1/4

HIGH PERFORMANCE LOW-NOISE OPERATIONAL AMPLIFIER

GENERAL DESCRIPTION

The NJM5534 is a high performance low noise operational amplifier. This amplifier features popular pin-out, superior noise performance, and high output drive capability.

The amplifier also features guaranteed noise performance with substantially higher gain-bandwidth product, power bandwidth, and selw rate which far exceeds that of the NJM741 type amplifiers. The NJM5534 is internally compensated for a gain of three or higher and may be externally compensated for optimizing specific performance requirements of various applications such as unity-gain voltage followers, drivers for capacitive loads or fast setting.

The specially designed low noise input transistors allow the NJM5534 to be used in very low noise signal processing applications such as audio pre-amplifiers and servo error amplifiers.

■ PACKAGE OUTLINE



NJM5534D

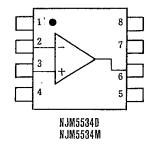


NJM5534M

■ FEATURES

- Operating Voltage
- $(\pm 3V \sim \pm 22V)$
- Single Circuit
- With Vio Trim Terminal
- Low Input Noise Voltage
- $(3.3 \text{nV}/\sqrt{\text{Hz}} \text{ typ. } \text{O1kHz})$
- Power Bandwidth
- (200kHz typ.)
- Slew Rate
- $(13V/\mu s typ.)$
- Package Outline
- DIP8, DMP8
- Bipolar Technology
- . ---

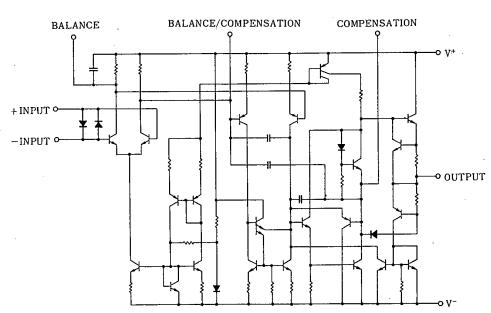
■ PIN CONFIGURATION



PIN FUNCTION

- 1. BALANCE
- 2. -INPUT 3. +INPUT
- 4 . V
- 5. COMPENSATION
- 6. OUTPUT
- 7. V
- 8. BALANCE/COMPENSATION

■ EQUIVALENT CIRCUIT



■ ABSOLUTE MAXIMUM RATINGS

(Ta=25℃)

PARAMETER	SYMBOL	RATINGS	UNIT	
Supply Voltage	(V+/V-)	±22	V	
Differential Input Voltage	V _{ID}	±0.5	V	
Input Voltage	V _{IC}	V+/V	V	
Power Dissipation	PD	(DIP8) 500	mW	
		(DMP8) 300	mW	
Operating Temperature Range	Topr	-20~+75	C	
Storage Temperature Range	Tstg	-40~+125	r	

■ ELECTRICAL CHARACTERISTICS

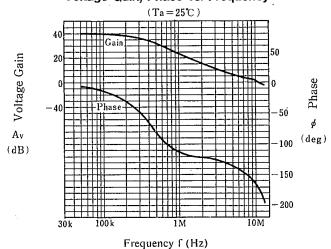
 $(Ta=25^{\circ}C, V^{+}/V^{-}=\pm 15V)$

PARAMETER		TEST CONDITION	NJM5534			
	SYMBOL		MIN.	TYP.	MAX.	UNIT
Input Offset Voltage	V _{IO}	R _S ≦10kΩ .	_	0.5	4	mV
Input Offset Current	I _{1O}		_	20	300	nA
Input Bias Current	I_{B}			500	1,500	nA
Input Resistance	R _{IN}		30	100	1	kΩ
Large Signal Voltage Gain	A _v	$R_{L} \ge 2k\Omega$, $V_{o} = \pm 10V$	88	100		dB
Maximum Output Voltage Swing	V _{OM}	R _{1.} ≥600Ω	±12	±13		v
Input Common Mode Voltage Range	V _{ICM}		±12	±13	_	v
Common Mode Rejection Ratio	CMR	R _S ≤10kΩ	70	100	_	dB
Supply Voltage Rejection Ratio	SVR	R _s ≤10kΩ	80	100		dB ·
Operating Current	1 _{cc}	$R_{L} = \infty$	_	4	8	mA
Transient Response Rise Time	t _R	$V_{1N}=50 \text{mV}, R_L=600\Omega, C_L=100 \text{pF}, C_c=22 \text{pF}$	_	35	_	nsec
Overshoot	ł	$V_{IN}=50 \text{mV}, R_L=600\Omega, C_L=100 \text{pF}, C_e=22 \text{pF}$		17	_	%
Slew Rate	SR	$C_c=0$	_	13		V/μS
Gain Bandwidth Product	GB	$C_c = 22pF, C_L = 100pF$		10	_	MHz
Power Bandwidth	WPG	$V_o = 20V_{p-p}, C_c = 0$	—	200	_	kHz
Equivalent Input Noise Voltage	V _{NI}	f=20Hz~20kHz		1.0		μVrms
Equivalent Input Noise Current	I _{NI}	f=20Hz~20kHz	_	25	—	pArms
Equivalent Input Noise Voltage 1	e _n 1	$f_0=30Hz$		5.5	_	nV/√Hz
Equivalent Input Noise Voltage 2	e _{n 2}	f _o =1kHz	-	3.3		nV/√Hz
Equivalent Input Noise Current 1	i _{n 1}	$\int_{c} = 30$ Hz	_	1.5	<u> </u>	pA/√Hz
Equivalent Input Noise Current 2	in 2	$f_0 = 1 \text{kHz}$		0.4		pA/√Hz
Broadband Noise Figure	NF	$f=10$ Hz \sim 20kHz, R _S =5k Ω	-	0.9	_	dB

