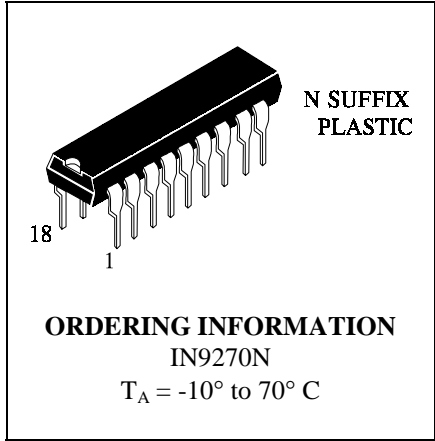**IN9270**

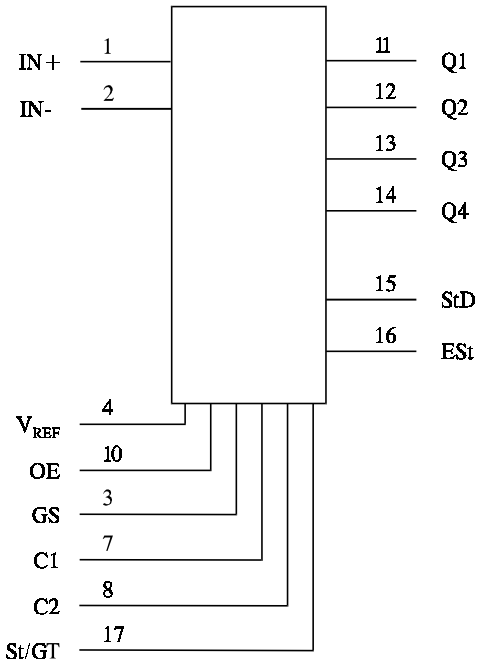# DTMF RECEIVER
High-Performance Silicon-Gate CMOS

The IN9270 is a complete DTMF receiver integrating both the bandsplit filter and digital decoder functions. The filter section uses switched capacitor techniques for high- and low-group filters and dial-tone rejection. Digital counting techniques are employed in the decoder to detect and decode all 16 DTMF tone-pairs into a 4-bit code. External component count is minimized by on-chip provision of a differential input amplifier, clock-oscillator and latched 3-state bus interface.

- Complete receiver in an 18-pin package.
- Excellent performance.
- CMOS, single 5 volt operation.
- Minimum board area.
- Central office quality.
- Low power consumption.

N SUFFIX
PLASTIC

18
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ORDERING INFORMATION
IN9270N
T_A = -10° to 70° C

