



General Description

The MAX1160 10-bit, monolithic analog-to-digital converter (ADC) is capable of 20Msps minimum word rates. An on-board track/hold ensures excellent dynamic performance without the need for external components. A 5pF input capacitance minimizes drive requirement problems.

Inputs and outputs are TTL compatible. An overrange output is provided to indicate overflow conditions. Output data format is straight binary. Power dissipation is low at only 1W with +5V and -5.2V power-supply voltages. The MAX1160 also accepts wide ±2V input voltages.

The MAX1160 is available in 28-pin DIP and SO packages in the commercial temperature range.

_Applications

Medical Imaging Professional Video Radar Receivers Instrumentation Digital Communications

_Features

- Monolithic 20Msps Converter
- On-Chip Track/Hold
- Bipolar, ±2V Analog Input
- ♦ 60dB SNR at 1MHz Input
- ♦ 5pF Input Capacitance
- TTL Outputs

Ordering Information

Pin Configuration

PART	TEMP. RANGE	PIN-PACKAGE
MAX1160ACPI	0°C to +70°C	28 Wide Plastic DIP
MAX1160BCPI	0°C to +70°C	28 Wide Plastic DIP
MAX1160ACWI	0°C to +70°C	28 SO
MAX1160BCWI	0°C to +70°C	28 SO



_Functional Diagram

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For the latest literature: http://www.maxim-ic.com, or phone 1-800-998-8800

ABSOLUTE MAXIMUM RATINGS

V _{CC}	6V
VEE	-6V
Analog Input	VFB ≤ VIN ≤ VFT
VFT, VFB	3V, -3V
Reference-Ladder Current	
CLK Input	Усс
Digital Outputs	30mA to -30mA

Continuous Power Dissipation ($T_A = +70^{\circ}C$)	
Plastic DIP	1.14W
SO	1W
Operating Temperature Range	0°C to +70°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10sec)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +5.0V, V_{EE} = -5.2V, DV_{CC} = +5.0V, V_{IN} = \pm 2.0V, VSB = -2.0V, VST = +2.0V, f_{CLK} = 20MHz, 50% clock duty cycle, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	CONDITIONS	TEST LEVEL	MIN	MAX1160A TYP	MAX	MIN	MAX1160 TYP	B MAX	UNITS
Resolution			10			10			Bits
DC ACCURACY (± full scal	e, 250kHz sample rate, T _A	= +25°C)				1			
Integral Nonlinearity				±1.0			±1.5		LSB
Differential Nonlinearity		I		±0.5			±0.75		LSB
No Missing Codes			1	Guaranteed	k		Guarantee	d	
ANALOG INPUT									
Input Voltage Range		VI		±2.0			±2.0		V
Input Bias Current	$V_{IN} = OV$	VI		30	60		30	60	μA
Input Bias Current	$T_A = -55^{\circ}C \text{ to } +125^{\circ}C$	VI			75			75	μA
Input Resistance		VI	100	300		100	300		kΩ
Input Resistance	$T_A = -55^{\circ}C \text{ to } +125^{\circ}C$	VI	75	300		75	300		kΩ
Input Capacitance		V		5			5		рF
Input Bandwidth	3dB small signal	V		120			120		MHz
Positive Full-Scale Error		V		±2.0			±2.0		LSB
Negative Full-Scale Error		V		±2.0			±2.0		LSB
REFERENCE INPUT									
Reference-Ladder Resistance		VI	500	800		500	800		Ω
Reference-Ladder Tempco		V		0.8			0.8		Ω/°C
TIMING CHARACTERISTIC	S								
Maximum Conversion Rate		VI	20			20			MHz
Overvoltage Recovery Time		V		20			20		ns
Pipeline Delay (Latency)		VI			1			1	Clock Cycle
Output Delay	$T_A = +25^{\circ}C$	V		14	18		14	18	ns
Aperture Delay Time	$T_A = +25^{\circ}C$	V		1			1		ns
Aperture Jitter Time	$T_A = +25^{\circ}C$	V		5			5		ps-RMS
Acquisition Time	$T_A = +25^{\circ}C$	V		20			20		ns

