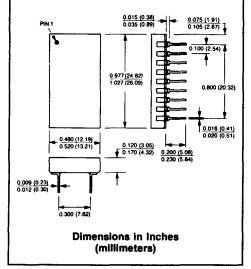


Digitally Controlled Programmable-Gain Amplifier

FEATURES

- Programmable Gain 1 to 128 in 8 Steps
- Gain Selected with a 3-Bit TTL Word
- Excellent Gain Accuracy: ± 0.002% @ G=1 ± 0.1% @ G=128
- Low Offset Voltage Drift ±5µV/°C
- High Input Impedance 1000 M Ω
- Small 18-Pin DIP
- Full Mil Operation -55°C to +125°C
- MIL-PRF-38534 Screening Optional.

18 PIN DIP



DESCRIPTION

MN2020 is a precision hybrid amplifier whose gain can be set to any one of 8 levels (1 to 128) with the application of a single 3-bit digital word (TTL logic levels) to its gain control inputs. This programmable-gain amplifier may be operated under direct computer or microprocessor control to provide fully automated, gain-range data acquisition.

The use of internal, laser-trimmed thin-film resistors result in excellent gain accuracy, linearity and drift characteristics. In addition, MN2020 has 100kHz of full power bandwidth and 5MHz of small signal bandwidth resulting in a rapid settling time of 5μ sec for a 20 volt step (@G=1).

The MN2020 Programmable-Gain Amplifier is packaged in a hermetically sealed, 18-pin dual-in-line package. The standard device is fully specified for either 0°C to +70°C or -55°C to +125°C ("H" model) operation. For military/aerospace or harshenvironment commercial/industrial applications, MN2020H/B CH is fully screened to MIL-PRF-38534.

MN2020 is an excellent choice for requirements where stable accurate gains are necessary. Typical applications include microprocessor-based data acquisition systems that have to handle a wide dynamic range of analog inputs.



MN2020 DIGITALLY CONTROLLED PROGRAMMABLE-GAIN AMPLIFIER

ABSOLUTE MAXIMUM RATINGS

Operating Temperature Specified Temperature

Storage Temperature +15V Supply (Pin 16) -15V Supply (Pin 6) +5V Supply (Pin 5) Analog Input (Pin 10)

Digital Inputs (Pins 2-4)

-55°C to +125°C 0°C to +70°C -55°C to +125°C ("H" Model)

-65°C to +150°C -0.5 to +18 Volts +0.5 to -18 Volts

-0.5 to +18 Volts ±15 Volts

0 to +Logic Supply (Note 1)

ORDERING INFORMATION

PART NUMBER MN2020/H/B CH Standard part is specified for 0°C to +70°C operation. Add "H" for specified -55°C to +125°C operation. Add "/B" to "H" models for Environmental Stress Screening. Add "CH" to "/B" models for 100% screening according to MIL-PRF-38534.

SPECIFICATIONS (TA=+25°C, Supply Voltages ±15V, unless otherwise specified).

