ML65F16245*

16-Bit Bidirectional Transceiver with 3-State Outputs

GENERAL DESCRIPTION

The ML65F16245 is a BiCMOS, non-inverting 16-bit transceiver with 3-state outputs. This device was specifically designed for high speed bus applications. Its 16 channels support propagation delay of 2ns maximum, and fast output enable and disable times of 5ns or less to minimize datapath delay.

This device is designed to minimize undershoot, overshoot, and ground bounce to decrease noise delays. These transceivers implement a unique digital and analog implementation to eliminate the delays and noise inherent in traditional digital designs. The device offers a new method for quickly charging up a bus load capacitor to minimize bus settling times, or FastBus™ Charge. FastBus Charge is a transition current, (specified as $I_{DYNAMIC}$) that injects between 60 to 200mA (depending on output load) of current during the rise time and fall time. This current is used to reduce the amount of time it takes to charge up a heavily-capacitive loaded bus, effectively reducing the bus settling times, and improving data/clock margins in tight timing budgets.

Micro Linear's solution is intended for applications for critical bus timing designs that include minimizing device propagation delay, bus settling time, and time delays due to noise. Applications include; high speed memory arrays, bus or backplane isolation, bus to bus bridging, and sub-2ns propagation delay schemes.

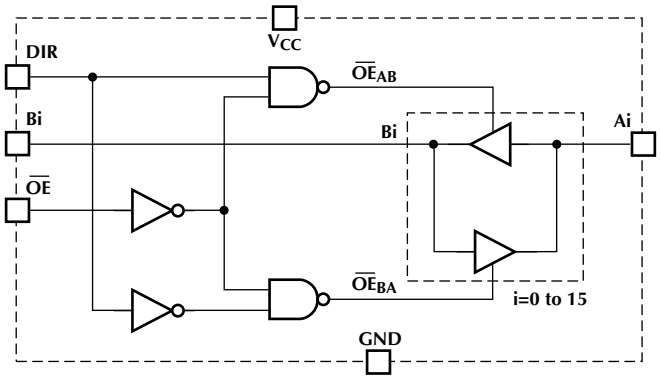
The ML65F16245 follows the pinout and functionality of the industry standard 2.7V to 3.6V-logic families.

FEATURES

- Low propagation delays — 2ns maximum for 3.3V, 2.5ns maximum for 2.7V
- Fast output enable/disable times of 5ns maximum
- FastBus Charge current to minimize the bus settling time during active capacitive loading
- 2.7V to 3.6V a V_{CC} supply operation; LV-TTL compatible input and output levels with 3-state capability
- Industry standard pinout compatible to FCT, ALV, LCX, LVT, and other low voltage logic families
- ESD protection exceeds 2000V
- Full output swing for increased noise margin
- Undershoot and overshoot protection to 400mV typically
- Low ground bounce design

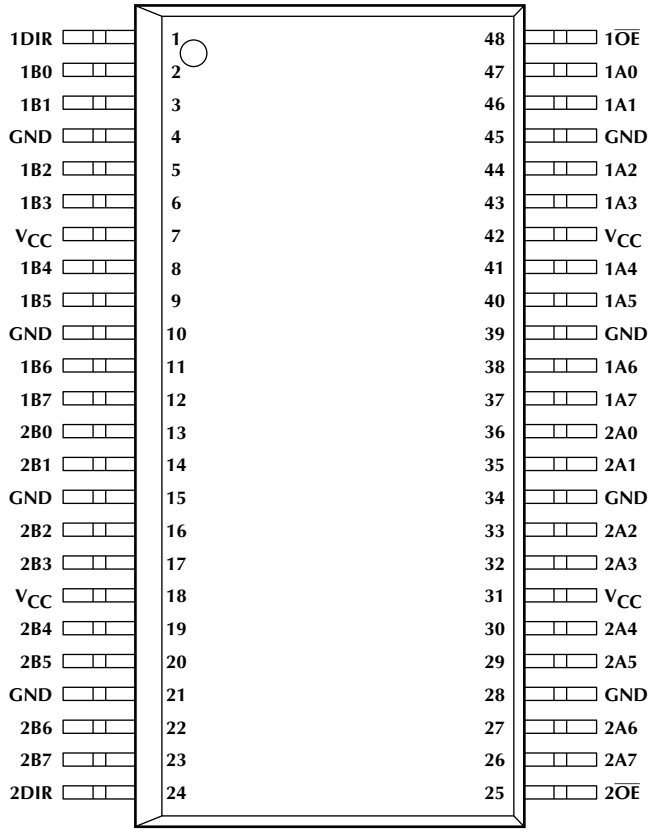
*This Part Is End Of Life As Of August 1, 2000

