Data Sheet

January 2003

10A, 30V, 0.013 Ohm, N-Channel, Logic Level UltraFET Power MOSFET



This N-Channel power MOSFET is manufactured using the innovative UltraFET process. This advanced process technology achieves the

lowest possible on-resistance per silicon area, resulting in outstanding performance. This device is capable of withstanding high energy in the avalanche mode and the diode exhibits very low reverse recovery time and stored charge. It was designed for use in applications where power efficiency is important, such as switching regulators, switching converters, motor drivers, relay drivers, low-voltage bus switches, and power management in portable and battery-operated products.

Formerly developmental type TA76131.

Ordering Information

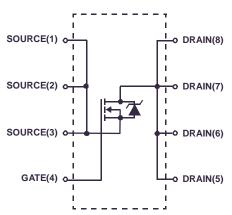
PART NUMBER	PACKAGE	BRAND
HUF76131SK8	MS-012AA	76131SK8

NOTE: When ordering, use the entire part number. Add the suffix T to obtain the variant in tape and reel, e.g., HUF76131SK8T.

Features

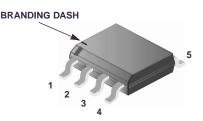
- · Logic Level Gate Drive
- 10A, 30V
- Ultra Low On-Resistance, $r_{DS(ON)} = 0.013\Omega$
- Temperature Compensating PSPICE[®] Model
- Thermal Impedance SPICE Model
- Peak Current vs Pulse Width Curve
- UIS Rating Curve
- · Related Literature
 - TB334, "Guidelines for Soldering Surface Mount Components to PC Boards"

Symbol



Packaging

JEDEC MS-012AA



HUF76131SK8

Absolute Maximum Ratings $T_A = 25^{\circ}C$, Unless Otherwise Specified

	HUF76131SK8	UNITS
Drain to Source Voltage (Note 1)	30	V
Drain to Gate Voltage ($R_{GS} = 20k\Omega$) (Note 1)	30	V
Gate to Source VoltageVGS	±20	V
Drain Current		
Continuous (Figure 2) (Notes 2, 3)	10	Α
Pulsed Drain CurrentI _{DM}	Figure 5	
Pulsed Avalanche RatingEAS	Figure 6	
Power Dissipation	2.5	W
Derate Above 25°C	0.02	W/oC
Operating and Storage Temperature	-55 to 150	°C
Maximum Temperature for Soldering		
Leads at 0.063in (1.6mm) from Case for 10s	300	οС
Package Body for 10s, See Techbrief 334	260	οС
prog.		

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

1. $T_J = 25^{\circ}C$ to $150^{\circ}C$.