| GAIN | MIN. | TYP. | MAX. | UNITS |
|---|------|----------------------------|---------------------------|-------------------------|
| Fixed Gain Settings | | 1, 2, 4, 8, 16, 32, 64, 1 | 28 | |
| Gain Nonlinearity (Note 2): G = 1 G = 128 | | ±0.002 ±0.04 | ±0.005 ±0.08 | % FSR (Note 3) % FSR |
| Gain Accuracy (Note 4) G = 1: +25°C 0°C to +70°C -55°C to +125°C | | ±0.002 ±0.003 ±0.004 | ±0.005 ±0.008 ±0.01 | % % % |
| G = 128: +25°C 0°C to +70°C -55°C to +125°C | | ±0.1 ±0.1 ±0.2 | ±0.2 ±0.2 ±0.4 | % % % |
| INPUT CHARACTERISTICS | ` | | | |
| Input Impedance Input Voltage Range (@G=1) Offset Voltage (RTI) (Notes 5 and 6) | | 1000 ±12 | | MΩ V |
| Initial 25°C Drift vs. Temperature -55°C to +125°C | | 100 5 | | μV μV/°C |
| Input Bias Current: +25° C 0° C to +70° C -55° C to +125° C ("H" Model) | | ± 20 ± 3 ±150 | ±200 ± 10 ±500 | pA nA nA |
| Voltage Noise (RTI) G=128 (0.1 to 10 Hz) | | 5 | | μ V p-p |
| OUTPUT CHARACTERISTICS | | | | |
| Output Voltage Swing Output Current | ±10 | ±12 | 5 | V mA |
| DYNAMIC CHARACTERISTICS | | | | |
| Small Signal Bandwidth G=1 G=128 | | 5 40 | | MHz KHz |
| Full Power Bandwidth (@G=1) Slew Rate | | 100 12 | | KHz V/μSec |
| Output Settling Time to ±0.1% 20V Step (Note 7) G=1 G=128 | | 5 65 | 75 | μSec μSec |
| GAIN SWITCHING | | | | |
| Gain Control Logic Inputs Logical 1 Logical 0 | +40 | | +0.8 | V V |
| Loading Gain Switching Time (Note 8) | | 0.6 | | μΑ μSec |
| POWER SUPPLY REQUIREMENTS (Vout=0) | | | | |
| Power Supply Range Current Drain (Analog Supply) Power Consumption | | ±15 ±9.2 275 | ±18 ±18 540 | V m A mW |

Note 1: Digital inputs should not exceed logic supply level. Logic supply (pin 5) must be at least +5V to maintain logic levels. Note 2: See definition of gain nonlinearity on Page 3.

Note 3: FSR Full Scale Range: If output swing = ±12V, FSR = 24V.

Note 4: Measured between endpoints of input (output) range in order to negate the effects of the offset voltage.

Note 5: RTI Referred to input.

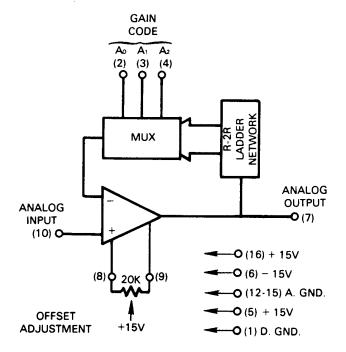
Note 6: Externally adjustable to zero

Note 7: For each gain value, the magnitude of the input step was chosen to make the output step 20V.

Note 8: Between any two gain values

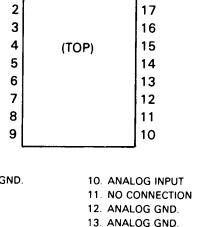
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BLOCK DIAGRAM



PINNING

1



- 1. DIGITAL GND. 2. Ao
- 3. A1
- 4. A2
- 5. +15 VOLTS (DIGITAL)
- 14. ANALOG GND.
- 6. -15 VOLTS
- 15. ANALOG GND. 16. +15 VOLTS (ANALOG)
- 7. ANALOG OUTPUT
- 17. NO CONNECTION
- 8. OFFSET ADJ. 9. OFFSET ADJ.
- 18. NO CONNECTION

PROGRAMMABLE GAIN AMPLIFIER SPECIFICATION DEFINITIONS

GAIN—The ratio of the amplitude of output signal voltage to the amplitude of input signal voltage.

GAIN ACCURACY—Either the percentage that actual gain differs from ideal gain (%) or the amount that the output, at a certain gain and input level, differs from the ideal value (volts, % FSR).

GAIN NONLINEARITY—Maximum deviation of the inputoutput voltage transfer function from the ideal, expressed as a percentage of the full output voltage range (FSR).

GAIN SWITCHING TIME—The time necessary for the amplifier gain to settle to within 0.1% of its new value following the appearance of a new digital code at its gain coding terminals.

INITIAL OFFSET VOLTAGE (Referred to Input)-The collection of internal voltage offsets summed and treated as a single offset voltage source appearing in series with the input. This offset, multiplied by the programmed gain, will appear at the amplifier output, even when the input signal is zero. This offset voltage can normally be zeroed out with an external trimpot.

