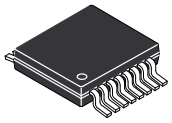
The MRFIC Line

1.9 GHz GaAs Upconverter

MRFIC1813

1.9 GHz UPMIXER AND EXCITER AMPLIFIER

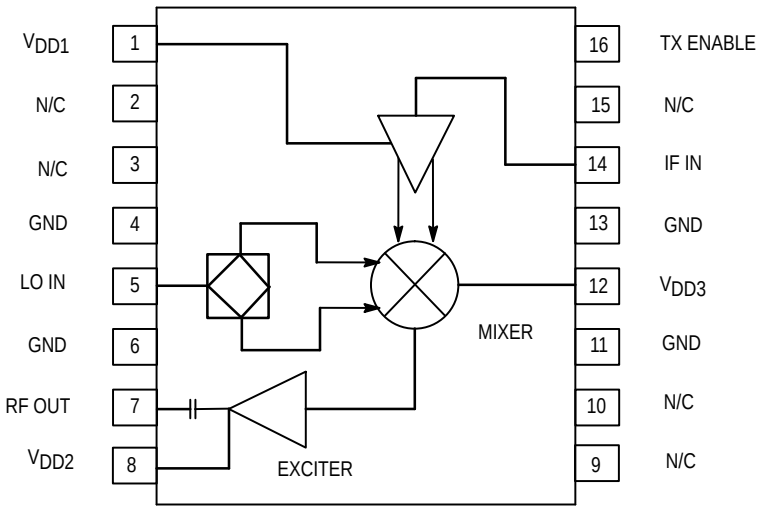


CASE 948C–03
(TSSOP–16)

Designed primarily for use in wireless Personal Communication Systems (PCS) applications such as Digital European Cordless Telephone (DECT), Japan's Personal Handy System (PHS) and the emerging North American systems. The MRFIC1813 is also applicable to 2.4 GHz ISM equipment. The device combines a balanced upmixer and a transmit exciter amplifier in a low-cost TSSOP–16 package. Minimal off-chip matching is required while allowing for maximum flexibility and efficiency. The mixer is optimized for low–side injection and provides more than 12 dB of conversion gain with over 0 dBm output at 1 dB gain compression. Image filtering is implemented off-chip to allow maximum flexibility. A CMOS compatible ENABLE pin allows standby operation where the current drain is less than 250 μA.

Together with other devices from the MRFIC180X or the MRFIC240X series, this GaAs IC family offers the complete transmit and receive functions, less LO and filters, needed for a typical 1.8 GHz cordless telephone or 2.4 GHz ISM band equipment.

- Usable Frequency Range = 1.7 to 2.5 GHz
- 15 dB Typ IF to RF Conversion Gain
- 3 dBm Power Output Typ, 0 dBm Minimum at 1 dB Gain Compression
- Simple Off-chip Matching for Maximum Flexibility
- Low Power Consumption = 75 mW (Typ)
- Single Bias Supply = 2.7 to 4.5 Volts
- Low LO Power Requirement = – 5 dBm (Typ)
- Low Cost Surface Mount Plastic Package
- Order MRFIC1813R2 for Tape and Reel.
  R2 Suffix = 2,500 Units per 16 mm, 13 inch Reel.
- Device Marking = M1813



Pin Connections and Functional Block Diagram

MOTOROLA

MAXIMUM RATINGS (T_A = 25°C unless otherwise noted)

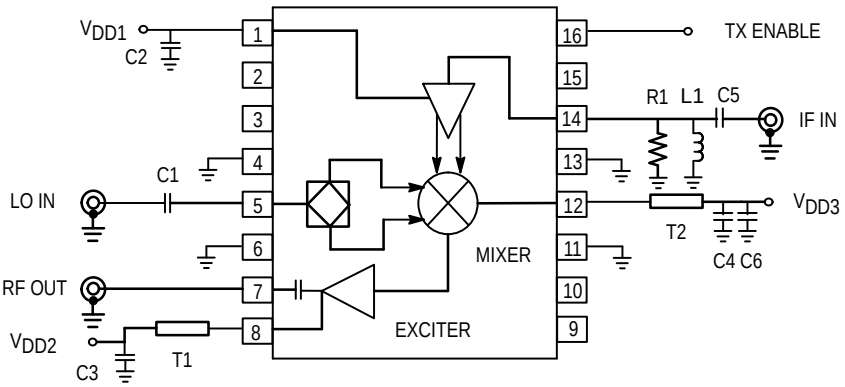
Ratings	Symbol	Limit	Unit
Supply Voltage	V_{DD1}, V_{DD2}, V_{DD3}	5.5	Vdc
IF Input Power	P_{IF}	3	dBm
LO Input Power	P_{LO}	3	dBm
Enable Voltage	TX ENABLE	5.5	Vdc
Storage Temperature Range	T_{stg}	– 65 to +150	°C
Operating Ambient Temperature	T_A	– 30 to +85	°C