Electrical Specifications $T_A = 25^{\circ}C$, Unless Otherwise Specified

PARAMETER	SYMBOL	TES	T CONDITIONS	MIN	TYP	MAX	UNITS
Drain to Source Breakdown Voltage	BV _{DSS}	$I_D = 250\mu A, V_{GS} = 0V \text{ (Figure 11)}$		30	-	-	V
Gate to Source Threshold Voltage	V _{GS(TH)}	$V_{GS} = V_{DS}$, $I_D = 250\mu A$ (Figure 10)		1	-	-	V
Zero Gate Voltage Drain Current	I _{DSS}	V _{DS} = 25V, V _{GS} = 0V		-	-	1	μΑ
		V _{DS} = 25V, V _{GS} = 0V, T _A = 150 ^o C		-	-	250	μΑ
Gate to Source Leakage Current	I _{GSS}	V _{GS} = ±20V		-	-	±100	nA
Drain to Source On Resistance	r _{DS(ON)}	I _D = 10A, V _{GS} =	I _D = 10A, V _{GS} = 4.5V (Figures 9,14)		0.017	0.018	Ω
		I _D = 10A, V _{GS} =	5V	-	0.015	0.017	Ω
		I _D = 10A, V _{GS} = 10V		-	0.011	0.013	Ω
Turn-On Time	toN	V_{DD} = 15V, $I_{D} \cong 10$ A, R_{L} = 1.5 Ω , V_{GS} = 5V, R_{GS} = 6.8 Ω (Figure 15)		-	-	115	ns
Turn-On Delay Time	t _d (ON)			-	15	-	ns
Rise Time	t _r			-	61	-	ns
Turn-Off Delay Time	t _d (OFF)			-	33	-	ns
Fall Time	t _f			-	36	-	ns
Turn-Off Time	tOFF			-	-	105	ns
Total Gate Charge	Q _{g(TOT)}	V _{GS} = 0V to 10V	$V_{DD} = 15V, I_{D} \cong 10A,$	-	39	47	nC
Gate Charge at 5V	Q _{g(5)}	V _{GS} = 0V to 5V	$V_{GS} = 0V \text{ to } 5V$ $R_{L} = 1.5\Omega$, $I_{g(REF)} = 1.0\text{mA}$		22	26	nC
Threshold Gate Charge	Q _{g(TH)}	V _{GS} = 0V to 1V	(Figure 13)	-	1.53	1.85	nC
Gate to Source Gate Charge	Q _{gs}			-	4.00	-	nC
Gate to Drain "Miller" Charge	Q _{gd}			-	9.50	-	nC
Input Capacitance	C _{ISS}	V _{DS} = 25V, V _{GS} = 0V, f = 1MHz (Figure 12)		-	1605	-	pF
Output Capacitance	Coss			-	685	-	pF
Reverse Transfer Capacitance	C _{RSS}			-	115	-	pF
Thermal Resistance Junction to Ambient	$R_{\theta JA}$	Pad Area = 0.76 in ² (Note 2)		-	-	50	°C/W
Pad Area = $0.054 \text{ in}^2 \text{ (S)}$		in ² (See TB377)	-	-	143.4	°C/W	
		Pad Area = 0.011	5 in ² (See TB377)	-	-	177.3	oC/W

Source to Drain Diode Specifications

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Source to Drain Diode Voltage	V _{SD}	I _{SD} = 10A		-	1.25	V
		I _{SD} = 2.3A	-	-	1.1	V
Reverse Recovery Time	t _{rr}	$I_{SD} = 2.3A$, $dI_{SD}/dt = 100A/\mu s$	-	-	57	ns
Reverse Recovered Charge	Q _{RR}	$I_{SD} = 2.3A$, $dI_{SD}/dt = 100A/\mu s$	-	-	81	nC

NOTES:

- 50°C/W measured using FR-4 board with 0.76 in² footprint at 10 seconds.
 177.3°C/W measured using FR-4 board with 0.0115 in² footprint at 1000 seconds.

