



2MHz, High-Brightness LED Drivers with Integrated MOSFET and High-Side Current Sense

General Description

The MAX16832A/MAX16832C step-down constant-current high-brightness LED (HB LED) drivers provide a cost-effective design solution for automotive interior/exterior lighting, architectural and ambient lighting, LED bulbs, and other LED illumination applications.

The MAX16832A/MAX16832C operate from a +6.5V to +65V input voltage range. A high-side current-sense resistor adjusts the output current up to 700mA, and a dedicated pulse-width modulation (PWM) input enables pulsed LED dimming over a wide range of brightness levels.

These devices are well suited for applications requiring a wide input voltage range. The high-side current sensing and an integrated current-setting circuitry minimize the number of external components while delivering an average output current with $\pm 3\%$ accuracy. A hysteretic control method ensures excellent input supply rejection and fast response during load transients and PWM dimming. The MAX16832A allows 10% current ripple, and the MAX16832C allows 30% current ripple. Both devices operate up to a 2MHz switching frequency, thus allowing the use of small-sized components.

The MAX16832A/MAX16832C offer an analog dimming feature that reduces the output current by applying an external DC voltage below the internal 2V threshold voltage from TEMP_I to GND. TEMP_I also sources 25 μ A to a negative temperature coefficient (NTC) thermistor connected between TEMP_I and GND, thus providing an analog thermal-foldback feature that reduces the LED current when the temperature of the LED string exceeds a specified temperature point. Additional features include thermal-shutdown protection.

The MAX16832A/MAX16832C operate over the -40°C to $+125^{\circ}\text{C}$ automotive temperature range and are available in a thermally enhanced 8-pin SO package.

Applications

Architectural, Industrial, and Ambient Lighting
Automotive RCL, DRL, and Fog Lights
Heads-Up Displays
Indicator and Emergency Lighting
MR16 and MR111 LED Lights

Pin Configuration appears at end of data sheet.

Features

- ◆ High-Efficiency Solution
- ◆ 6.5V to 65V Input Voltage Range
- ◆ On-Board 65V, 0.45 Ω Power MOSFET
- ◆ Hysteretic Control: Up to 2MHz Switching Frequency
- ◆ $\pm 3\%$ LED Current Accuracy
- ◆ 200mV Current-Sense Reference
- ◆ Resistor-Programmable Constant LED Current
- ◆ Integrated High-Side Current Sense
- ◆ Thermal-Foldback Protection/Linear Dimming
- ◆ Thermal-Shutdown Protection
- ◆ Available in a Thermally Enhanced 8-Pin SO Package
- ◆ -40°C to $+125^{\circ}\text{C}$ Operating Temperature Range

Ordering Information

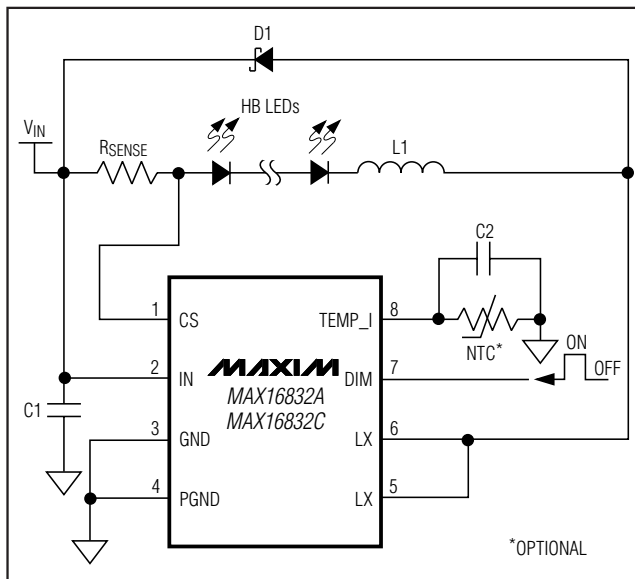
PART	TEMP RANGE	PIN-PACKAGE
MAX16832AASA+	-40°C to $+125^{\circ}\text{C}$	8 SO-EP**
MAX16832CASA+*	-40°C to $+125^{\circ}\text{C}$	8 SO-EP**

+Denotes a lead-free package.