BLOCK DIAGRAM



PIN CONFIGURATION

ML65F16245
48-Pin SSOP (R48)
48-Pin TSSOP (T48)



TOP VIEW

FUNCTION TABLE

(Each 8-bit section)

$\overline{OE}$	DIR	Ai	Bi	FUNCTION
H	X	Z	Z	Disable
L	L	Output	Input	Bus B to Bus A
L	H	Input	Output	Bus A to Bus B

L = Logic Low, H = Logic High, X = Don't Care, Z = High Impedance

i = 0 to 7

PIN DESCRIPTION

PIN	NAME	FUNCTION
1	1DIR	Direction Select
2	1B0	Data Bus 1B
3	1B1	Data Bus 1B
4	GND	Signal Ground
5	1B2	Data Bus 1B
6	1B3	Data Bus 1B
7	V_{CC}	2.7V to 3.6V Supply
8	1B4	Data Bus 1B
9	1B5	Data Bus 1B
10	GND	Signal Ground
11	1B6	Data Bus 1B
12	1B7	Data Bus 1B
13	2B0	Data Bus 2B
14	2B1	Data Bus 2B
15	GND	Signal Ground
16	2B2	Data Bus 2B
17	2B3	Data Bus 2B
18	V_{CC}	2.7V to 3.6V Supply
19	2B4	Data Bus 2B
20	2B5	Data Bus 2B
21	GND	Signal Ground
22	2B6	Data Bus 2B
23	2B7	Data Bus 2B
24	2DIR	Direction Select
25	$2\overline{OE}$	Output Enable
26	2A7	Data Bus 2A
27	2A6	Data Bus 2A
28	GND	Signal Ground
29	2A5	Data Bus 2A
30	2A4	Data Bus 2A
31	V_{CC}	2.7V to 3.6V Supply
32	2A3	Data Bus 2A
33	2A2	Data Bus 2A
34	GND	Signal Ground
35	2A1	Data Bus 2A
36	2A0	Data Bus 2A
37	1A7	Data Bus 1A
38	1A6	Data Bus 1A
39	GND	Signal Ground
40	1A5	Data Bus 1A
41	1A4	Data Bus 1A
42	V_{CC}	2.7V to 3.6V Supply
43	1A3	Data Bus 1A
44	1A2	Data Bus 1A
45	GND	Signal Ground
46	1A1	Data Bus 1A
47	1A0	Data Bus 1A
48	$1\overline{OE}$	Output Enable

ABSOLUTE MAXIMUM RATINGS

Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

V_{CC} 7V
DC Input Voltage ..... –0.3V to V_{CC} + 0.3V
AC Input Voltage (PW < 20ns) ..... –3.0V
DC Output Voltage ..... –0.3V to 7VDC
Output Current, Source or Sink 180mA

Storage Temperature Range ..... –65°C to 150°C
Junction Temperature ..... 150°C
Lead Temperature (Soldering, 10sec) ..... 150°C
Thermal Impedance (θ_{JA}) ..... 76°C/W

OPERATING CONDITIONS

Temperature Range ..... 0°C to 70°C
V_{IN} Operating Range ..... 2.7V to 3.6V

ELECTRICAL CHARACTERISTICS

Unless otherwise specified, V_{IN} = 3.3V, T_A = Operating Temperature Range (Note 1).