RECOMMENDED OPERATING RANGES

Parameter	Symbol	Value	Unit
RF Output Frequency	f_{RF}	1.7 to 2.5	GHz
LO Input Frequency	f_{LO}	1.5 to 2.4	GHz
IF Input Frequency	f_{IF}	70 to 350	MHz
Supply Voltage	V_{DD}	2.7 to 4.5	Vdc
TX Enable Voltage, ON	TX ENABLE	2.7 to V_{DD}	Vdc
TX Enable Voltage, OFF	TX ENABLE	0 to 0.2	Vdc

ELECTRICAL CHARACTERISTICS ($V_{DD1,2,3}$, TX ENABLE= 3 V, T_A = 25°C, f_{LO} = 1.65 GHz @ – 5 dBm, f_{IF} = 250 MHz @ –15 dBm)

Characteristic	Min	Typ	Max	Unit
IF to RF Small Signal Conversion Gain (P_{RF} = –35 dBm)	12	15	—	dB
RF Output 1 dB Gain Compression	0	3	—	dBm
RF Output 3rd Order Intercept	—	11	—	dBm
LO Feedthrough to RF Port	—	–15	–10	dBm
Noise Figure	—	11	—	dB
Lower Sideband Output Power at RF Port	—	–10	–6	dBm
Supply Current TX Mode	—	25	35	mA
Supply Current Standby Mode (TX ENABLE = 0 V, LO Off)	—	100	250	μA
TX Enable Current	—	3	—	μA



C1, C2, C3, C4 15 pF
C6 1 μF
R1 300 Ω
T1, T2 MICROSTRIP, Z_0 = 73 Ω, θ = 29° @ 1.9 GHz
BOARD MATERIAL – GLASS/EPOXY, ε_r = 4.45,
DIELECTRIC THICKNESS = 0.018 INCH

IF Frequency	L1	C5
110 MHz	180 nH	12 pF
250 MHz	68 nH	5.6 pF
350 MHz	39 nH	3.9 pF