*Future product—contact factory for availability.

**EP = Exposed pad.

Typical Application Circuit



MAX16832A/MAX16832C

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ABSOLUTE MAXIMUM RATINGS

IN, CS, LX, DIM to GND	-0.3V to +70V
TEMP_I to GND	-0.3V to +6V
PGND to GND	-0.3V to +0.3V
CS to IN	-0.3V to +0.3V
Maximum Current into Any Pin (except IN, LX, and PGND).....	20mA
Continuous Power Dissipation (T _A = +70°C) 8-Pin SO (derate 23.3mW/°C above +70°C).....	1860.5mW

Junction-to-Ambient Thermal Resistance (θ _{JA}) (Note 1).....	43°C/W
Operating Temperature Range	-40°C to +125°C
Junction Temperature	+150°C
Storage Temperature Range	-65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Pin-to-Pin ESD Ratings.....	±2.5kV

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{IN} = +24V, V_{DIM} = V_{IN}, T_A = T_J = -40°C to +125°C, unless otherwise noted. Typical values are at T_A = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Voltage Range	V _{IN}		6.5		65	V
Ground Current		No switching		1.5		mA
Supply Current		V _{DIM} < 0.6V, V _{IN} = 12V		350		μA
UNDERVOLTAGE LOCKOUT (UVLO)						
Undervoltage Lockout	UVLO	V _{CS} = V _{IN} - 100mV, V _{IN} rising until V _{LX} < 0.5V _{IN}		6.25	6.5	V
		V _{CS} = V _{IN} - 100mV, V _{IN} falling until V _{LX} > 0.5V _{IN}			6.0	
Undervoltage-Lockout Hysteresis				0.5		V
SENSE COMPARATOR						
Sense Voltage Threshold High	V _{SNSHI}	MAX16832A, V _{IN} - V _{CS} rising from 140mV until V _{LX} > 0.5V _{IN} , V _{DIM} = 5V	201	210	216	mV
		MAX16832C, V _{IN} - V _{CS} rising from 140mV until V _{LX} > 0.5V _{IN} , V _{DIM} = 5V	218	230	236	
Sense Voltage Threshold Low	V _{SNSLO}	MAX16832A, V _{IN} - V _{CS} falling from 260mV until V _{LX} < 0.5V _{IN} , V _{DIM} = 5V	185	190	198	mV
		MAX16832C, V _{IN} - V _{CS} falling from 260mV until V _{LX} < 0.5V _{IN} , V _{DIM} = 5V	166	170	180	
Propagation Delay to Output High	t _{DPDH}	Falling edge of V _{IN} - V _{CS} from 140mV to 260mV to V _{LX} = 0.5V _{IN}		50		ns
Propagation Delay to Output Low	t _{DPDL}	Rising edge of V _{CS} - V _{IN} from 260mV to 140mV to V _{LX} < 0.5V _{IN}		50		ns
CS Input Current	I _{CSIN}	V _{IN} - V _{CS} = 200mV, V _{IN} = V _{CS}			3.5	μA
INTERNAL MOSFET						
Drain-to-Source Resistance	R _{DS(on)}	V _{IN} = V _{DIM} = 24V, V _{CS} = 23.9V, I _{LX} = 700mA		0.45	0.9	Ω
		V _{IN} = V _{DIM} = 6.0V, V _{CS} = 5.9V, I _{LX} = 700mA		1	2	
LX Leakage Current	I _{LX_LEAK}	V _{DIM} = 0V, V _{LX} = 65V			10	μA

