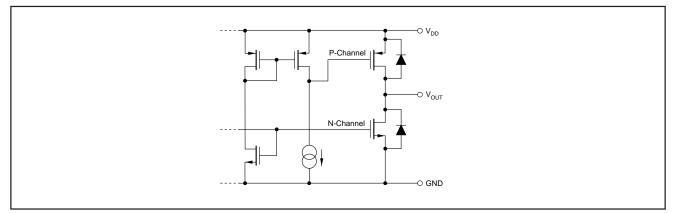


FIGURE 3. Simplified Driver Section of Output Amplifier.



APPLICATIONS

POWER AND GROUNDING

The DAC7612 can be used in a wide variety of situations from low power, battery operated systems to large-scale industrial process control systems. In addition, some applications require better performance than others, or are particularly sensitive to one or two specific parameters. This diversity makes it difficult to define definite rules to follow concerning the power supply, bypassing, and grounding. The following discussion must be considered in relation to the desired performance and needs of the particular system.

A precision analog component requires careful layout, adequate bypassing, and a clean, well-regulated power supply. As the DAC7612 is a single-supply, +5V component, it will often be used in conjunction with digital logic, microcontrollers, microprocessors, and digital signal processors. The more digital logic present in the design and the higher the switching speed, the more difficult it will be to achieve good performance.

Because the DAC7612 has a single ground pin, all return currents, including digital and analog return currents, must flow through this pin. The GND pin is also the ground reference point for the internal bandgap reference. Ideally, GND would be connected directly to an analog ground plane. This plane would be separate from the ground connection for the digital components until they are connected at the power entry point of the system (see Figure 4).

The power applied to V_{DD} should be well regulated and lownoise. Switching power supplies and DC/DC converters will often have high-frequency glitches or spikes riding on the output voltage. In addition, digital components can create similar high frequency spikes as their internal logic switches states. This noise can easily couple into the DAC output voltage through various paths between V_{DD} and V_{OUT} .

As with the GND connection, V_{DD} should be connected to a +5V power supply plane or trace that is separate from the connection for digital logic until they are connected at the power entry point. In addition, the 10µF and 0.1µF capacitors shown in Figure 4 are strongly recommended and should be installed as close to V_{DD} and ground as possible. In some situations, additional bypassing may be required such as a 100µF electrolytic capacitor or even a "Pi" filter made up of inductors and capacitors—all designed to essentially lowpass filter the +5V supply, removing the high frequency noise (see Figure 4).

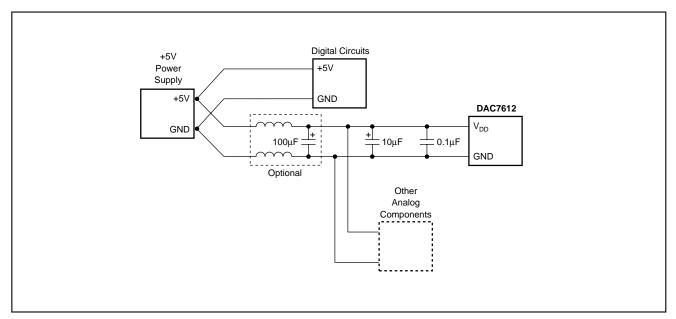


FIGURE 4. Suggested Power and Ground Connections for a DAC7612 Sharing a +5V Supply with a Digital System.