Typical Performance Curves

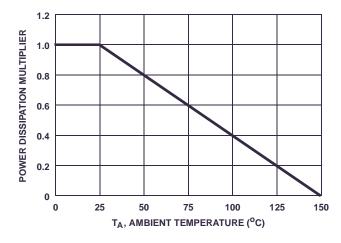


FIGURE 1. NORMALIZED POWER DISSIPATION vs AMBIENT TEMPERATURE

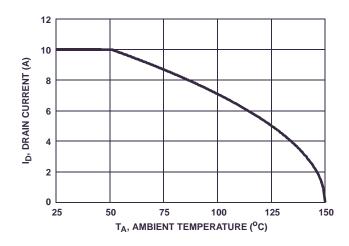


FIGURE 2. MAXIMUM CONTINUOUS DRAIN CURRENT vs AMBIENT TEMPERATURE

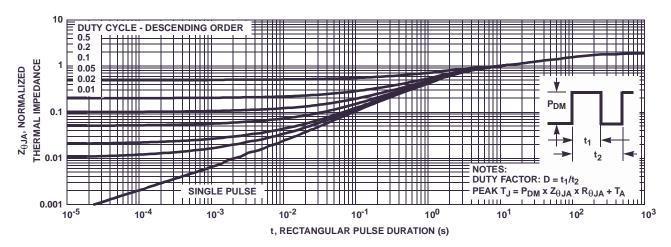


FIGURE 3. NORMALIZED MAXIMUM TRANSIENT THERMAL IMPEDANCE

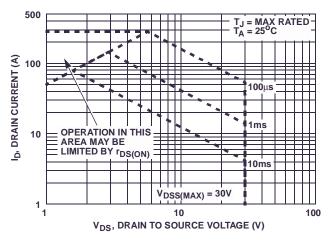


FIGURE 4. FORWARD BIAS SAFE OPERATING AREA

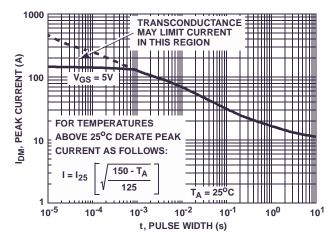
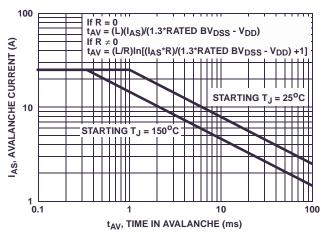


FIGURE 5. PEAK CURRENT CAPABILITY

Typical Performance Curves (Continued)



NOTE: Refer to Fairchild Application Notes AN9321 and AN9322.

FIGURE 6. UNCLAMPED INDUCTIVE SWITCHING CAPABILITY

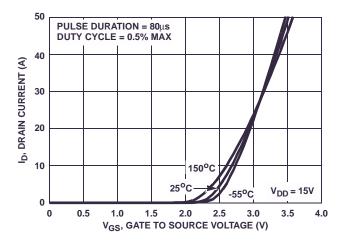


FIGURE 8. TRANSFER CHARACTERISTICS

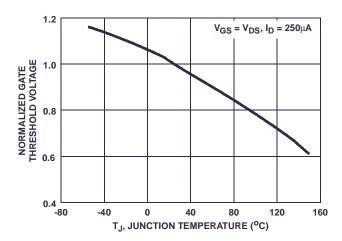


FIGURE 10. NORMALIZED GATE THRESHOLD VOLTAGE vs JUNCTION TEMPERATURE

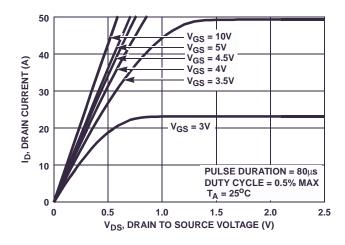


FIGURE 7. SATURATION CHARACTERISTICS

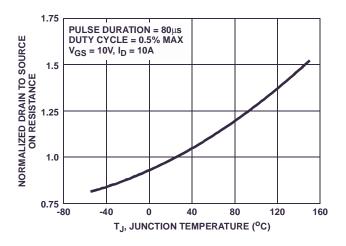


FIGURE 9. NORMALIZED DRAIN TO SOURCE ON RESISTANCE vs JUNCTION TEMPERATURE

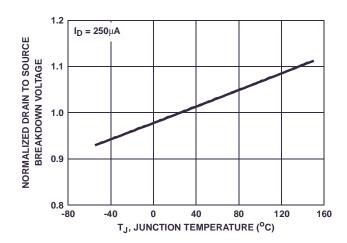


FIGURE 11. NORMALIZED DRAIN TO SOURCE BREAKDOWN VOLTAGE vs JUNCTION TEMPERATURE

Typical Performance Curves (Continued)