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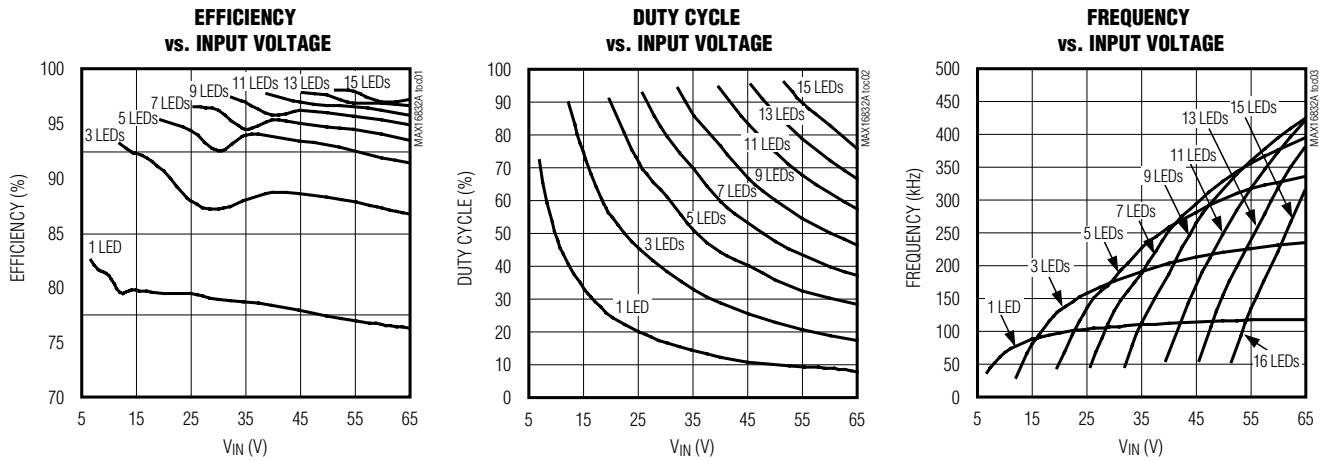
ELECTRICAL CHARACTERISTICS (continued)

($V_{IN} = +24V$, $V_{DIM} = V_{IN}$, $T_A = T_J = -40^{\circ}C$ to $+125^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DIM INPUT						
DIM Input-Voltage High	V_{IH}	$V_{IN} - V_{CS} = 100mV$	2.8			V
DIM Input-Voltage Low	V_{IL}	$V_{CS} - V_{IN} = 100mV$			0.6	V
DIM Turn-On Time	t_{DIM_ON}	V_{DIM} rising edge to $V_{LX} < 0.5V_{IN}$		200		ns
DIM Input Leakage High		$V_{DIM} = V_{IN}$		8	15	μA
DIM Input Leakage Low		$V_{DIM} = 0V$	-3	-1.5	0	μA
THERMAL SHUTDOWN						
Thermal-Shutdown Threshold		Temperature rising		+165		$^{\circ}C$
Thermal-Shutdown Threshold Hysteresis				10		$^{\circ}C$
THERMAL FOLDBACK						
Thermal-Foldback Enable Threshold Voltage	V_{TFB_ON}	$V_{DIM} = 5V$	1.9	2.0	2.12	V
Thermal-Foldback Slope	FB_{SLOPE}	$V_{DIM} = 5V$		0.75		1/V
TEMP_I Output Bias Current	I_{TEMP_I}		25	26.5	28	μA