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +5.0V, V_{EE} = -5.2V, DV_{CC} = +5.0V, V_{IN} = \pm 2.0V, VSB = -2.0V, VST = +2.0V, f_{CLK} = 20MHz, 50% clock duty cycle, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	CONDITIONS		TEST LEVEL	M MIN	AX1160A TYP MAX	M MIN	AX1160B TYP MAX	UNITS
DYNAMIC PERFORMANC	E							1
	f _{IN} = 1MHz				9.2		8.7	
Effective Number of Bits	f _{IN} = 3.58MHz				8.8		8.3	Bits
	f _{IN} = 10MHz				7.5		7.0	1
		$T_A = +25^{\circ}C$	I	57	60	54	57	
	$f_{IN} = 1MHz$	$T_A = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_A = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	55	58	52	55	-
Signal-to-Noise Ratio		$T_A = +25^{\circ}C$	I	56	58	53	55	1
(without harmonics) (SNR)	f _{IN} = 3.58MHz	$T_{A} = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_{A} = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	54	56	51	53	dB
		$T_A = +25^{\circ}C$	I	50	53	47	49	-
	f _{IN} = 10MHz	$T_{A} = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_{A} = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	47	50	44	46	
		$T_A = +25^{\circ}C$	1	57	60	54	57	
Total Harmonic Distortion (THD)	$f_{IN} = 1MHz$	$T_{A} = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_{A} = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	54	57	51	54	
	f _{IN} = 3.58MHz	$T_A = +25^{\circ}C$	I	56	58	53	55	1
		$T_A = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_A = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	53	55	50	52	dB
	f _{IN} = 10MHz	$T_A = +25^{\circ}C$	I	46	48	43	45	-
		$T_{A} = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_{A} = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	45	47	42	44	
	f _{IN} = 1MHz	$T_A = +25^{\circ}C$	I	55	57	52	54	
		$T_A = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_A = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	52		49		-
Signal-to-Noise and		$T_A = +25^{\circ}C$	I	54	55	51	52	
Distortion Ratio (SINAD)	$f_{IN} = 3.58MHz$	$T_A = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_A = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	51		48		dB
		$T_A = +25^{\circ}C$	I	44	47	41	44	1
	$f_{IN} = 10MHz$	$T_A = 0^{\circ}C \text{ to } +70^{\circ}C,$ $T_A = -25^{\circ}C \text{ to } +85^{\circ}C$	IV	43		40		-
Spurious-Free Dynamic Range (SFDR)	f _{IN} = 1MHz	T _A = +25°C	V		67		67	dB
Differential Phase	$f_{IN} = 3.58MHz$ and 4.35MHz	$T_A = +25^{\circ}C$	V		0.2		0.2	Degrees
Differential Gain	$f_{IN} = 3.58MHz$ and 4.35MHz	T _A = +25°C	V		0.5		0.7	%

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +5.0V, V_{EE} = -5.2V, DV_{CC} = +5.0V, V_{IN} = \pm 2.0V, VSB = -2.0V, VST = +2.0V, f_{CLK} = 20MHz, 50% clock duty cycle, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	CONDITIONS	TEST LEVEL	MIN	AX116 TYP	0A MAX	MIN	AX116 TYP	0B MAX	UNITS
DIGITAL INPUTS	L	1	1						
Logic 1 Voltage		V	2.4		4.5	2.4		4.0	V
Logic 0 Voltage		V			0.8			0.8	V
Maximum Input Current Low	T _A = +25°C	IV	0	5	20	0	5	20	μA
Maximum Input Current High	TA = +25°C	IV	0	5	20	0	5	20	μA
Pulse Width Low (CLK)		IV	20			20			ns
Pulse Width High (CLK)		IV	20		300	20		300	ns
DIGITAL OUTPUTS									
Logic 1 Voltage		IV	2.4			2.4			V
Logic 0 Voltage		IV			0.6			0.6	V
POWER-SUPPLY REQUIR	EMENTS								
	Vcc	IV	4.75		5.25	4.75		5.25	
Voltages	DV _{CC}	IV	4.75	5.0	5.25	4.75	5.0	5.25	V
	-V _{EE}	IV	-4.95	-5.2	-5.45	-4.95	-5.2	-5.45	
	Icc	VI		118	145		118	145	
Currents	DICC	VI		40	55		40	55	mA
	-IEE	VI		40	57		40	57	
Power Dissipation		VI		1.0	1.3		1.0	1.3	W
Power-Supply Rejection	$V_{CC} = 5V \pm 0.25V$, $V_{EE} = -5.2V \pm 0.25V$	V		1.0			1.0		LSB

TEST LEVEL CODES

All electrical characteristics are subject to the following conditions: All parameters having min/max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Assurance inspection. Any blank section in the data column indicates that the specification is not tested at the specified condition. Unless otherwise noted, all tests are pulsed; therefore, $T_{\rm I}=T_{\rm C}=T_{\rm A}.$

TEST LEVEL TEST PROCEDURE

Ш

Ш

IV

V

VI

100% production tested at the specified temperature.

100% production tested at TA = +25°C, and sample tested at the specified temperatures.

QA sample tested only at the specified temperatures.

Parameter is guaranteed (but not tested) by design and characterization data.