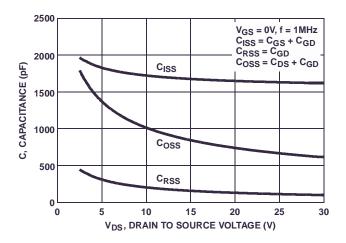


FIGURE 12. CAPACITANCE vs DRAIN TO SOURCE VOLTAGE

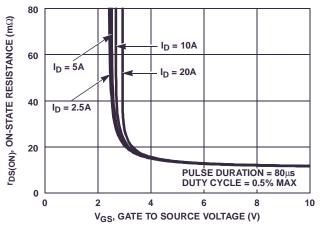
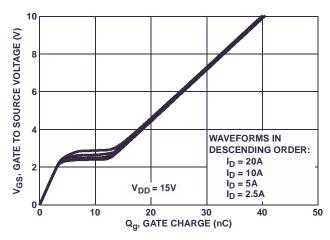


FIGURE 14. DRAIN TO SOURCE ON RESISTANCE vs GATE VOLTAGE AND DRAIN CURRENT



NOTE: Refer to Fairchild Application Notes 7254 and 7260.

FIGURE 13. GATE CHARGE WAVEFORMS FOR CONSTANT GATE CURRENT

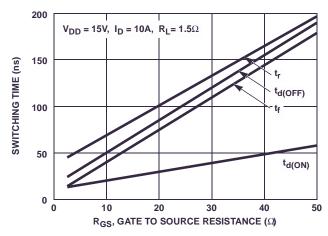


FIGURE 15. SWITCHING TIME vs GATE RESISTANCE

Test Circuits and Waveforms

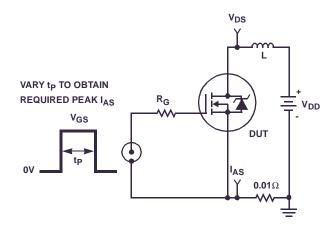


FIGURE 16. UNCLAMPED ENERGY TEST CIRCUIT

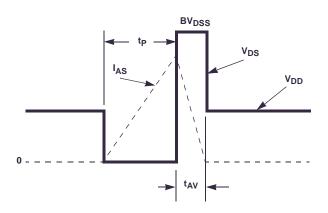


FIGURE 17. UNCLAMPED ENERGY WAVEFORMS

Test Circuits and Waveforms (Continued)

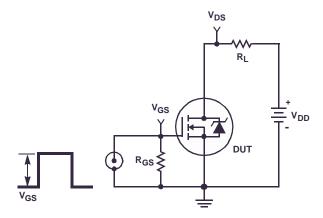


FIGURE 18. SWITCHING TIME TEST CIRCUIT

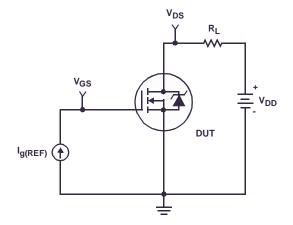


FIGURE 20. GATE CHARGE TEST CIRCUIT

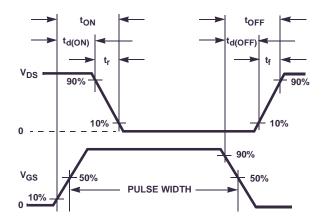


FIGURE 19. SWITCHING TIME WAVEFORM

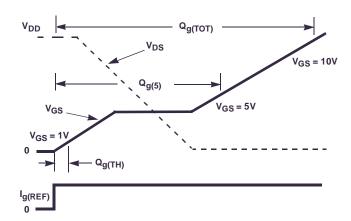


FIGURE 21. GATE CHARGE WAVEFORMS

Thermal Resistance vs Mounting Pad Area

The maximum rated junction temperature T_{JMAX} constrains the maximum allowable device power dissipation P_{Dmax} in an application. The application ambient temperature T_A (^{O}C) and thermal impedance $Z_{\theta JA}$ ($^{O}C/W$) must be reviewed to ensure that T_{JMAX} (^{O}C) is never exceeded. Equation 1 mathematically represents the relationship.

$$P_{DMAX} = \frac{(T_{JMAX} - T_A)}{Z_{\theta JA}}$$
 (EQ. 1)

Fairchild provides thermal information to assist the designer's preliminary application evaluation. Precise determination of P_{DMAX} is complex and influenced by many factors:

- PC heat sink area and location (top and bottom), copper leads and mounting pad area.
- 2. Air Flow, board orientation and type.
- 3. Power pulse width and duty factor.