Typical Operating Characteristics

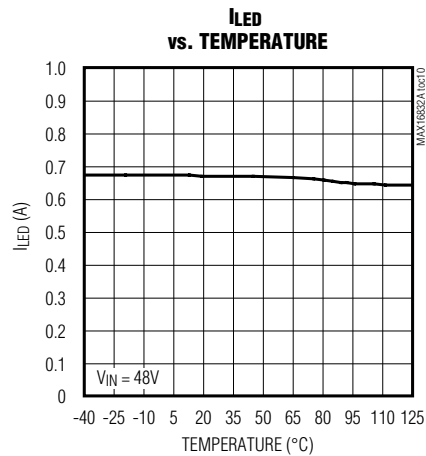
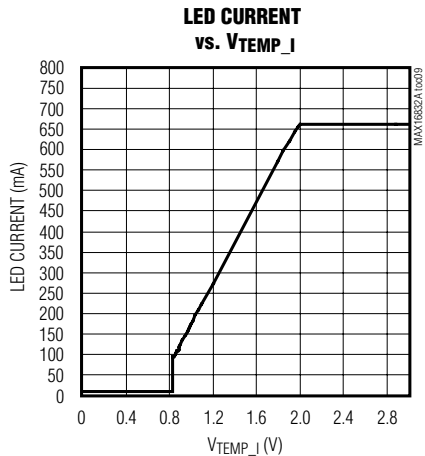
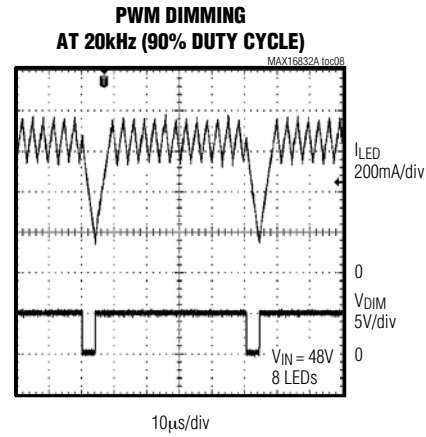
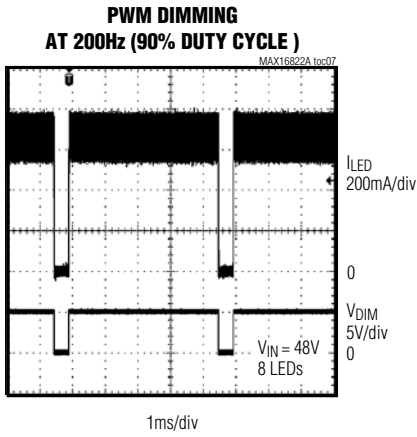
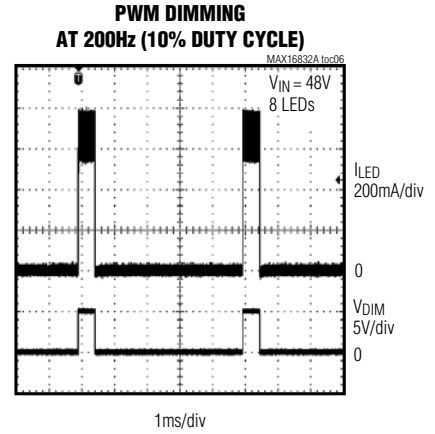
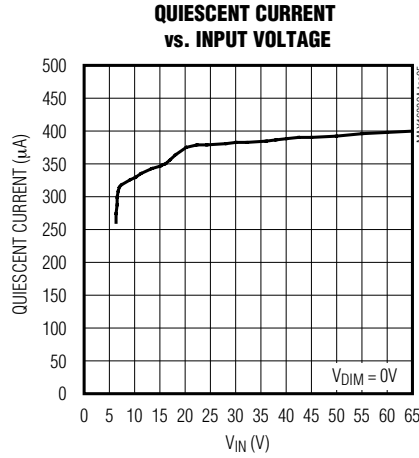
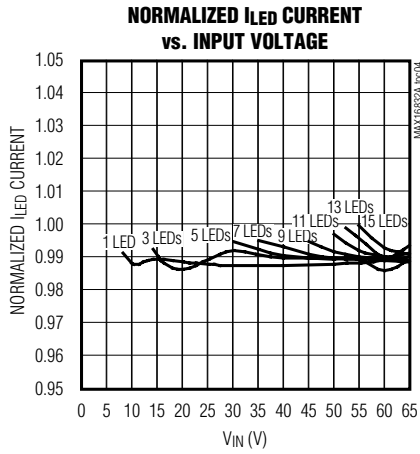
($V_{IN} = V_{DIM} = 48V$, $R_{SENSE} = 0.3\Omega$, $L = 220\mu H$ (connected between IN and CS). Typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.)



2MHz, High-Brightness LED Drivers with Integrated MOSFET and High-Side Current Sense

Typical Operating Characteristics (continued)