Figure 1. Applications Circuit Configuration

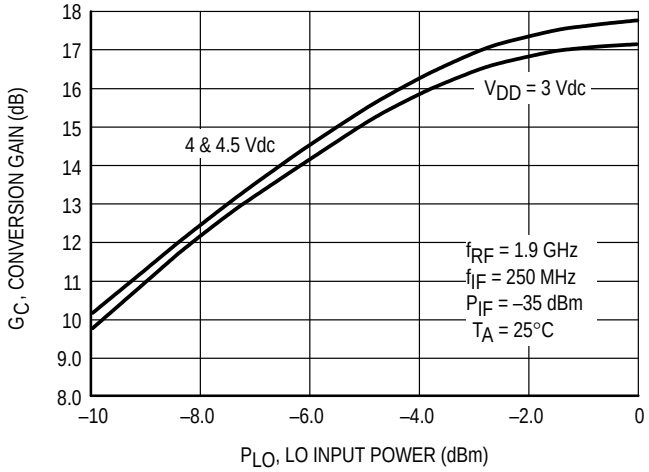


Figure 2. Conversion Gain versus LO Power

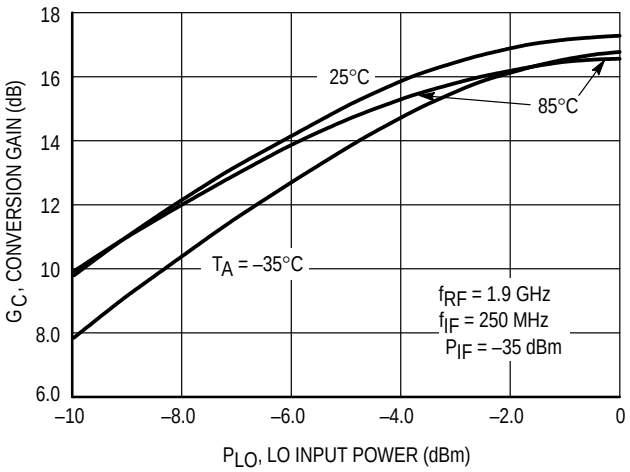


Figure 3. Conversion Gain versus LO Power

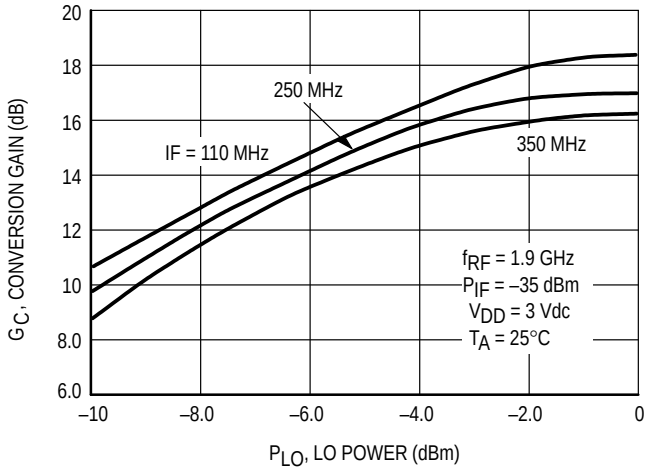


Figure 4. Conversion Gain versus LO Power

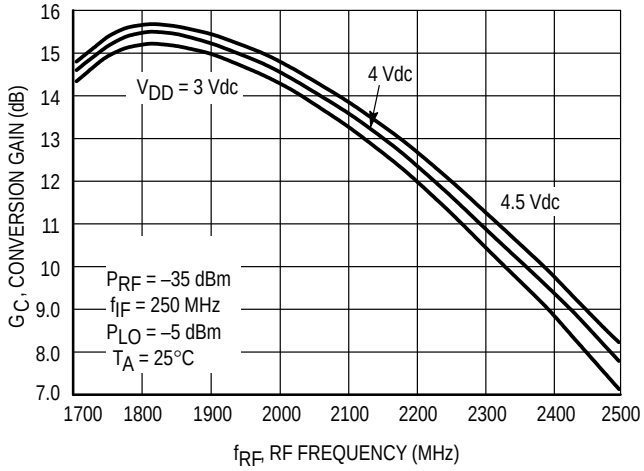


Figure 5. Conversion Gain versus RF Frequency

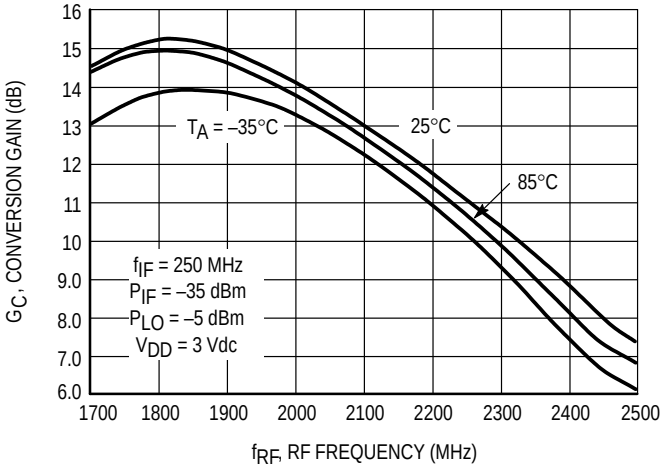


Figure 6. Conversion Gain versus RF Frequency

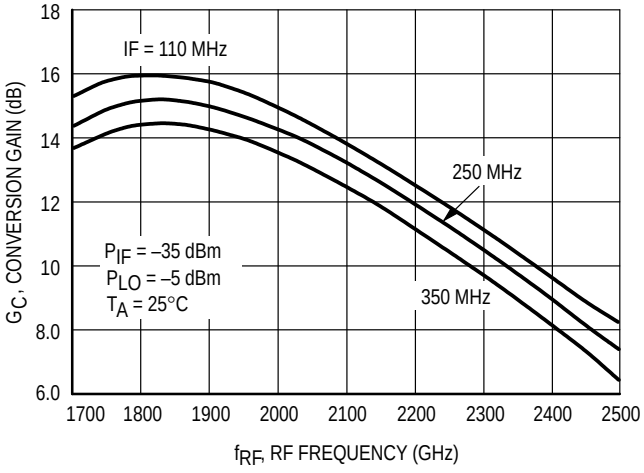


Figure 7. Conversion Gain versus RF Frequency

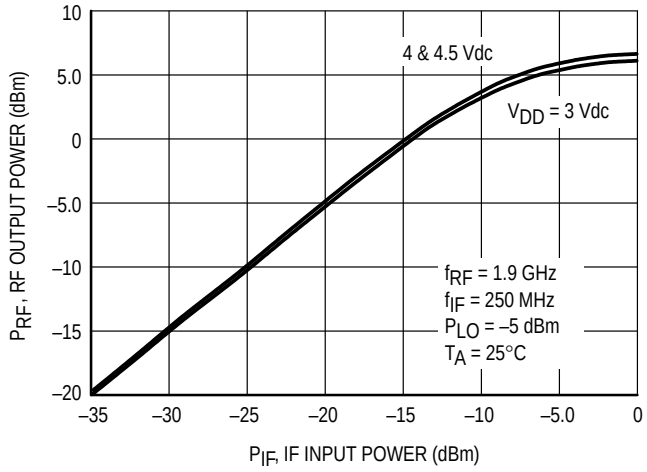


Figure 8. RF Output versus Input Power

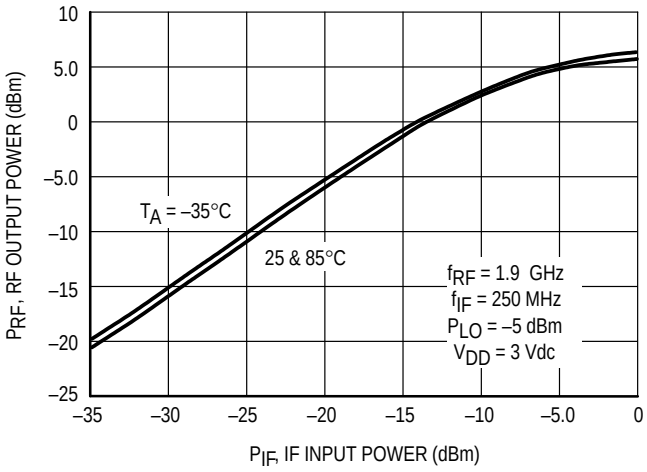