LOGIC DIAGRAM

IN+ 1
IN- 2
11 Q1
12 Q2
13 Q3
14 Q4
15 StD
16 ESt
V_{REF} 4
OE 10
GS 3
C1 7
C2 8
St/GT 17

PIN 9 = GND
PIN 18 = V_{CC}
PINS 5,6 = NO CONNECTION

PIN ASSIGNMENT

IN+	1●	18	V_{CC}
IN-	2	17	St/GT
GS	3	16	ESt
V_{REF}	4	15	StD
NC*	5	14	Q4
NC*	6	13	Q3
C1	7	12	Q2
C2	8	11	Q1
GND	9	10	OE

* Connect to GND

PIN DESCRIPTIONS

NAME	PIN	DESCRIPTION	
ESt	16	Early steering output. Presents a logic high immediately when the digital algorithm detects a recognizable tone-pair (signal condition). Any momentary loss of signal condition will cause ESt to return to a logic low.	
GS	3	Gain Select. Gives access to output of front-end differential amplifier for connection of feedback resistor.	
IC	5,6	Internal Connection. Must be tied to GND.	
IN+	1	Non-Inverting Input	Connections to the front-end differential amplifier.
IN-	2	Inverting Input	
C1	7	Clock Input	3.579545 MHz crystal connected between these pins completes internal oscillator.
C2	8	Clock Output	
Q1-Q4	11-14	3-state data outputs. When enabled by OE, provide the code corresponding to the last valid tone-pair received.	
StD	15	Delayed steering output. Presents a logic high when a received tone-pair has been registered and the output latch updated; returns to logic low when the voltage on St/GT falls below V_{TSt}.	
St/GT	17	Steering input/guard time output (bi-directional). A voltage greater than V_{TSt}, detected at St causes the device to register the detected tone-pair and update the output latch. A voltage less than V_{TSt} frees the device to accept a new tone-pair. The GT output acts to reset the external steering time-constant; its state is a function of ESt and the voltage on St.	
OE	10	3-state output enable (input). Logic high enables the outputs Q1-Q4. Internal pull-up.	
V_{CC}	18	Positive power supply, +5 V.	
V_{REF}	4	Reference voltage output, nominally $V_{CC}/2$. May be used to bias the inputs at mid-rail.	
GND	9	Negative power supply, normally connected to 0 V.	

FUNCTIONAL DESCRIPTION

The IN9270 monolithic DTMF receivers offer small size, low power consumption and high performance. The architecture consists of a bandsplit filter section, which separates the high and low tones of a receiver pair, followed by a digital counting section which verifies the frequency and duration of the received tones before passing the corresponding code to the output bus.

Filter Section

Separation of the low-group and high-group tones is achieved byapplying the dual-tone signal to the inputs of two filters - a sixth order for the high group and an eight order for the low group. The band-widths of which correspond to the bands enclosing the low-group and high-group tones (see Figure 1). The filter section also incorporates notches at 350 Hz and 440 Hz for exceptional dial-tone rejection. Each filter output is followed by a second order switched-capacitor section which smooths the signals prior to limiting. Limiting is performed by high-gain comparators which are provided with hysteresis to prevent detection of unwanted low-level signals and noise; the outputs of the comparators provide full-rail logic swings at the frequencies of the incoming tones.

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Decoder Section

The decoder uses digital counting techniques to determine the frequencies of the limited tones and to verify that they correspond to standard DTMF frequencies. A complex averaging algorithm protects against tone simulation by extraneous signals, such as voice, while providing tolerance to small frequency deviations and variations. This averaging algorithm has been developed to ensure an optimum combination of immunity to "talk-off" and tolerance to the presence of interfering signals ("third tones") and noise. When the detector recognizes the simultaneous presence of two valid tones (referred to as "signal condition" in some industry specifications), it raises the "early steering" flag (ESt). Any subsequent loss of signal-condition will cause Est to fall.

Steering Circuit

Before registration of a decoded tone-pair, the receiver checks for a valid signal duration (referred to as "character-recognition-condition"). This check is performed by an external RC time-constant driven by ESt. A logic high on ESt causes V_C (see Figure 2) to rise as the capacitor discharges. Provided signal-condition is maintained (ESt remains high) for the validation period (t_{GTP}), V_C reaches the threshold (V_{TSt}) of the steering logic to register the tone-pair, latching its corresponding 4-bit code (see Figure 3) into the output latch. At this point, the GT output is activated and drives V_C to V_{CC}. GT continues to drive high as long as ESt remains high. Finally after a short delay to allow the output latch to settle, the "delayed-steering" output flag, StD, goes high, signaling that a received tone-pair has been registered. The contents of the output latch are made available on the 4-bit output bus by raising the 3-state control input (OE) to a logic high. The steering circuit works in reverse to validate the interdigit pause between signals. Thus, as well as rejecting signals too short to be considered valid, the receiver will tolerate signal interruptions ("drop-out") too short to be considered a valid pause. The facility, together with the capability of selecting the steering time-constants externally, allows the designer to tailor performance to meet a wide variety of system requirements.

Guard Time Adjustment

In many situations not requiring independent selection of receive and pause, the simple steering circuit of Figure 2 is applicable. Component values are chosen according to the following formula:

$t_{REC} = t_{DP} + t_{GTP}$

$t_{ID} = t_{DA} + t_{GTA}$

The value of t_{DP} is a parameter of the device and t_{REC} is the minimum signal duration to be recognized by the receiver. A value for C of 0.1 µF is recommended for most applications, leaving R to be selected by the designer. For example, a suitable value of R for a t_{REC} of 40 ms would be 300 k.