Figure 22 addresses these points by depicting $R_{\theta JA}$ values vs. top copper (component side) heat sink area. The measurements were performed in still air using a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power.

Figure 22 also displays the two $R_{\theta JA}$ values listed in the Electrical Specifications table. The two points were chosen to graphically depict the compromise between copper board area, thermal resistance and ultimately power dissipation.

Thermal resistance values corresponding to other component side copper areas can be obtained from Figure 22 or by calculation using Equation 2. Area in Equation 2 is the top copper area including the gate and source pads.

$$R_{\theta JA} = 79.3 - 21.8 \times ln(Area)$$
 (EQ. 2)

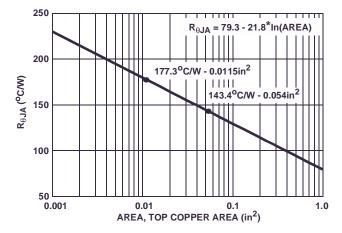


FIGURE 22. THERMAL RESISTANCE vs MOUNTING PAD AREA

Figure 22 provides the necessary information for steady state junction temperature or power dissipation calculations. Transient pulse applications are best studied using the Fairchild device SPICE thermal model.

PSPICE Electrical Model

SUBCKT HUF76131 2 1 3: rev 12/31/97 CA 12 8 2.22-9 CB 15 14 2.13e-9 **LDRAIN** CIN 6 8 1.52e-9 **DPLCAP** DRAIN 5 **-o** 2 10 RLDRAIN DBODY 7 5 DBODYMOD ≷RSLC1 DBREAK 5 11 DBREAKMOD DBREAK T 51 **DPLCAP 10 5 DPLCAPMOD** RSLC₂ **ESLC** 11 EBREAK 11 7 17 18 37.4 50 EDS 14 8 5 8 1 EGS 13 8 6 8 1 17 18 DBODY **≻**RDRAIN 6 8 **EBREAK ESG** ESG 6 10 6 8 1 **EVTHRES** EVTHRES 6 21 19 8 1 16 21 <u>19</u> 8 EVTEMP 20 6 18 22 1 **MWEAK LGATE EVTEMP RGATE** GATE 18 **←**_MMED 22 IT 8 17 1 20 MSTRC **RLGATE** LDRAIN 2 5 1e-9 LSOURCE LGATE 1 9 1.04e-9 CIN SOURCE 8 m LSOURCE 3 7 1.29e-10 3 **RSOURCE** MMED 16 6 8 8 MMEDMOD RLSOURCE MSTRO 16 6 8 8 MSTROMOD S1A S2A MWEAK 16 21 8 8 MWEAKMOD **RBREAK** 13 8 15 14 17 18 13 RBREAK 17 18 RBREAKMOD 1 RDRAIN 50 16 RDRAINMOD 1.94e-3 S1B **RVTEMP** RGATE 9 20 2.20 13 CB 19 RLDRAIN 2510 CA IT 14 **RLGATE 1 9 10.4** VRAT **RLSOURCE 371.29** <u>5</u> **EGS EDS** RSLC1 5 51 RSLCMOD 1e-6 RSLC2 5 50 1e3 8 RSOURCE 8 7 RSOURCEMOD 8.75e-3 **RVTHRES 22 8 RVTHRESMOD 1 RVTHRES RVTEMP 18 19 RVTEMPMOD 1** S1A 6 12 13 8 S1AMOD S1B 13 12 13 8 S1BMOD S2A 6 15 14 13 S2AMOD S2B 13 15 14 13 S2BMOD VBAT 22 19 DC 1 ESLC 51 50 VALUE={(V(5,51)/ABS(V(5,51)))*(PWR(V(5,51)/(1e-6*275),3))} .MODEL DBODYMOD D (IS = 2.25e-12 RS = 6.05e-3 IKF=16.00 TRS1 = 1.14e-4 TRS2 = 1.23e-6 CJO = 2.35e-9 TT = 2.71e-8 M = 0.44) MODEL DBREAKMOD D (RS = 1.05e-1 TRS1 = 1.01e-4 TRS2 = 1.11e-7) .MODEL DPLCAPMOD D (CJO = 1.08e-9 IS = 1e-30 N = 10 M = 0.69) .MODEL MMEDMOD NMOS (VTO = 1.89 KP = 5.05 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 2.20) .MODEL MSTROMOD NMOS (VTO = 2.22 KP = 125.00 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u) MODEL MWEAKMOD NMOS (VTO = 1.62 KP = 0.10 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 22.0 RS = 0.1) .MODEL RBREAKMOD RES (TC1 = 9.54e-4 TC2 = 1.07e-7) .MODEL RDRAINMOD RES (TC1 = 1.61e-2 TC2 = 5.17e-5) .MODEL RSLCMOD RES (TC1 = 1.03e-5 TC2 = 7.67e-7) .MODEL RSOURCEMOD RES (TC1 = 0 TC2 = 0) .MODEL RVTHRESMOD RES (TC = -2.81e-3 TC2 = -8.75e-6) .MODEL RVTEMPMOD RES (TC1 = -6.68e-4 TC2 = 8.80e-7) .MODEL S1AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -5.80 VOFF= -1.50) .MODEL S1BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -1.50 VOFF= -5.80) .MODEL S2AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -0.50 VOFF= -0.00) .MODEL S2BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = 0.00 VOFF= -0.50) .ENDS