Figure 9. RF Output Power versus IF Input Power

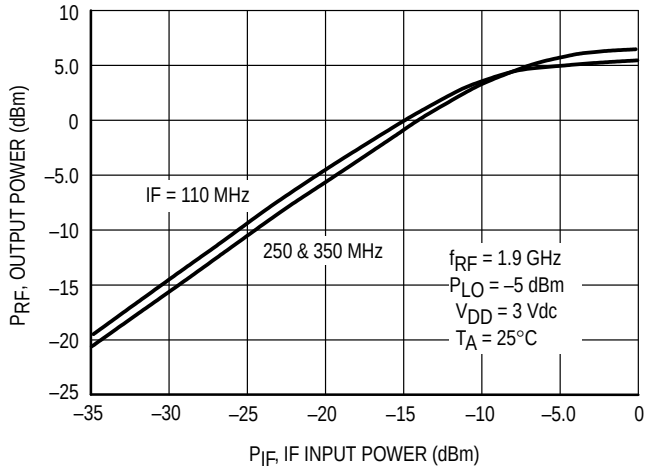


Figure 10. RF Output versus IF Input Power

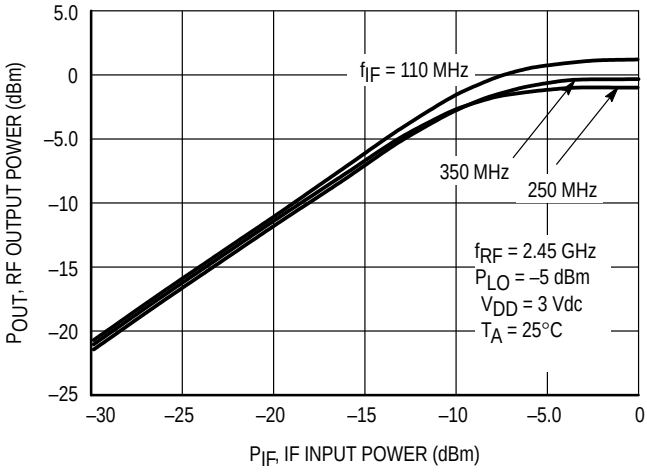


Figure 11. Output Power versus IF Input Power

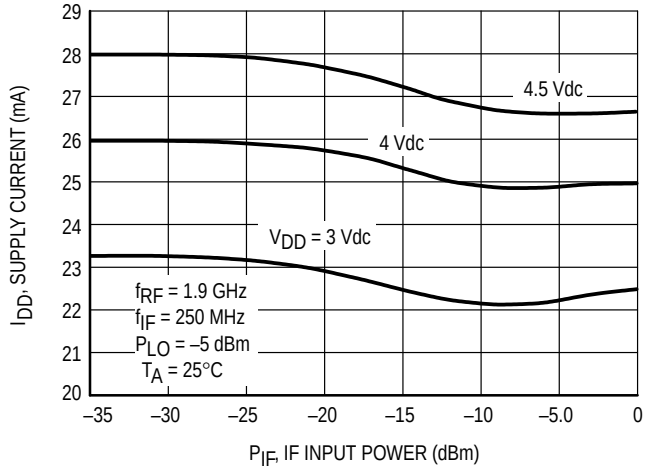


Figure 12. Supply Current versus IF Input Power

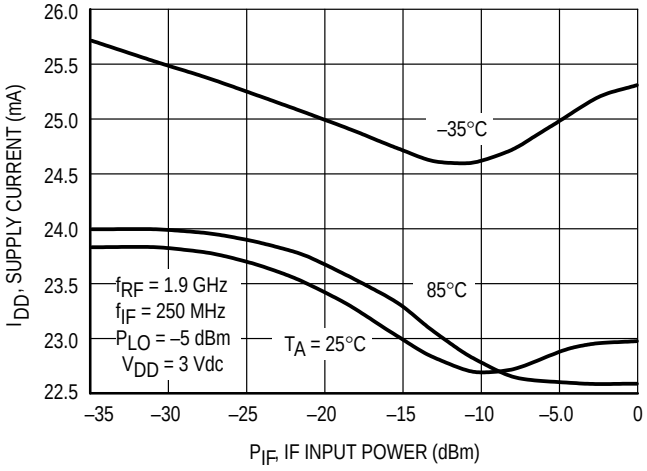


Figure 13. Supply Current versus IF Input Power

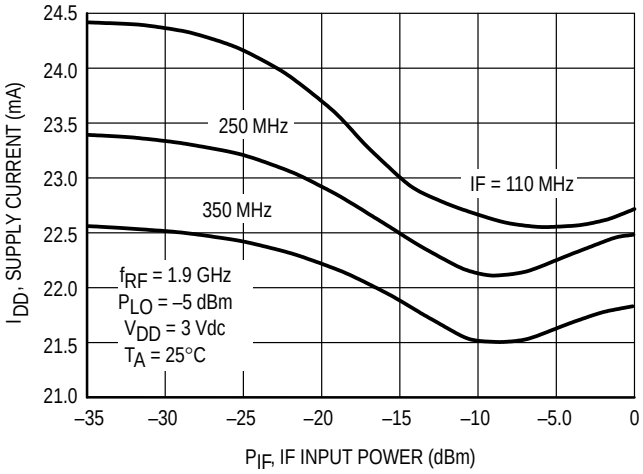


Figure 14. Supply Current versus IF Input Power

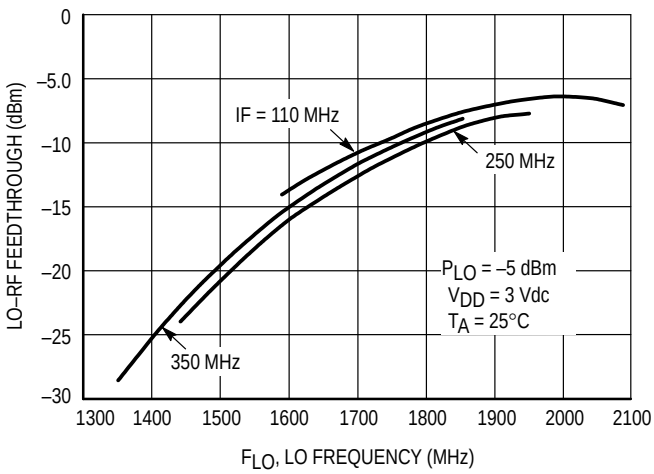


Figure 15. LO to RF Feedthrough versus LO Frequency

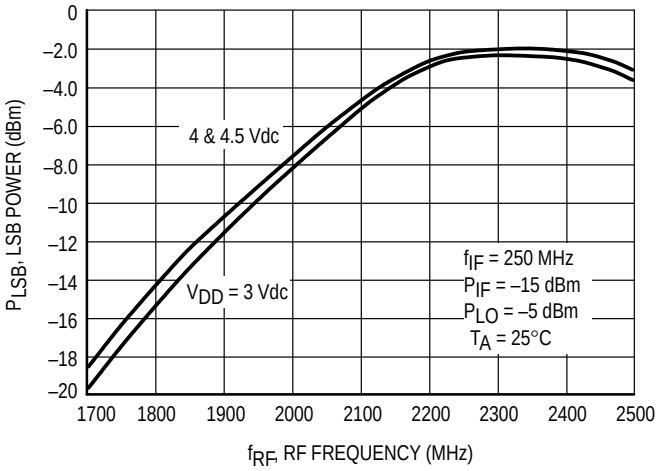


Figure 16. Lower Side Band Power versus RF Frequency