Different steering arrangements may be esed to select independently the guard-times for tone-present (t_{GTP}) and tone-absent (t_{GTA}). This may be necessary to meet system specifications which place both accept and reject limits on both tone duration and inter-digital pause.

Guard-time adjustment also allows the designer to tailor system parameters such as talk-off and noise immunity. Increasing t_{REC} improves talk-off performance, since it reduces the probability that tones simulated by speech will maintain signal condition for long enough to be registered. On the other hand, a relatively short t_{REC} with a long t_{DO} would be appropriate for extremely noisy environments where fast acquisition time and immunity to drop-outs would be requirements. Design information for guard-time adjustment is show in Figure 4.

Input Configuration

The input arrangement of the IN9270 provides a differential-input operational amplifier as well as a bias source (V_{REF}) which is used to bias the inputs at mid-rail.

Provision is made for connection of a feedback resistor to the op-amp output (GS) for adjustment of gain.

In a single-ended configuration, the input pins are connected as shown in Figure 5 with the op-amp connected for unity gain and V_{REF} biasing the input at $1/2V_{CC}$. Figure 6 shows the differential configuration, which permits the adjustment of gain with the feedback resistor R_5.

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MAXIMUM RATINGS[*]

Symbol	Parameter	Value	Unit
V_{CC}	DC Supply Voltage (Referenced to GND)	-0.3 to +6.0	V
V_{IN}	DC Input Voltage (Referenced to GND)	-0.3 to V_{CC} +0.3	V
I_{IN}	DC InputCurrent, per Pin	10	mA
P_D	Power Dissipation in Still Air, Plastic DIP[**]	500	mW
Tstg	Storage Temperature	-65 to +150	°C

[*] Maximum Ratings are those values beyond which damage to the device may occur.
Functional operation should be restricted to the Recommended Operating Conditions.
[**] Derating: -10 $^{mW}/_{°C}$ from 65°C to 70°C.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Max	Unit
V_{CC}	DC Supply Voltage (Referenced to GND)	4.75	5.25	V
V_{IN}	DC Input Voltage (Referenced to GND)	1.5	3.5	V
T_A	Operating Temperature	-10	+70	°C
P_O	Power Consumption (f = 3.579 MHz, V_{CC} = 5 V)	-	45	mW
t_r, t_f	Input Rise and Fall Time	0	110	ns

      This device contains protection circuitry to guard against damage due to high static voltages or electric fields. However, precautions must be taken to avoid applications of any voltage higher than maximum rated voltages to this high-impedance circuit. For proper operation, V_{IN} and V_{OUT} should be constrained to the range GND≤(V_{IN} or V_{OUT})≤V_{CC}.

      Unused inputs must always be tied to an appropriate logic voltage level (e.g., either GND or V_{CC}). Unused outputs must be left open.

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DC ELECTRICAL CHARACTERISTICS(Voltages Referenced to GND, V_{CC} = 5 V ± 5%, T_A = -10 to +70°C)