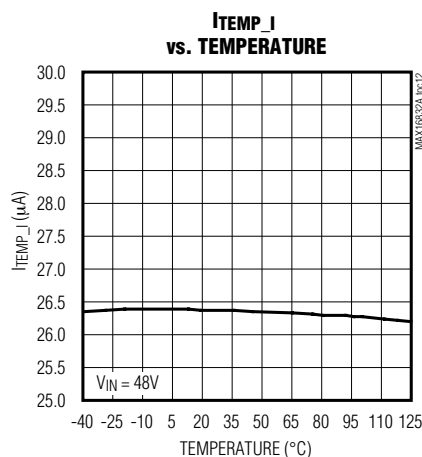
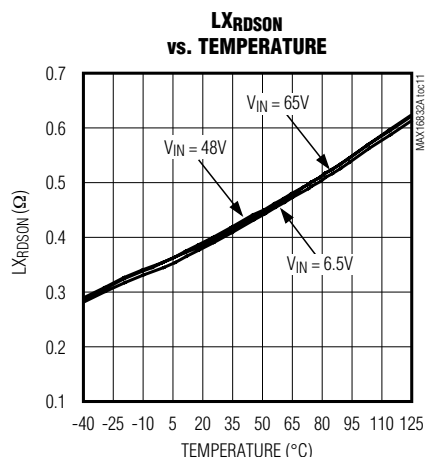
($V_{IN} = V_{DIM} = 48V$, $R_{SENSE} = 0.3\Omega$, $L = 220\mu H$ (connected between IN and CS). Typical values are at $T_A = +25^\circ C$, unless otherwise noted.)



2MHz, High-Brightness LED Drivers with Integrated MOSFET and High-Side Current Sense

Typical Operating Characteristics (continued)

($V_{IN} = V_{DIM} = 48V$, $R_{SENSE} = 0.3\Omega$, $L = 220\mu H$ (connected between IN and CS). Typical values are at $T_A = +25^\circ C$, unless otherwise noted.)



Pin Description

PIN	NAME	FUNCTION
1	CS	Current-Sense Input. Connect a resistor between IN and CS to program the LED current.
2	IN	Positive Supply Voltage Input. Bypass with a 1μF or higher value capacitor to GND.
3	GND	Ground
4	PGND	Power Ground
5, 6	LX	Switching Node
7	DIM	Logic-Level Dimming Input. Drive DIM low to turn off the current regulator. Drive DIM high to enable the current regulator.
8	TEMP_I	Thermal Foldback Control and Linear Dimming Input. Bypass with a 0.01μF capacitor to GND if thermal foldback or analog dimming is used. See the <i>Thermal Foldback</i> section.
—	EP	Exposed Pad. Connect EP to a large-area ground plane for effective power dissipation. Do not use as the IC ground connection.

Detailed Description

The MAX16832A/MAX16832C are step-down, constant-current, HB LED drivers. These devices operate from a +6.5V to +65V input voltage range and deliver up to 700mA of output current. A high-side current-sense resistor sets the output current and a dedicated PWM

dimming input enables pulsed LED dimming over a wide range of brightness levels.

A high-side current-sensing scheme and an on-board current-setting circuitry minimize the number of external components while delivering LED current with ±3% accuracy, using a 1% sense resistor. See Figure 1 for an internal block diagram.

2MHz, High-Brightness LED Drivers with Integrated MOSFET and High-Side Current Sense