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
AC ELECTRICAL CHARACTERISTICS (C_{LOAD} = 50pF)							
t_{PHL}, t_{PLH}	Propagation Delay	Ai to/from Bi	3.3V	1.35	1.7	2	ns
			2.7V	1.25	1.9	2.5	ns
t_{OE}	Output Enable Time	OE to Ai/Bi	3.3V			5	ns
			2.7V			6	ns
		DIR to Ai/Bi	3.3V			5	ns
			2.7V			6	ns
t_{OD}	Output Disable Time	OE to Ai/Bi	3.3V			5	ns
			2.7V			6	ns
		DIR to Ai/Bi	3.3V			5	ns
			2.7V			6	ns
T_{OS}	Output-to-Output Skew					300	ps
C_{IN}	Input Capacitance				8		pF
DC ELECTRICAL CHARACTERISTICS (C_{LOAD} = 50pF, R_{LOAD} = Open)							
V_{IH}	Input High Voltage	Logic high		2.0			V
V_{IL}	Input Low Voltage	Logic low				0.8	V
I_{IH}	Input High Current	Per pin, V_{IN} = 3V				300	µA
I_{IL}	Input Low Current	Per pin, V_{IN} = 0V				300	µA
I_{HI-Z}	Three-State Output Current	V_{CC} = 3.6V, 0 < V_{IN} < V_{CC}				5	µA
V_{IC}	Input Clamp Voltage	V_{CC} = 3.6V, I_{IN} = 18mA			–0.7	–0.2	V
$I_{DYNAMIC}$	Dynamic Transition Current (FastBus Charge)	Low to high transitions			80		mA
		High to low transitions			80		mA
V_{OH}	Output High Voltage		V_{CC} = 3.6V	2.4	3.4		V
			V_{CC} = 2.7V	2.25	2.35		V
V_{OL}	Output Low Voltage	V_{CC} = 3.6V and 2.7V				0.6	V
I_{CC}	Quiescent Power Supply Current	V_{CC} = 3.6V, f = 0Hz, inputs = V_{CC} or 0V				3	µA

Note 1: Limits are guaranteed by 100% testing, sampling, or correlation with worst-case test conditions.