Symbol	Parameter	Test Conditions	Guaranteed Limits			Unit
			Min	Typ	Max	
V_{IH}	Minimum High-Level Input Voltage	V_{OUT} =5 V	3.6			V
V_{IL}	Maximum Low-Level Input Voltage	V_{OUT} =5 V			1.4	V
V_{OH}	Minimum High-Level Output Voltage	No Load	4.97			V
V_{OL}	Maximum Low-Level Output Voltage	No Load			0.05	V
I_{IN}	Maximum Input Leakage Current	V_{IN}=V_{CC} or GND			±0.1	µA
I_{SO}	Maximum Pull Up (Source) Current	OE = 0 V			24	µA
I_{OL}	Minimum Output-Low (Sink) Current	V_{OUT} =0.4 V	0.8			mA
I_{OH}	Minimum Output-High (Source) Current	V_{OUT} =4.6 V	0.35			mA
I_{CC}	Maximum Quiescent Supply Current (per Package)	V_{IN} =V_{CC}			11	mA
V_{TSt}	Steering Threshold Voltage		2.2		2.5	V
R_{IN}	Input Impedance (Signal Inputs 1,2)	@ 1 KHz	8			MΩ
V_{REF}	Output Voltage	No Load	2.4		2.8	V
R_{OR}	Output Resistance			10		kΩ
I_{OZ}	Maximum Three-State Leakage current	Output in High-Impedance State V_{IN} =V_{IL} V_{OUT}=V_{CC} or GND			±0.1	µA

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OPERATING CHARACTERISTICS
Gain Setting Amplifier

Symbol	Parameter	Test Conditions	Typ	Unit
I_N	Input Leakage Current	GND< V_{IN} < V_{CC}	±100	nA
R_{IN}	Input Resistance		10	MΩ
V_{OS}	Input Offset Voltage		±25	mV
PSRR	Power Supply Rejection	1 KHz	60	dB
CMRR	Common Mode Rejection	-3.0 V < V_{IN} < 3.0 V	60	dB
A_{VOL}	DC Open Loop Voltage Gain		65	dB
f_C	Open Loop Unity Gain Bandwidth		1.5	MHz
V_O	Output Voltage Swing	R_L ≥100 KΩ to GND	4.5	V_{PP}
C_L	Tolerable Resistive Load (GS)		100	pF
$R_{L:}$	Tolerable Resistive Load (GS)		50	KΩ
V_{CM}	Common Mode Range	No Load	3.0	V_{PP}

AC ELECTRICAL CHARACTERISTICS(All Voltages referenced to GND. V_{CC} = 5.0 V, GND = 0 V, T_A = -10 to +70°C, F_{CLK} = 3.579545 Mhz, using test circuit of Figure 5)

Parameter		Guaranteed Limits			Unit	Notes
		Min	Typ	Max		
SIGNAL CONDITION						
Valid Input Signal	MIN			-29	dBm	1,2,3,5,6,9
Level (each tone of				27.5	m V_{RMS}	1,2,3,5,6,9
composite signal)	MAX	+1			dBm	1,2,3,5,6,9
		883			m V_{RMS}	
NON-ACCEPT LEVEL						
Freq. Deviation Accept Limit				±1.5% ±2 Hz	Nom.	2,3,5,9
Freq. Deviation Reject Limit		±3.5%			Nom.	2,3,5
Third Tone Tolerance			-2.5		dB	2,3,4,5,7,9,10
Dial Tone Tolerance			+18		dB	2,3,4,5,8,9,10

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TIMING REQUIREMENTS (All Voltages referenced to GND. V_{CC} = 5.0 V, GND = 0 V, T_A = -10 to +70°C, F_{CLK} = 3.579545 Mhz, using test circuit of Figure 5)

Symbol	Parameter		Guaranteed Limits		Unit	Notes
			Min	Max		
t_{DP}	Tone Present Detection Time (Figure 7)		5	18	ms	Refer to
t_{DA}	Tone Absent Detection Time (Figure 7)		0.5	10	ms	Fig. 7
t_{REC}	Maximum Tone Detection Accept (Figure 7)			40	ms	(User Adjustable)
$\overline{t_{REC}}$	Minimum Tone Detection Reject (Figure 7)		20		ms	
t_{ID}	Maximum Interdigit Pause Accept (Figure 7)			40	ms	Refer to "Guard Time
t_{DO}	Minimum Interdigit Pause Reject (Figure 7)		20		ms	Adjustment"
t_{PQ}	Maximum Propagation Delay (St to Q) (Figure 7)			11	µs	OE = V_{CC}
t_{PSED}	Maximum Propagation Delay (St to StD) (Figure 7)			16	µs	
t_{QSED}	Maximum Output Data Set Up (Q to StD) (Figure 7)			5	µs	
t_{PTE}	Maximum Propagation Delay	ENABLE		75	ns	R_L= 10 KΩ C_L= 50 pF
t_{PTD}	(OE to Q) (Figure 7)	DISABLE		460	ns	
f_{CLK}	Crystal/Clock Frequency		3.5759	3.581	MHz	
C_{LO}	Clock Output (C2)	Capacitive Load		30	pF	