Parameter is a typical value for information purposes only. 100% production tested at $T_A = +25^{\circ}$ C. Parameter is guaranteed over specified

temperature range.

_Pin Description

PIN	NAME	FUNCTION	
1, 13	DGND	Digital Ground	
2	D0	TTL Output (LSB)	
3–10	D1-D8	TTL Outputs	
11	D9	TTL Output (MSB)	
12	D10	TTL Output Overrange	
14, 28	DVcc	+5V Supply (digital)	
15	CLK	Clock	
16, 27	V _{EE}	-5.2V Supply (analog)	

PIN	NAME	FUNCTION		
17, 26	AGND	Analog Ground		
18, 25	Vcc	+5V Supply (analog)		
19	VFT	Force for Top of Reference Ladder		
20	VST	Sense for Top of Reference Ladder		
21	VIN	Analog Input		
22	VRM	Middle of Voltage Reference Ladder		
23	VSB	Sense for Bottom of Reference Ladder		
24	VFB	Force for Bottom of Reference Ladder		



Typical Operating Characteristics

SIGNAL-TO-NOISE RATIO vs. TOTAL HARMONIC DISTORTION vs. SIGNAL-TO-NOISE AND DISTORTION INPUT FREQUENCY INPUT FREQUENCY vs. INPUT FREQUENCY 80 80 80 f_S = 20Msps f_S = 20Msps f_S = 20Msps 70 70 70 60 60 60 SINAD (dB) SNR (dB) THD (dB) 50 50 50 40 40 40 30 30 30 20 20 20 10 100 1 10 100 1 1 10 100 INPUT FREQUENCY (MHz) INPUT FREQUENCY (MHz) **INPUT FREQUENCY (MHz)** SNR, THD, SINAD vs. SNR, THD, SINAD vs. SPECTRAL RESPONSE SAMPLE RATE TEMPERATURE 80 65 0 f_{IN} = 1MHz $f_S = 20Msps$ SNR SNR f_{IN} = 1MHz 70 SNR, THD 60 -30 (dB) (qB) ۱_{THD} 60 AMPLITUDE (dB) SINAD (SINAD (THD 55 SINAD 50 -60 SNR, THD, SNR, THD, SINAD 50 40 -90 45 30 f_S = 20Msps $f_{IN} = 1MHz$ 40 -120 20 25 75 100 -25 0 50 0 2 8 9 10 1 10 1 3 4 5 6 7 TEMPERATURE (°C) SAMPLE RATE (Msps) INPUT FREQUENCY (MHz)

Detailed Description

The MAX1160 requires few external components to achieve the stated operation and performance. Figure 2 shows the typical interface requirements when using the MAX1160 in normal circuit operation. The following section provides a description of the pin functions, and outlines critical performance criteria to consider for achieving the optimal device performance.

 $(T_A = +25^{\circ}C, unless otherwise noted.)$

Power Supplies and Grounding

The MAX1160 requires -5.2V and +5V analog supply voltages. The +5V supply is common to analog V_{CC} and digital DV_{CC}. A ferrite bead in series with each supply line reduces the transient noise injected into the analog V_{CC}. Connect these beads as close to the device as possible. The connection between the beads and the MAX1160 should not be shared with any other device. Bypass each power-supply pin as close to the device as possible. Use 0.1μ F for V_{EE} and V_{CC}, and 0.01μ F for DV_{CC} (chip capacitors are recommended).

MAX1160

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Figure 1b. Single-Event Clock

Figure 1a. Timing Diagram

Table 1. Timing Parameters

PARAMETER	DESCRIPTION	MIN	TYP	MAX	UNITS
t _d	CLK to Data Valid Propagation Delay		14	18	ns
tpwH	CLK High Pulse Width	20		300	ns
tpwL	CLK Low Pulse Width	20			ns

The MAX1160 has two grounds: AGND and DGND. These internal grounds are isolated on the device. Use ground planes for optimum device performance. Use DGND for the DV_{CC} return path (typically 40mA) and for the return path for all digital output logic interfaces. Separate AGND and DGND from each other, connecting them together only through a ferrite bead at the device.

Connect a Schottky or hot carrier diode between AGND and V_{EE} . The use of separate power supplies between V_{CC} and DV_{CC} is not recommended due to potential power-supply-sequencing latchup conditions. For optimum performance, use the recommended circuit shown in Figure 2.