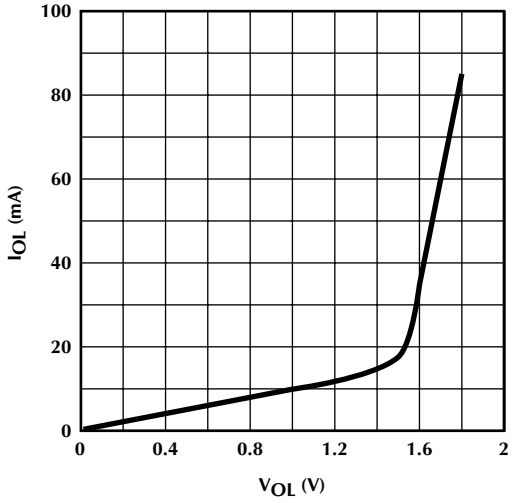


Figure 1a. Typical V_{OL} vs. I_{OL} for 3.3V V_{CC}. One Buffer Output

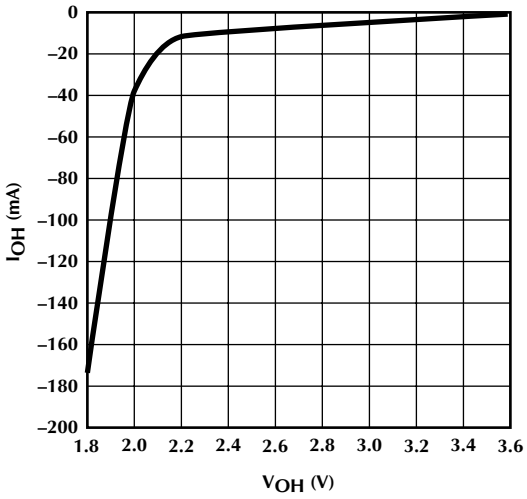


Figure 1b. Typical V_{OH} vs. I_{OH} for 3.3V V_{CC}. One Buffer Output

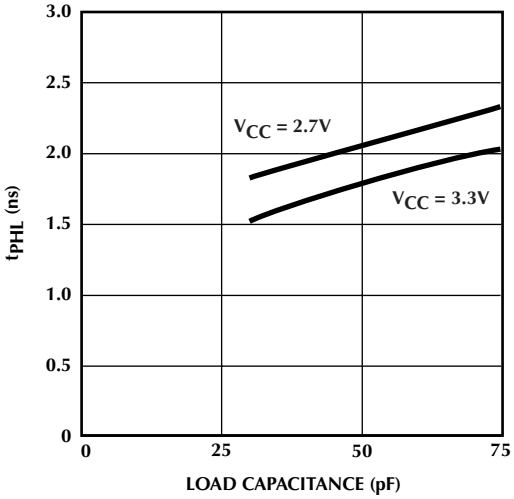


Figure 2a. Propagation Delay vs. Load Capacitance: 3.3V, 50MHZ

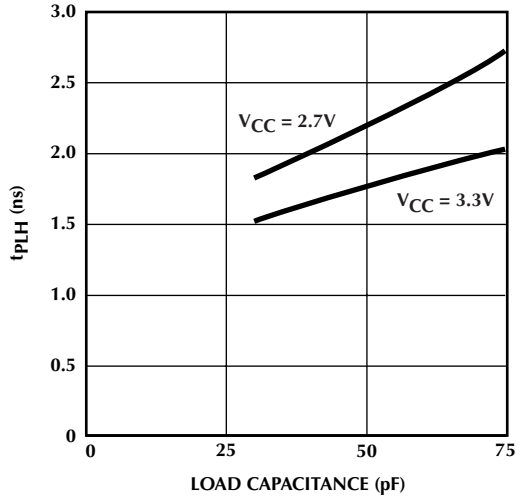


Figure 2b. Propagation Delay vs. Load Capacitance: 2.7V, 50MHZ

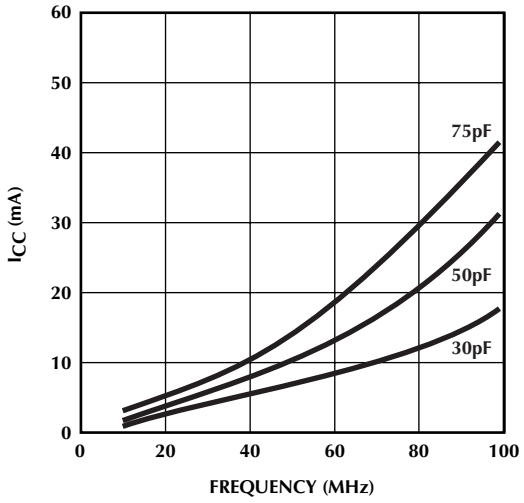


Figure 3a. I_{CC} vs. Frequency: V_{CC} = V_{IN} = 3.3V. One Buffer Output

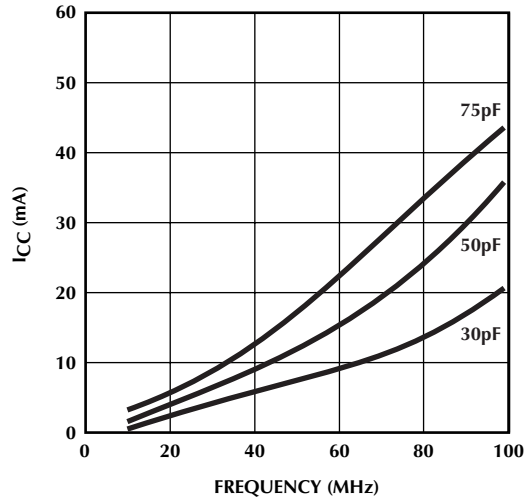


Figure 3b. I_{CC} vs. Frequency: V_{CC} = V_{IN} = 2.7V. One Buffer Output

FUNCTIONAL DESCRIPTION

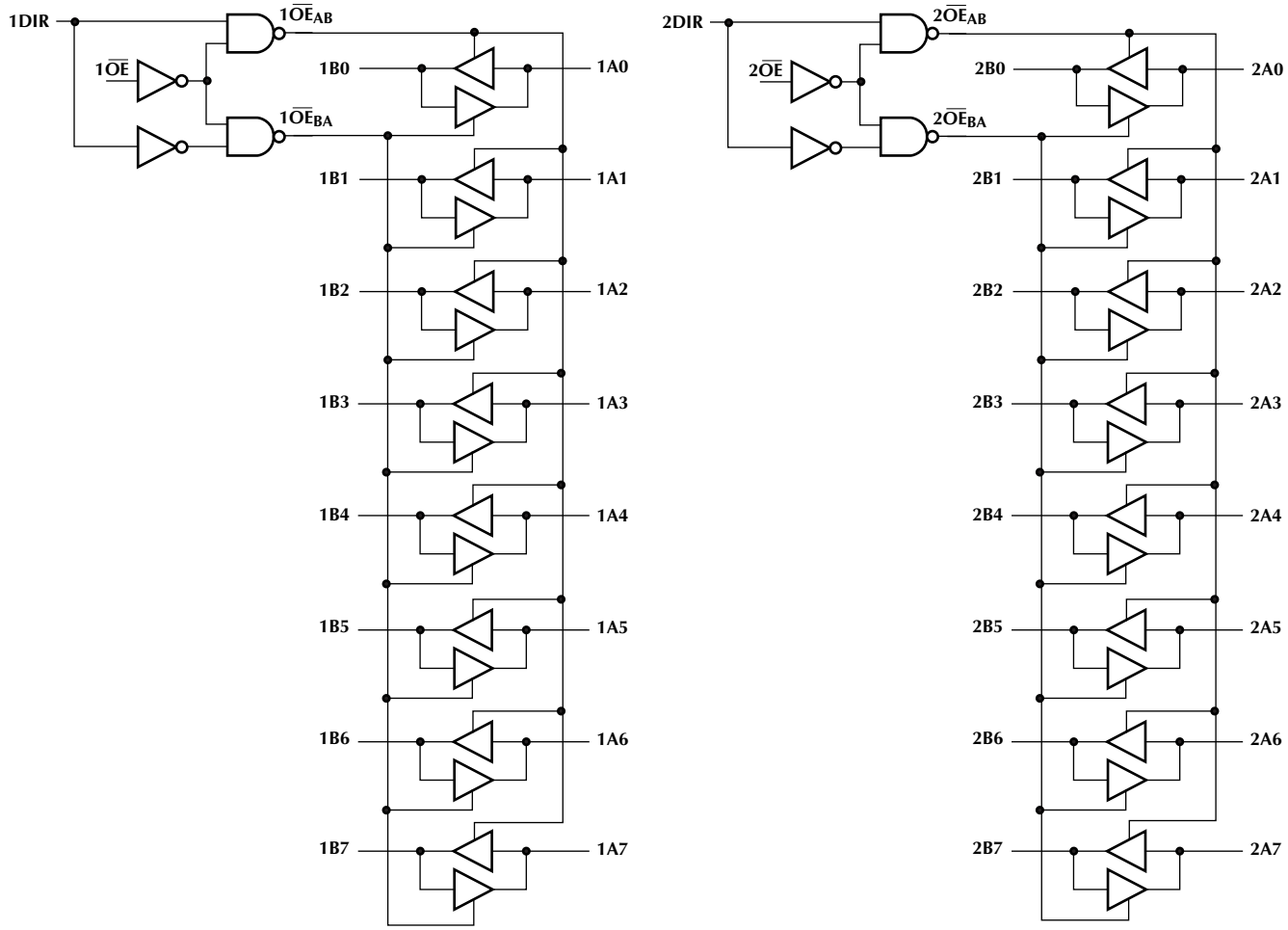


Figure 4. Logic Diagram

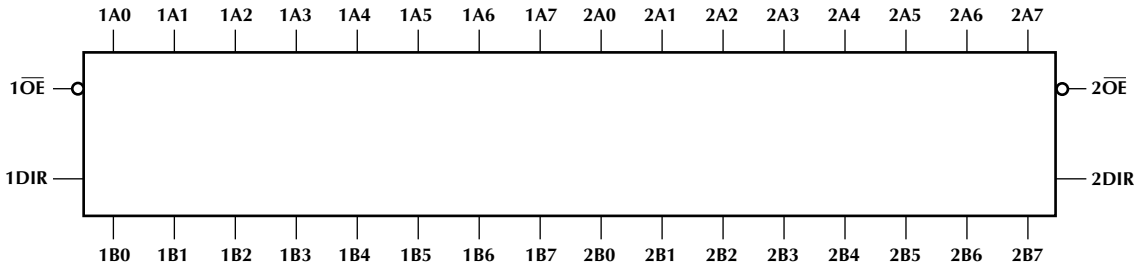


Figure 5. Logic Symbol

ARCHITECTURAL DESCRIPTION

The ML65F16245 is a 16-bit (dual-octal) non-inverting bus transceiver with 3-state outputs designed for 2.7V to 3.6V V_{CC} operation. This device is designed for asynchronous communication between data buses. The ML65F16245 can be used as two 8-bit transceivers or as one 16-bit transceiver and can be designated as Port-A bus and