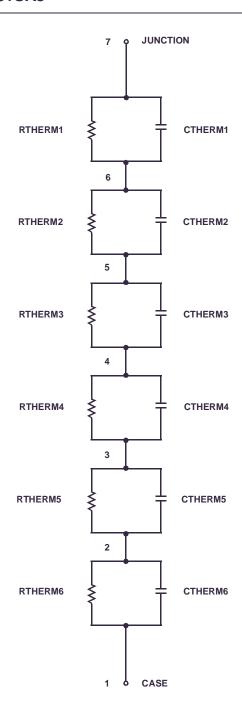
NOTE: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.

SPICE Thermal Model

REV 20 Feb 98

HUF76131

CTHERM1 7 6 3.75e-4
CTHERM2 6 5 3.05e-3
CTHERM3 5 4 3.70e-2
CTHERM4 4 3 2.52e-2
CTHERM5 3 2 8.50e-2
CTHERM6 2 1 7.95e-1
RTHERM1 7 6 3.95e-2
RTHERM2 6 5 2.50e-1
RTHERM3 5 4 4.00e-1
RTHERM4 4 3 6.35
RTHERM5 3 2 2.02e1
RTHERM6 2 1 4.80e1



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CoolFET™	FASTr™	MicroFET™	PowerTrench [®]	SuperSOT™-6
CROSSVOLT™	FRFET™	MicroPak™	QFET™	SuperSOT™-8
DOME™	GlobalOptoisolator™	MICROWIRE™	QS™	SyncFET™
EcoSPARK™	GTO™	MSX™	QT Optoelectronics™	TinyLogic [®]
E ² CMOS™	HiSeC™	MSXPro™	Quiet Series™	TruTranslation™
EnSigna™	I^2C^{TM}	OCX^{TM}	RapidConfigure™	UHC™
Across the board	. Around the world.™	OCXPro™	RapidConnect™	UltraFET [®]
The Power Franc	hise™	OPTOLOGIC [®]	SILENT SWITCHER®	VCX^{TM}
Programmable Active Droop™		OPTOPLANAR™	SMART START™	

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