Voltage Reference

The MAX1160 requires the use of two voltage references: VFT and VFB. VFT is the force for the top of the voltage-reference ladder (typically +2.5V); VFB (typically -2.5V) is the force for the bottom of the voltage-reference ladder. Both voltages are applied across an 800Ω internal reference-ladder resistance. The +2.5V voltage source for reference VFT must be current limited to 20mA (max) if a different driving circuit is used in place of the recommended reference circuit shown in Figures 2 and 3. In addition, there are three reference-ladder taps (VST, VRM, and VSB). VST is the sense for the top of the reference ladder (+2V), VRM is the midpoint of the ladder (typically 0V), and VSB is the sense for the

bottom of the reference ladder (-2V). The voltages at VST and VSB are the device's true full-scale input voltages when VFT and VFB are driven to the recommended voltages (typically +2.5V and -2.5V, respectively). These points should be used to monitor the device's actual full-scale input range. When not being used, a decoupling capacitor of 0.01μ F (chip capacitor preferred) connected to AGND from each tap is recommended to minimize high-frequency noise injection.

Figure 2 shows an example of a recommended reference-driver circuit. IC1 is a MAX6225, a 2.5V reference with an accuracy of 0.2%. The 10k Ω potentiometer R1 supports a minimum adjustable range of 0.6%. Use an OP07 or equivalent device for IC2. R2 and R3 must be matched to within 0.1% with good TC tracking to maintain 0.3LSB matching between VFT and VFB. If 0.1% matching is not met, then R4 can be used to adjust the VFB voltage to the desired level. Adjust VFT and VFB such that VST and VSB are exactly +2V and -2V, respectively.

The analog input range scales proportionally with respect to the reference voltage if a different input range is required. The maximum scaling factor for device operation is $\pm 20\%$ of the recommended reference voltages of VFT and VFB. However, because the device is laser trimmed to optimize performance with $\pm 2.5V$ references, its accuracy degrades if operated beyond a $\pm 2\%$ range.

The following errors are defined:

+FS error = top of ladder offset voltage
=
$$\Delta$$
(+FS - VST + 1LSB)
-FS error = bottom of ladder offset voltage
= Δ (-FS - VSB - 1LSB)

where the +FS (full-scale) input voltage is defined as the output transition between 11 1111 1110 and 11 1111 1111, and the -FS input voltage is defined as the output transition between 00 0000 0000 and 00 0000 0001 (Table 2).

Analog Input

VIN is the analog input. The full-scale input range will be 80% of the reference voltage, or $\pm 2V$ with VFB = -2.5V and VFT = +2.5V.

The analog input's drive requirements are minimal when compared to conventional flash converters. This is due to the MAX1160's extremely low (5pF) input capacitance and very high ($300k\Omega$) input resistance. For example, for an input signal of $\pm 2Vp$ -p with a 10MHz input frequency, the peak output current required for the driving circuit is only 628µA.

Clock Input

The MAX1160 is driven from a single-ended TTL input (CLK). The CLK pulse width (t_{pwH}) must be kept between 20ns and 300ns to ensure proper operation of the internal track/hold amplifier (Figure 1a). When operating the MAX1160 at sampling rates above 3Msps, it is recommended that the clock input duty cycle be kept at



Figure 2. Typical Operating Circuit

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Table 2. Output Data Information

ANALOG INPUT	OVERRANGE D10	OUTPUT CODE D9–D0
> +2V + 1/2LSB	1	11 1111 1111
+2V - 1LSB	0	11 1111 111Ø
OV	0	00 0000 0000
-2V + 1LSB	0	00 0000 000Ø
< 2V	0	00 0000 0000

(Ø indicates the flickering bit between logic 0 and 1.)

50% to optimize performance, but performance will not be degraded if kept within the 40% to 60% range. The analog input signal is latched on the rising edge of CLK.

The clock input must be driven from fast TTL logic (VIH ≤ 4.5 V, t_{RISE} < 6ns). In the event the clock is driven from a high current source, use a 100 Ω resistor (R5) in series to limit current to approximately 45mA.

Digital Outputs

The format of the output data (D0–D9) is straight binary (Table 2). The outputs are latched on the rising edge of CLK with a typical propagation delay of 14ns. There is a one-clock-cycle latency between CLK and the valid output data (Figure 1a).

Overrange Output

The overrange output (D10) is an indication that the analog input signal has exceeded the positive full-scale input voltage by 1LSB. When this condition occurs, D10 will switch to logic 1. All other data outputs (D0–D9) will remain at logic 1 as long as D10 remains at logic 1. This feature makes it possible to include the MAX1160 in higher-resolution systems.

Evaluation Board

The MAX1160 EV kit is available to help designers demonstrate the MAX1160's full performance. This board includes a reference circuit, a clock-driver circuit, output data latches, and an on-board reconstruction of the digital data. A separate data sheet describing the operation of this board is also available. Contact the factory for price and availability.

Figure 3. Analog Equivalent Input Circuit

Figure 4. Digital Output Characteristics

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