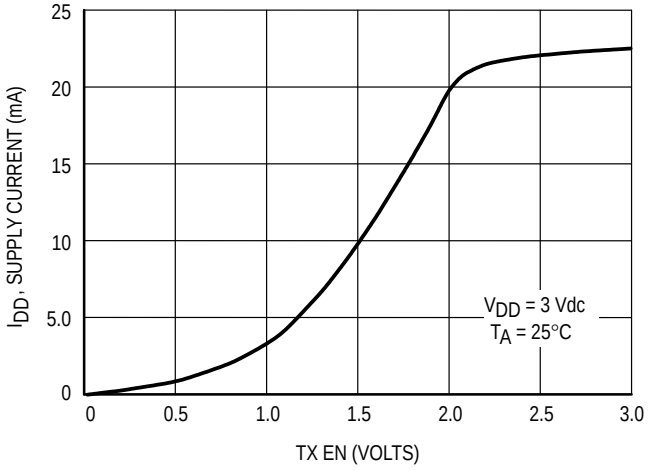


Figure 17. Supply Current versus Transmit Enable Voltage

f	IF Input		LO Input		RF Output (1)	
(MHz)	R	jX	R	jX	R	jX
70	8.3	–452.4				
100	7.3	–318.5				
150	7.1	–211.3				
200	6.6	–156.4				
250	6.5	–123.1				
300	6.1	–100.7				
350	5.7	–84.2				
1100			62.5	3.1		
1200			58.1	4.3		
1300			53.7	4.7		
1400			50.2	4.2		
1500			47.3	3.9		
1600			44.4	3.2		
1700			42.0	1.6	30.4	33.6
1800			40.6	0.5	42.6	16.9
1900			39.6	–0.7	49.1	2.3
2000			38.7	–2.2	40.6	14.2
2100			38.2	–3.6	33.8	17.7
2200			38.4	–5.1	33.3	15.7
2300			38.9	–6.5	32.9	13.7
2400			39.5	–7.8	29.6	13.2
2500					27.4	11.9

(1) Includes T1 shown in Figure 1.

Table 1. Port Impedances versus Frequency (V_{D1}, V_{D2}, V_{D3}, TX EN = 3 Vdc)

APPLICATIONS INFORMATION

DESIGN CONSIDERATIONS

The MRFIC1813 combines a single–balanced MESFET mixer with an exciter amplifier. It is usable for transmit frequencies from 1.7 to 2.5 GHz and IF frequencies from 70 to 350 MHz. The design is optimized for low–side local oscillator injection in hetrodyne transmit applications.

Minimal off–chip matching is required while allowing for flexibility and performance optimization. An active balun is employed at the IF port which gives good balance down to at least 70 MHz. A passive splitter is used at the LO input to complete the single–balanced configuration.