Notes:
1. dBm = decibels above or below a reference power of 1 mW into a 600 Ω load.
2. Digit sequence consists of all 16 DTMF tones.
3. Tone duration = 40 ms, Tone pause = 40 ms.
4. Nominal DTMF frequencies are used.
5. Both tones in the composite signal have an equal amlitude.
6. Tone pair is deviated by ±1.5% ±2 Hz.
7. Bandwidth limited (3 KHz) Gaussian Noise.
8. The precise dial tone frequencies are (350 Hz and 440 Hz) ±2%.
9. For an error rate of less than 1 in 10,000.
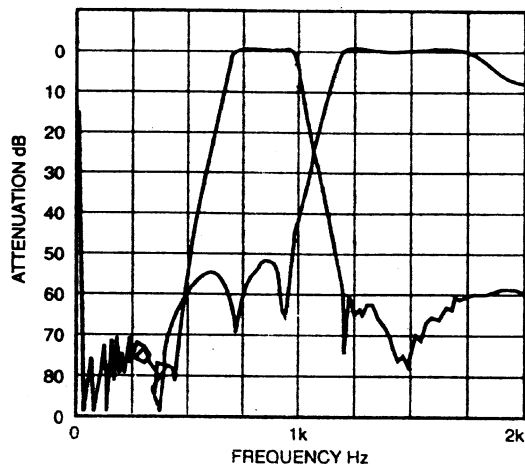10. Referenced to the lowest level frequency component in DTMF signal.
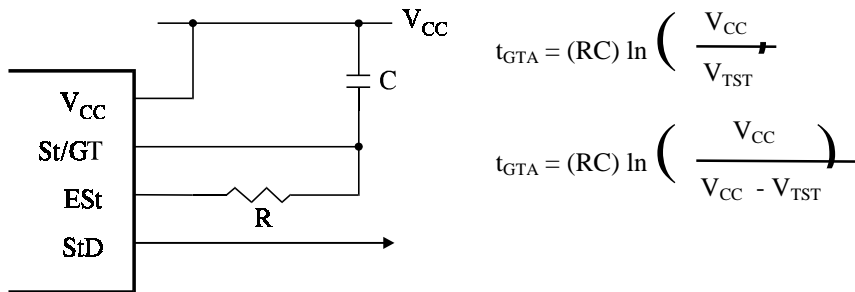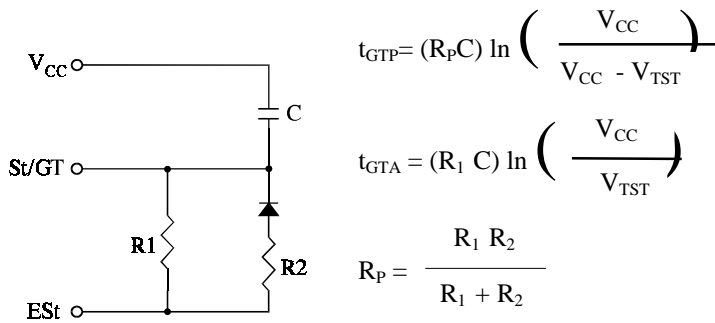