INITIAL OFFSET VOLTAGE (RTI) DRIFT vs TEMPERA-TURE-Drift in initial offset voltage resulting from temperature variations. Usually expressed as V/°C or ppm of FSR /°C.

INPUT BIAS CURRENT-The current drawn into (or out of) the input terminals of the amplifier when the amplifier is turned on and the input signal is zero (input grounded).

INPUT IMPEDANCE—Total impedance seen looking into the amplifier input terminal (with the load connected) with respect to analog ground.

LOADING—The apparent load that the digital gain coding inputs of the amplifier present to their driving circuits. Usually expressed as standard logic loads (e.g. 3 TTL Loads) or in terms of the current sourced or sank when the input is a logic "0" or "1".

FPUT DRIVE CURRENT—Current that the amplifier will source or sink to the load while remaining within specification.

OUTPUT VOLTAGE SWING—Maximum allowable output excursion for faithful reproduction of the input signal. This is limited to several volts less than the associated power supply voltage range.

SETTLING TIME—The interval from the application of either an input step at a fixed gain or a new gain code at a fixed input level to the output's settling within a specified error band (usually 0.1%) of its final value.

SMALL SIGNAL BANDWIDTH—Frequency at which the amplifiers gain drops 3 dB from its D.C. value.

SLEW RATE-Maximum rate of change (V/Sec) in the output in response to a step change at the input or a gain change.

VOLTAGE NOISE (RTI) — Sum of the internal noise sources treated as a single source appearing in series with the input signal. The noise, multipled by the programmed gain, will appear at the amplifier output. Voltage noise is dependent upon bandwidth and may be reduced by using the minimum bandwidth necessary for a given application.



TYPICAL CHARACTERISTICS

(TA=25°C, Supplies ±15V)

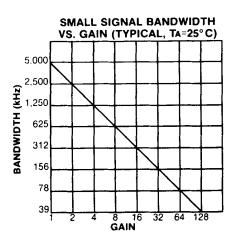
GAIN CODES AND SETTLING TIMES

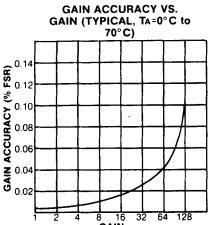
| GAIN | DIGITAL CODE A2 A1 A0 | | | OUTPUT SETTLING TIME* (±0.1% 20V Step) | | |
|-----------|--------------------------|---|--------|---|--|--|
| 1 2 | 0 | 0 | 0 | 2.5 μSec 3 μSec | | |
| 4 | 0 | 1 | 0 | 4 μSec | | |
| 8 16 | 0 | 1 | 1 0 | 6 μSec 8 μSec | | |
| 32 | 1 | 0 | 1 | 17 μSec | | |
| 64 128 | 1 | 1 | 0 1 | 33 μSec 65 μSec | | |

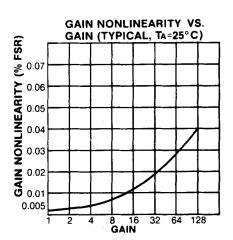
GAIN ACCURACIES

| GAIN | ACCURACY (%) | | | | | | | | |
|------|--------------|-------|-------------|-------|-----------------|-------|--|--|--|
| | 25° C | | 0°C to 70°C | | -55°C to +125°C | | | | |
| | TYPICAL | MAX. | TYPICAL | MAX. | TYPICAL | MAX. | | | |
| 1 2 | 0.002 | 0.005 | 0.003 | 0.008 | 0.004 | 0.010 | | | |
| | 0.005 | 0.015 | 0.005 | 0.020 | 0.008 | 0.020 | | | |
| 4 | 0.005 | 0.015 | 0.005 | 0.020 | 0.015 | 0.040 | | | |
| 8 | 0.010 | 0.020 | 0.015 | 0.040 | 0.020 | 0.080 | | | |
| 16 | 0.020 | 0.030 | 0.020 | 0.040 | 0.025 | 0.080 | | | |
| 32 | 0.020 | 0.040 | 0.020 | 0.040 | 0.040 | 0.100 | | | |
| 64 | 0.040 | 0.100 | 0.040 | 0.100 | 0.100 | 0.300 | | | |
| 128 | 0.100 | 0.200 | 0.100 | 0.200 | 0.200 | 0.400 | | | |

^{*}For each gain value the magnitude of the input step was chosen to make the output step 20V.