CIRCUIT CONSIDERATIONS

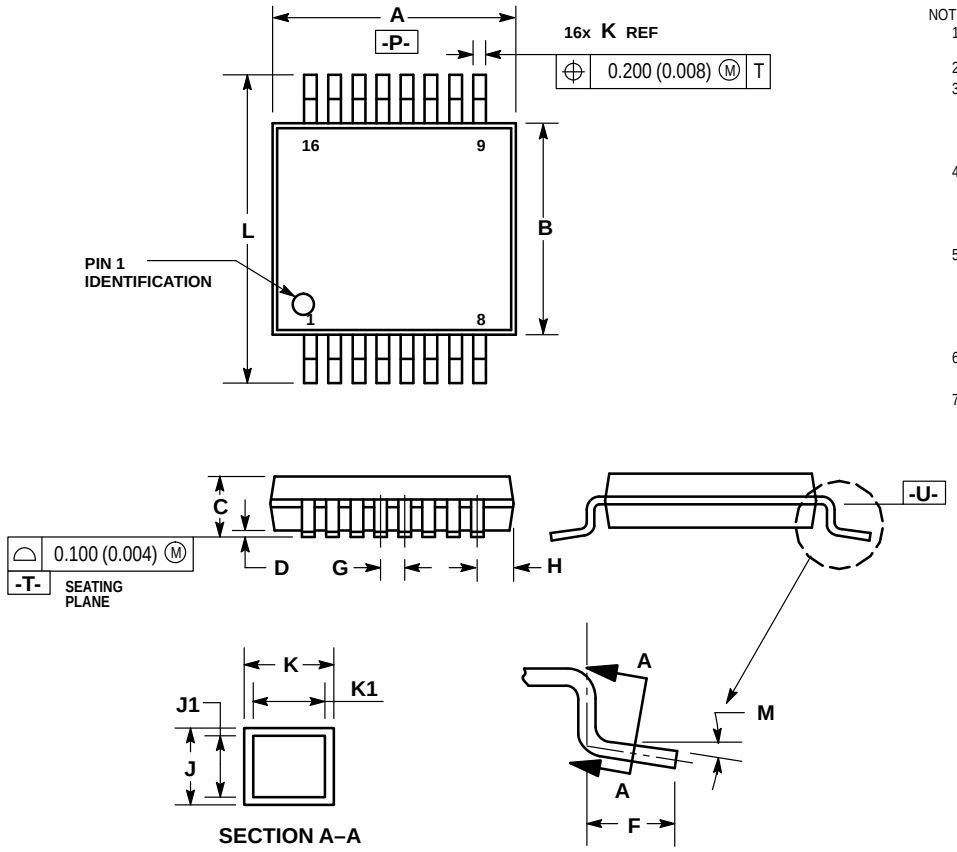
Figure 1 shows the application circuit used to gather the data presented in the characterization curves. As shown in Table 1, the IF port impedance is very high. Three hundred ohms was chosen for R1 to shunt the IF port as a compromise of gain and bandwidth. A 50 Ω resistor can be used and L1 and C5 eliminated to provide a broadband match. The conversion gain is reduced to about 8 dB. Microstrip inductors T1 and T2 combine with inductance internal to the device to form RF chokes. Some tuning of the RF output can be achieved with T1.

As with all RF devices, circuit layout is important. Controlled impedance lines should be used for all RF and IF interconnects. As shown in Figure 1, power supply by–passing should be used to avoid device instability. Ground vias should be included near all ground connections indicated in the schematic. Off–chip components should be mounted as close to the IC leads as possible.

EVALUATION BOARDS

Evaluation boards are available for RF Monolithic Integrated Circuits by adding a “TF” to the device type. For a complete list of currently available boards and one in development for newly introduced products, please contact your local Motorola Distributor or Sales Office.

PACKAGE DIMENSIONS



NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.
4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
5. DIMENSION K DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 (0.003) TOTAL EXCESS OF THE K DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. TERMINAL NUMBERS ARE SHOWN FOR REFERENCE ONLY.
7. DIMENSIONS A AND B ARE TO BE DETERMINED AT DATUM PLANE –U–.

	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
A	——	5.10	——	0.200
B	4.30	4.50	0.169	0.177
C	——	1.20	——	0047
D	0.05	0.25	0.002	0.010
F	0.45	0.55	0.018	0.022
G	0.65 BSC		0.026 BSC	
H	0.22	0.23	0.009	0.010
J	0.09	0.24	0.004	0.009
J1	0.09	0.18	0.004	0.007
K	0.16	0.32	0.006	0.013
K1	0.16	0.26	0.006	0.010
L	6.30	6.50	0.248	0.256
M	0°	10°	0°	10°

CASE 948C–03
ISSUE B

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