Port-B bus. The Direction and Output Enable controls are designed to operate these configurations. The direction control pin (iDIR) controls the direction of the data flow. The output enable pin ($1\overline{OE}$, $2\overline{OE}$) overrides the direction control and disables both ports.

Until now, these transceivers were typically implemented in CMOS logic and made to be TTL compatible by sizing the input devices appropriately. In order to buffer large capacitances with CMOS logic, it is necessary to cascade an even number of inverters, each successive inverter larger than the preceding, eventually leading to an inverter that will drive the required load capacitance at the required frequency. Each inverter stage represents an additional delay in the gating process because in order for a single gate to switch, the input must slew more than half of the supply voltage. The best of these 16-bit CMOS buffers has managed to drive 50pF load capacitance with a delay of 3ns.

Micro Linear has produced a 16-bit transceiver with a delay less than 2ns (at 3.3V) by using a unique circuit architecture that does not require cascade logic gates.

The basic architecture of the ML65F16245 is shown in Figure 6. In this circuit, there are two paths to the output. One path sources current to the load capacitance where the signal is asserted, and the other path sinks current from the output when the signal is negated.

The assertion path is the Darlington pair consisting of transistors Q1 and Q2. The effect of transistor Q1 is to increase the current gain through the stage from input to output, to increase the input resistance and to reduce input capacitance. During the transition state (the input from low-to-high) the output transistor Q2 sources large amount of current to quickly charge up a highly capacitive load which in effect reduces the bus settling time. This current is specified as $I_{DYNAMIC}$.

The negation path is also the Darlington pair consisting of transistor Q3 and transistor Q4. With M1 connecting to the input of the Darlington pair, Transistor Q4 then sinks a large amount of current during the input transition from high-to-low.

Inverter X2 is a helpful buffer that not only drives the output toward the upper rail but also pulls the output to the lower rail.

There are a number of MOSFETs not shown in Figure 6. These MOSFETs are used to 3-state the buffers.