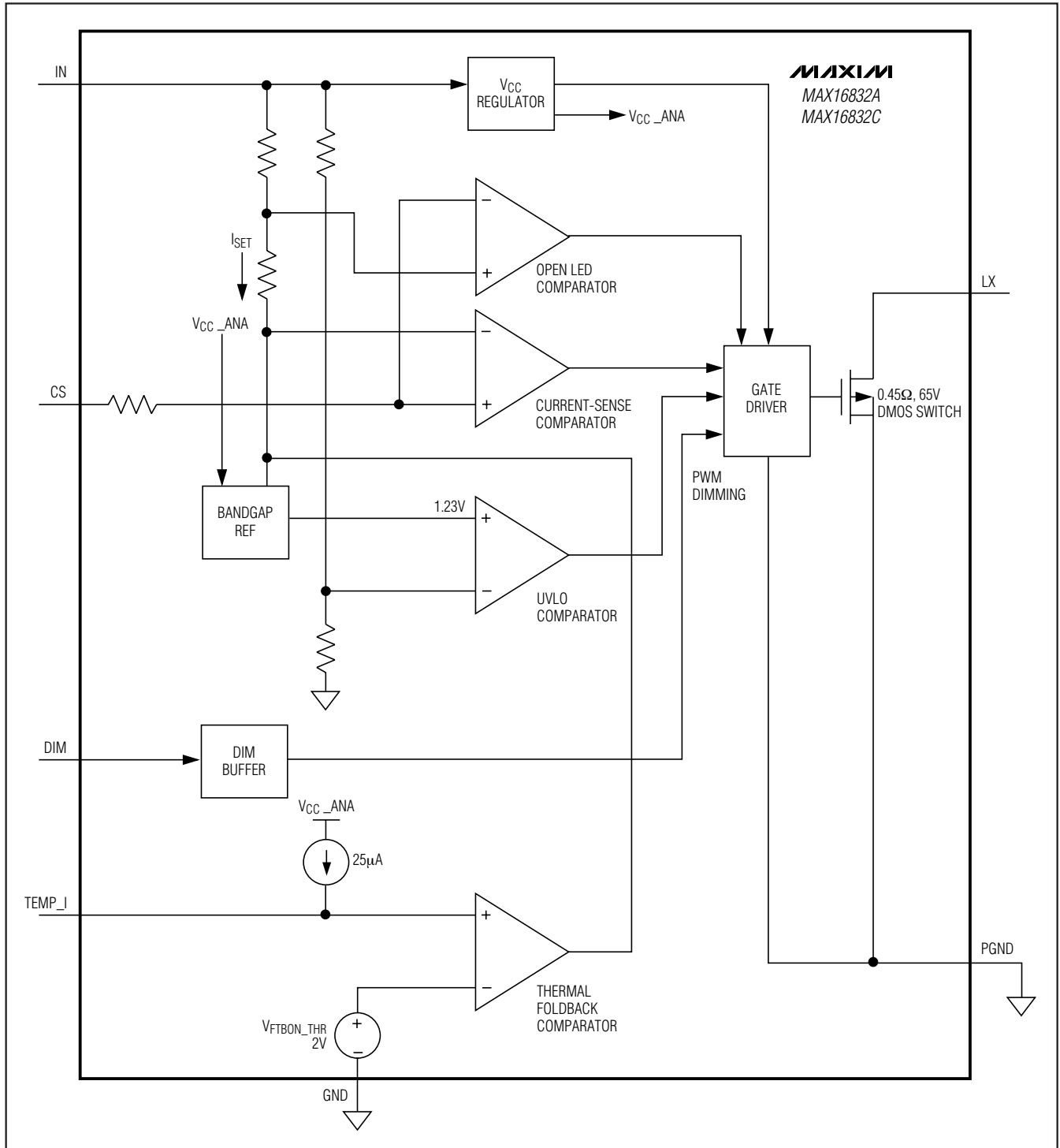


Figure 1. Internal Block Diagram

2MHz, High-Brightness LED Drivers with Integrated MOSFET and High-Side Current Sense

MAX16832A/MAX16832C

Undervoltage Lockout (UVLO)

The MAX16832A/MAX16832C include a UVLO with 500mV hysteresis. The internal MOSFET turns off when V_{IN} falls below 5.5V to 6.0V.

DIM Input

LED dimming is achieved by applying a PWM signal at DIM. A logic level below 0.6V at DIM forces the MAX16832A/MAX16832Cs' output low, thus turning off the LED current. To turn the LED current on, the logic level at DIM must be greater than 2.8V.

Thermal Shutdown

The MAX16832A/MAX16832C thermal-shutdown feature turns off the LX driver when the junction temperature exceeds +165°C. The LX driver turns back on when the junction temperature drops 10°C below the shutdown temperature threshold.

Analog Dimming

The MAX16832A/MAX16832C offer an analog-dimming feature that reduces the output current when the voltage at TEMP_I is below the internal 2V threshold voltage. The MAX16832A/MAX16832C achieve analog dimming by either an external DC voltage source connected between TEMP_I and ground or by a voltage on a resistor connected across TEMP_I and ground induced by an internal current source of 25µA. When the voltage at TEMP_I is below the internal 2V threshold limit, the MAX16832A/MAX16832C reduce the LED current. Use the following formula to set the analog dimming current:

$$I_{TF}(A) = I_{LED}(A) \times \left[1 - FB_{SLOPE} \left(\frac{1}{V} \right) \times (V_{TFB_ON} - V_{AD})(V) \right]$$

where $V_{TFB_ON} = 2V$ and $FB_{SLOPE} = 0.75$ are obtained from the *Electrical Characteristics* table and V_{AD} is the voltage at TEMP_I.