Figure 1. Typical Filter Characteristic

$$t_{GTA} = (RC) \ln \left(\frac{V_{CC}}{V_{TST}} \right)$$

$$t_{GTA} = (RC) \ln \left(\frac{V_{CC}}{V_{CC} - V_{TST}} \right)$$

Figure 2. Basic Steering Circuit

F_{LOW}	F_{HIGH}	KEY	OE	Q4	Q3	Q2	Q1
697	1209	1	H	0	0	0	1
697	1336	2	H	0	0	1	0
697	1477	3	H	0	0	1	1
770	1209	4	H	0	1	0	0
770	1336	5	H	0	1	0	1
770	1477	6	H	0	1	1	0
852	1209	7	H	0	1	1	1
852	1336	8	H	1	0	0	0
852	1477	9	H	1	0	0	1
941	1336	0	H	1	0	1	0
941	1209	*	H	1	0	1	1
941	1477	#	H	1	1	0	0
697	1633	A	H	1	1	0	1
770	1633	B	H	1	1	1	0
852	1633	C	H	1	1	1	1
941	1633	D	H	0	0	0	0
		ANY	L	Z	Z	Z	Z

"L = Logic Low, H = Logic High, Z = High Impedance"

Figure 3. Logic Table

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$$t_{GTP} = (R_P C) \ln \left(\frac{V_{CC}}{V_{CC} - V_{TST}} \right)$$

$$t_{GTA} = (R_1 C) \ln \left(\frac{V_{CC}}{V_{TST}} \right)$$

$$R_P = \frac{R_1 R_2}{R_1 + R_2}$$

a) Descreasing t_{GTP} ($t_{GTP} < t_{GTA}$)

$$t_{GTP} = (R_1 C) \ln \left(\frac{V_{CC}}{V_{CC} - V_{TST}} \right)$$

$$t_{GTA} = (R_P C) \ln \left(\frac{V_{CC}}{V_{TST}} \right)$$

$$R_P = \frac{R_1 R_2}{R_1 + R_2}$$
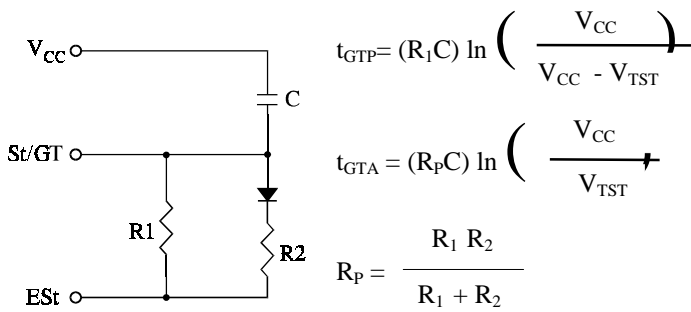
b) Descreasing t_{GTA} ($t_{GTP} > t_{GTA}$)