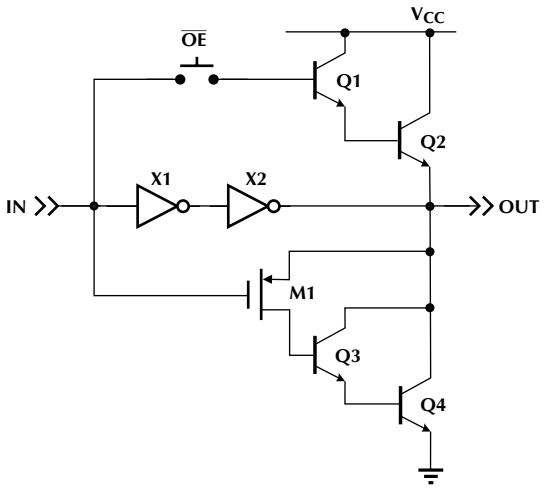


Figure 6. One Buffer Cell of the ML65F16245

CIRCUITS AND WAVE FORMS

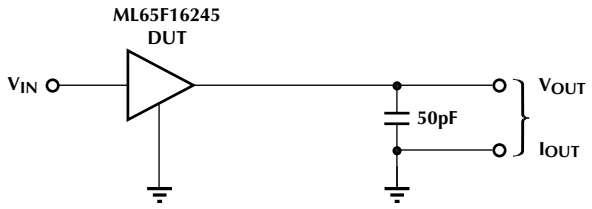


Figure 7. Test Circuits for All Outputs

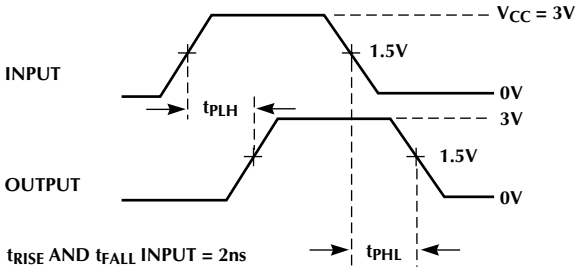


Figure 8. Propagation Delay

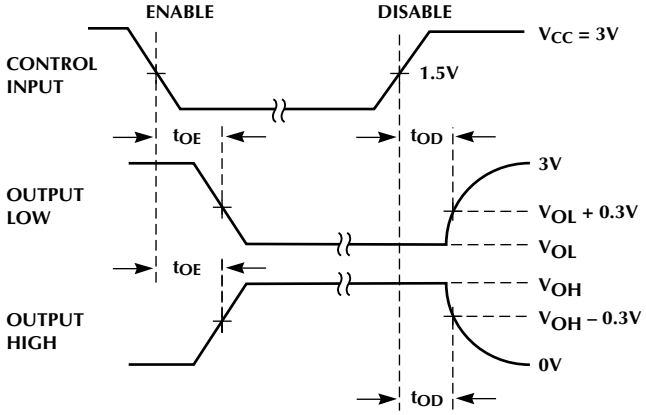


Figure 9. Enable and Disable Times

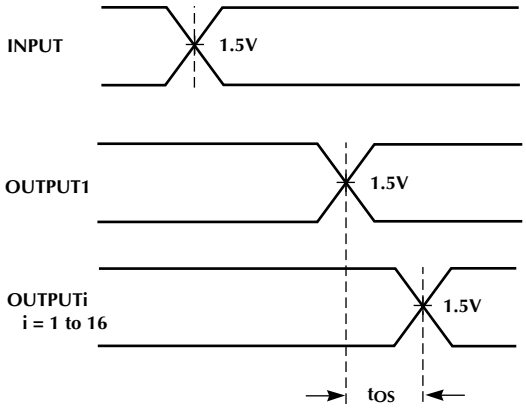
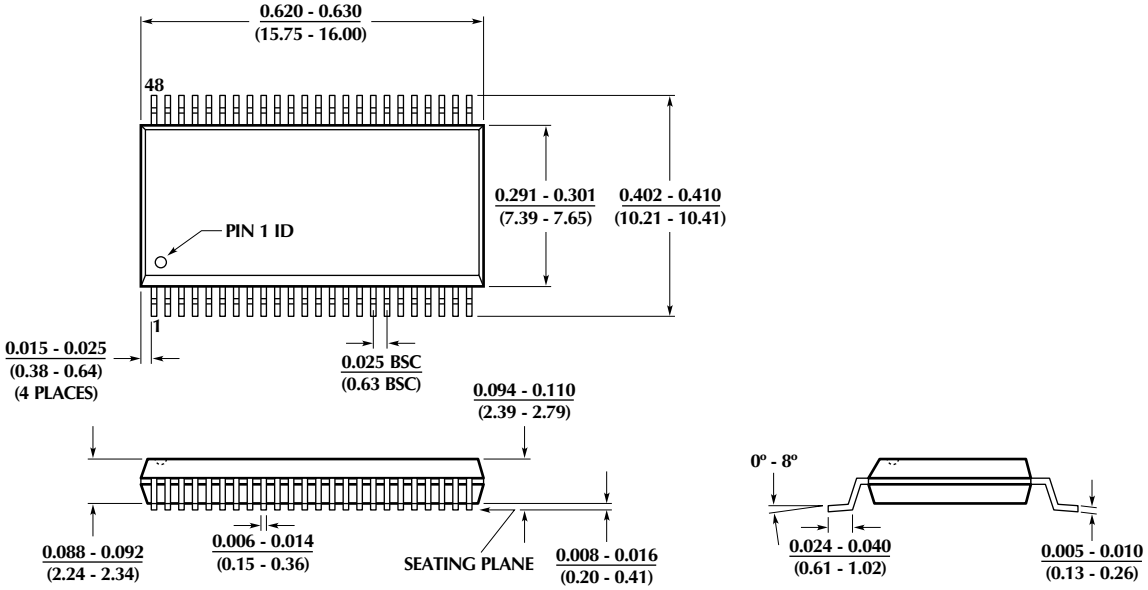