APPLICATIONS INFORMATION:

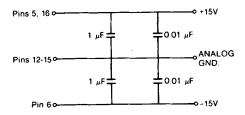
OFFSET ADJUSTMENT:

The MN2020 meets all specifications without adjustment. However, the initial offset voltage may be adjusted to zero with the addition of a trimpot between pins 8 and 9 as shown in the block diagram. A 20K, 10 turn, <100 ppm/°C TC trimpot should be used to minimize drift with temperature.

LAYOUT CONSIDERATIONS:

Proper attention to layout and decoupling is necessary to obtain specified accuracies. Analog and digital grounds are not connected internally. The four (4) analog commons (pins 12-15) and the digital common (pin 1) should be tied together as close to the package as possible, preferably to a large ground plane underneath the package. If these commons must be run separately, wide conductor runs should be used.

Power supplies should be decoupled with tantalum or electrolytic capacitors located close to the device package. For optimum results, $1\mu F$ capacitors paralleled by $0.01~\mu F$ ceramic capacitors should be connected as shown in the adjacent diagram.





INTERFACING THE MN2020 TO POPULAR MICROPROCESSORS

The MN2020 can be easily interfaced to microprocessors for fully automated data acquisition or other applications where it is desirable to change gain under program control. Memory mapped I/O is recommended to take advantage of the powerful memory reference instructions available

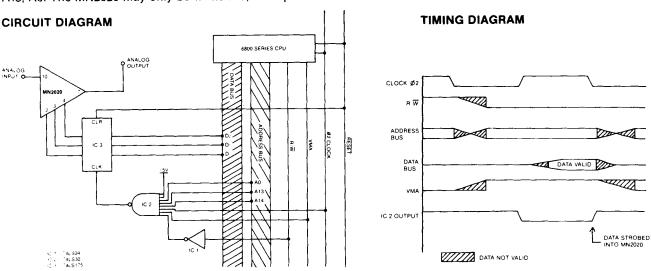
in most microprocessor instruction sets. Detailed information is provided below for the 6800 Series and 8080 Series processors. Interfacing to other processors would be similar.

INTERFACING TO 6800 SERIES MICROPROCESSORS

Wiring and Timing Diagrams for interfacing the MN2020 with the 6800 family of microprocessors are shown in Figure 1. In this example, the MN2020's gain control inputs are addressed and written to as a memory location with the gain code in the three lowest order bits (Do, D1, D2) of the accumulator. The address bus connections shown (A14, A13, and A0) correspond to memory address 6001 hexadecimal. Redundant addressing of the MN2020 would occur at any address containing the bit combination A14, A13, A0. The MN2020 may only be written to; attempts to

read the memory address assigned to the MN2020 will result in indeterminate (floating CMOS inputs) data.

The connections to the VMA, \emptyset 2 clock and R/ \overline{W} lines are used to insure correct timing and to prevent spurious data that may be present on the address bus during non-memory transfer operations from addressing the MN2020. When connected as shown, the MN2020 will be reset to a gain of 1 by a reset pulse from the hardware reset control line of the 6800 System.



INTERFACING TO 8080 SERIES MICROPROCESSORS

Wiring and Timing Diagrams for interfacing the MN2020 with the 8080 family of microprocessors are shown in Figure 2. In this example, an 8228 system controller is used, and the MN2020 is treated as a standard I/O device. The gain code is written, as an output instruction, to the MN2020 which is located at I/O address 80 hexadecimal.

When the values of A7, A0, and $\overline{I/O}$ W are "1", "0" and "0" respectively, the output of \underline{IC} 1A will go to logic zero. When either of the address or $\overline{I/O}$ W outputs leaves the above state, the output of IC2 is forced to a logic one, with the rising edge triggering the latch IC2 which strobes the gain code information into the MN2020.