Thermal Foldback

The MAX16832A/MAX16832C include a thermal-foldback feature that reduces the output current when the temperature of the LED string exceeds a specified temperature point. These devices enter thermal-foldback mode when the voltage drop on the NTC thermistor, thermally attached to the LEDs and electrically connected between TEMP_I and ground, drops below the internal 2V threshold limit.

Applications Information

Selecting RSENSE to Set LED Current

The LED current is programmed with a current-sense resistor connected between IN and CS. Use the following equation to calculate the value of this resistor:

$$R_{SENSE}(\Omega) = \frac{1}{2} \frac{(V_{SNSHI} + V_{SNSLO})(V)}{I_{LED}(A)}$$

where V_{SNSHI} is the sense voltage threshold high and V_{SNSLO} is the sense voltage threshold low (see the *Electrical Characteristics* table for values).

Current-Regulator Operation

The MAX16832A/MAX16832C regulate the LED current using a comparator with hysteresis (see Figure 2). As the current through the inductor ramps up and the voltage across the sense resistor reaches the upper threshold, the internal MOSFET turns off. The internal MOSFET turns on again when the inductor current ramps down through the freewheeling diode until the voltage across the sense resistor equals the lower threshold. Use the following equation to determine the operating frequency:

$$f_{SW} = \frac{(V_{IN} - nV_{LED}) \times nV_{LED} \times R_{SENSE}}{V_{IN} \times \Delta V \times L}$$

where n is the number of LEDs, V_{LED} is the forward voltage drop of 1 LED, and $\Delta V = (V_{SNSHI} - V_{SNSLO})$.

Inductor Selection

The MAX16832A/MAX16832C operate up to a switching frequency of 2MHz. For space-sensitive applications, the high switching frequency allows the size of the inductor to be reduced. Use the following formula to calculate an approximate inductor value and use the closest standard value:

$$L(\text{approx.}) = \frac{(V_{IN} - nV_{LED}) \times nV_{LED} \times R_{SENSE}}{V_{IN} \times \Delta V \times f_{SW}}$$

For component selection, use the MAX16832A/C Design Tool available at: www.maxim-ic.com/MAX16832-software.

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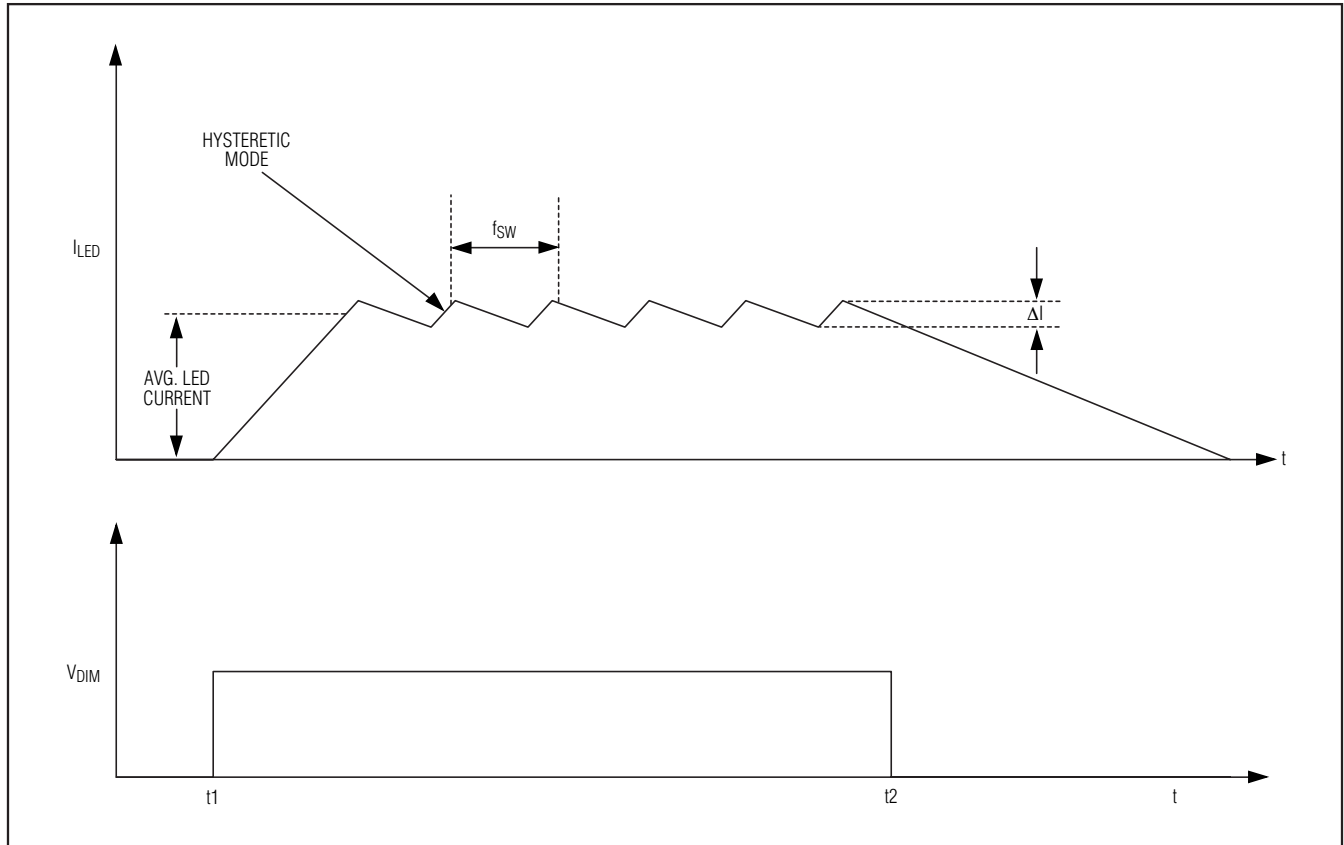