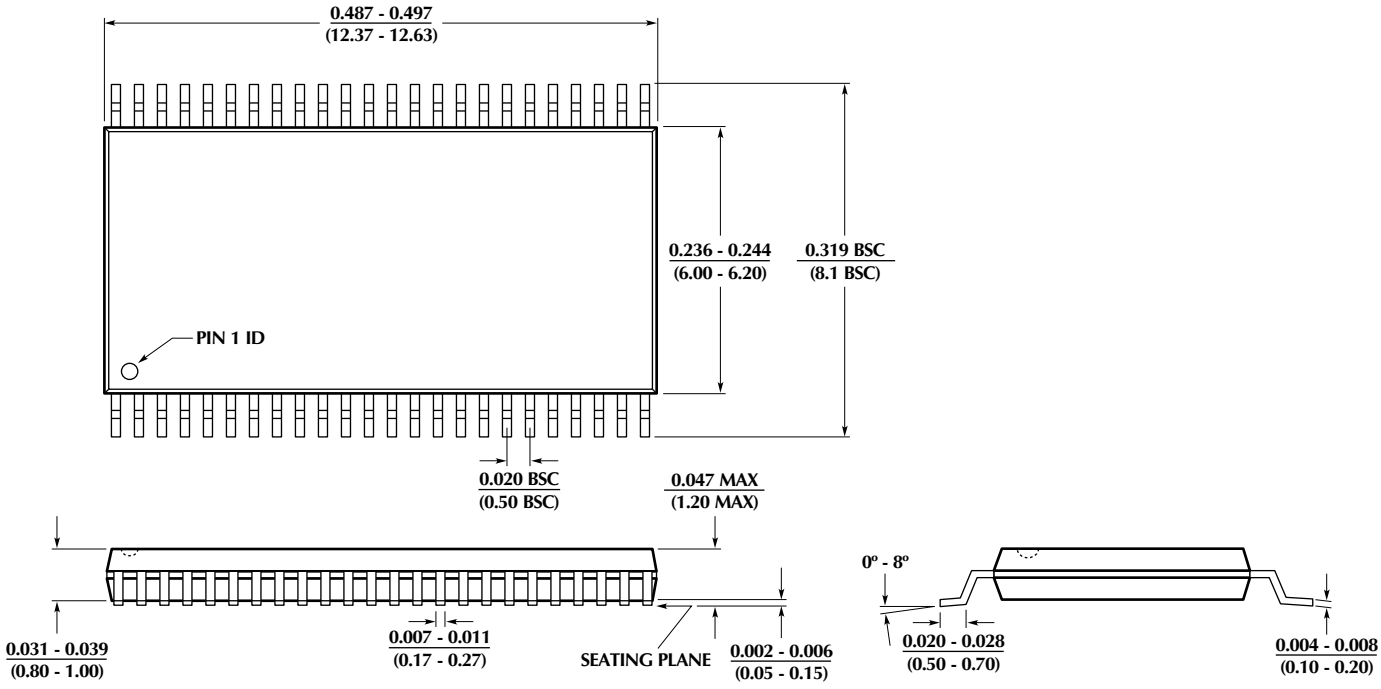
Figure 10. Output Skew

PHYSICAL DIMENSIONS inches (millimeters)

Package: R48
48-Pin SSOP



Package: T48
48-Pin TSSOP



ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
ML65F16245CR (Obsolete)	0°C to 70°C	48-Pin SSOP (R48)
ML65F16245CT (EOL)	0°C to 70°C	48-Pin TSSOP (T48)

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