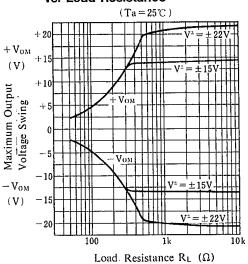
Note: JRC's general selected products D rank are also prepared for the noise standard ($R_S=2.2k\Omega$, RIAA, $V_N=1.4\mu V$ Max.)

■ TYPICAL CHARACTERISTICS

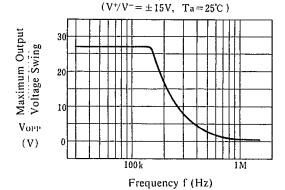
Voltage Gain, Phase vs. Frequency



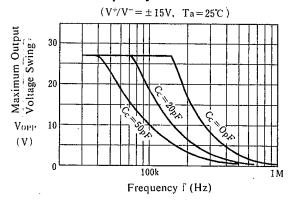
Maximum Output Voltage Swing vs. Load Resistance



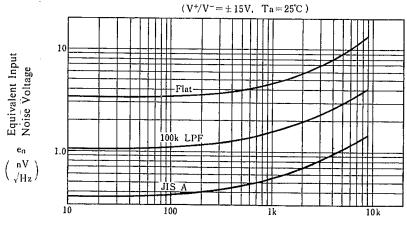
Maximum Output Voltage Swing vs. Frequency



Maximum Output Voltage Swing vs. Frequency



Equivalent Input Noise Voltage vs. Rs

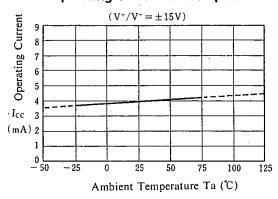


Source Resistance R_S (Ω)

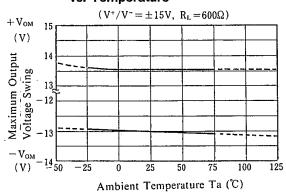
1

TYPICAL CHARACTERISTICS

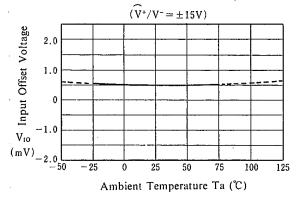
Operating Current vs. Temperature



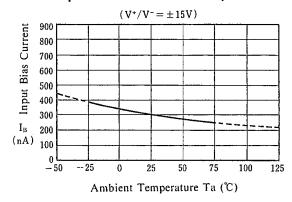
Maximum Output Voltage Swing vs. Temperature



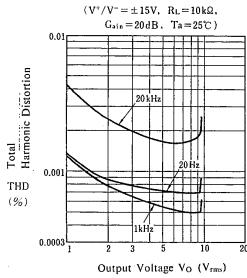
Input Offset Voltage vs. Temperature

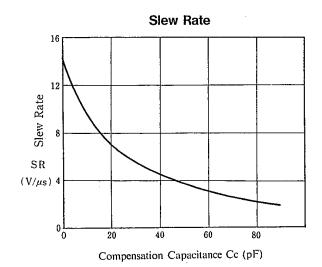


Input Bias Current vs. Temperature



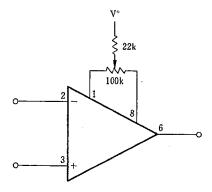
Total Harmonic Distortion vs. Output Voltage



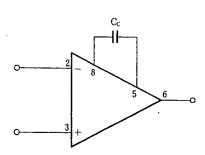


■ ADJUSTMENT METHOD

Offset Adjustment



Frequency Compensation



■ NOTICE

When used in voltage follower circuit, put a current limit resistor into non-inverting input terminal in order to avoid inside input diode destruction when the power supply is turned on. (ref. Fig. 1)

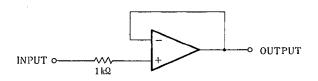


Fig. 1

MEMO

[CAUTION]
The specifications on this databook are only given for information , without any guarantee as regards either mistakes or omissions. The application circuits in this databook are described only to show representative usages of the product and not intended for the guarantee or permission of any right including the industrial rights.