Figure 4. Guard Time Adjustment
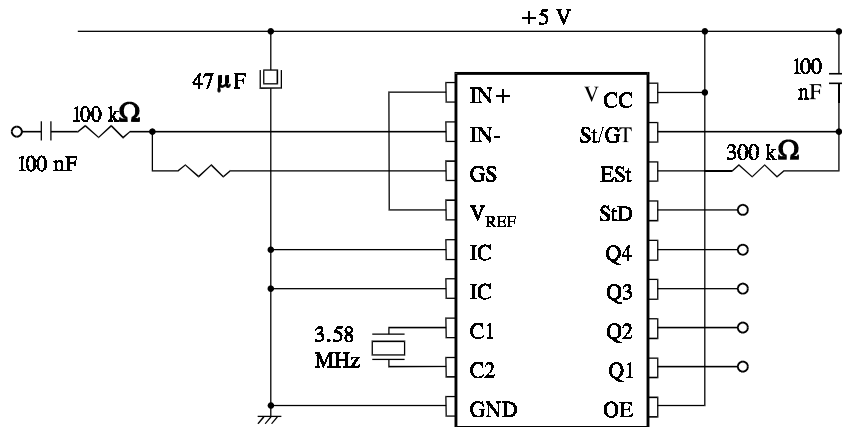
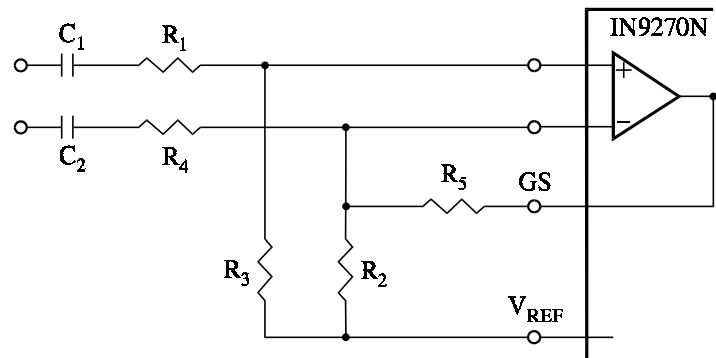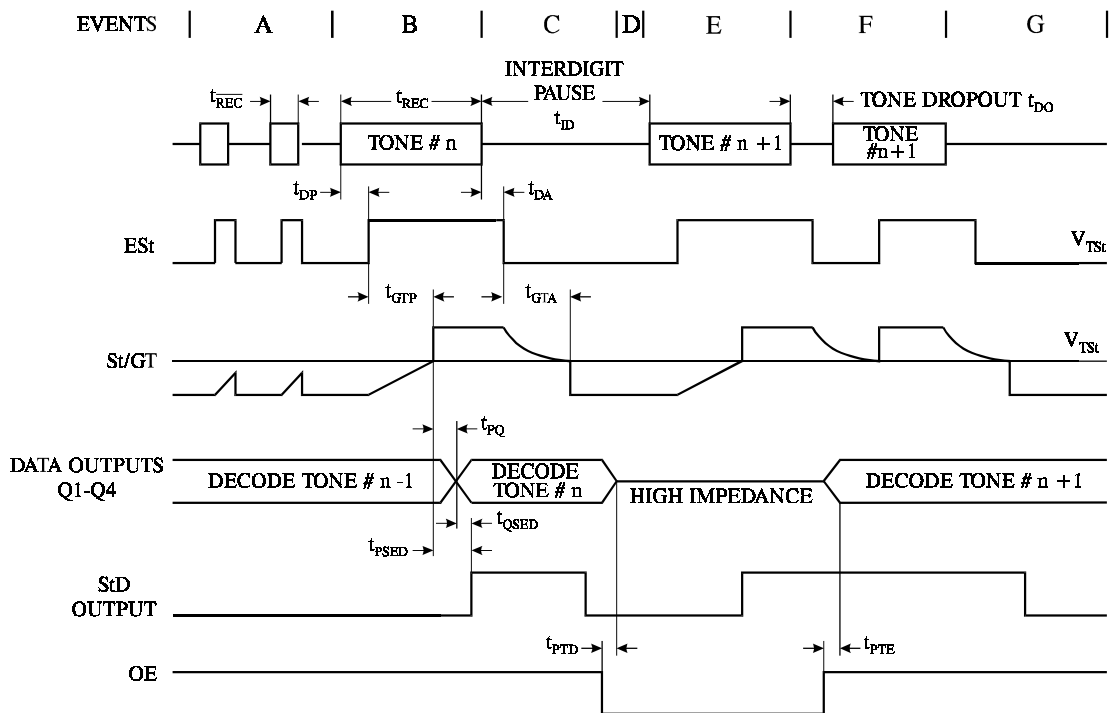**Figure 5. Single Ended Input Configuration**

Figure 6. Differential Input Configuration

TIMING DIAGRAM

A. Short tone bursts: detected. Tone duration is invalid.
B. Tone #n is detected. Tone duration is valid. Decoded to outputs.
C. End of tone #n is detected and validated.
D. 3 State outputs disabled (high impedance).
E. Tone #n+1 detected. Tone duration is valid. Decoded to outputs.
F. Tristate outputs are enabled. Acceptable drop out of tone #n+1 does not register at outputs.
G. End of tone #n+1 is detected and validated.

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EXPANDED LOGIC DIAGRAM