Figure 2. Current-Regulator Operation

Freewheeling-Diode Selection

For stability and best efficiency, a low forward-voltage drop diode with fast reverse-recovery time and low capacitance is recommended. A Schottky diode is a good choice as long as its breakdown voltage is high enough to withstand the maximum operating voltage.

PCB Layout Guidelines

Careful PCB layout is critical to achieve low switching losses and stable operation. Use a multilayer board whenever possible for better noise immunity. Minimize ground noise by connecting high-current ground returns, the input bypass-capacitor ground lead, and the output-filter ground lead to a single point (star ground configuration). In normal operation, there are two power loops. One is formed when the internal MOSFET

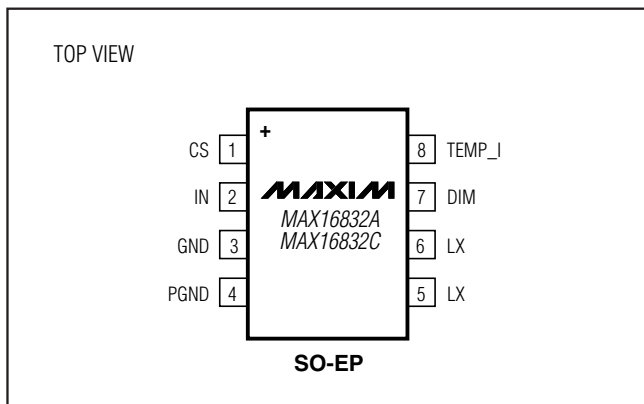
is on and the high current flows through IN, R_{SENSE}, LED load, the inductor, the internal MOSFET, and GND. The other loop is formed when the internal MOSFET is off and the high current circulates through R_{SENSE}, LED load, the inductor, and the freewheeling diode. Minimize each loop area to reduce noise interaction.

Place R_{SENSE} as close as possible to CS and IN. For better noise immunity, a Kelvin connection between CS and R_{SENSE} is strongly recommended.

Due to the integrated power MOSFET, the SO-EP package has an exposed pad to transfer the heat from the chip to the PCB. To make the thermal resistance between the chip and PCB lower, the exposed pad must be soldered to the PCB. The exposed pad is connected to GND.

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Pin Configuration



Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information, go to www.maxim-ic.com/packages.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
8 SO-EP	S8E-12	21-0111

MAX16832A/MAX16832C

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