

DATA SHEET

OQ2536HP **SDH/SONET STM16/OC48** **demultiplexer**

Product specification
File under Integrated Circuits, IC19

1998 Mar 10

SDH/SONET STM16/OC48 demultiplexer

OQ2536HP

FEATURES

- Normal and loop (test) modes
- 1.2 V GTL (Gunning Transceiver Logic) level compatible data and clock outputs (low speed interface)
- Differential CML (Current-Mode Logic) data and clock inputs
- High input sensitivity (100 mV for the high speed inputs)
- Boundary Scan Test (BST) at low speed interface, in accordance with "IEEE Std 1149.1-1990"
- Low power dissipation (typically 1.45 W).

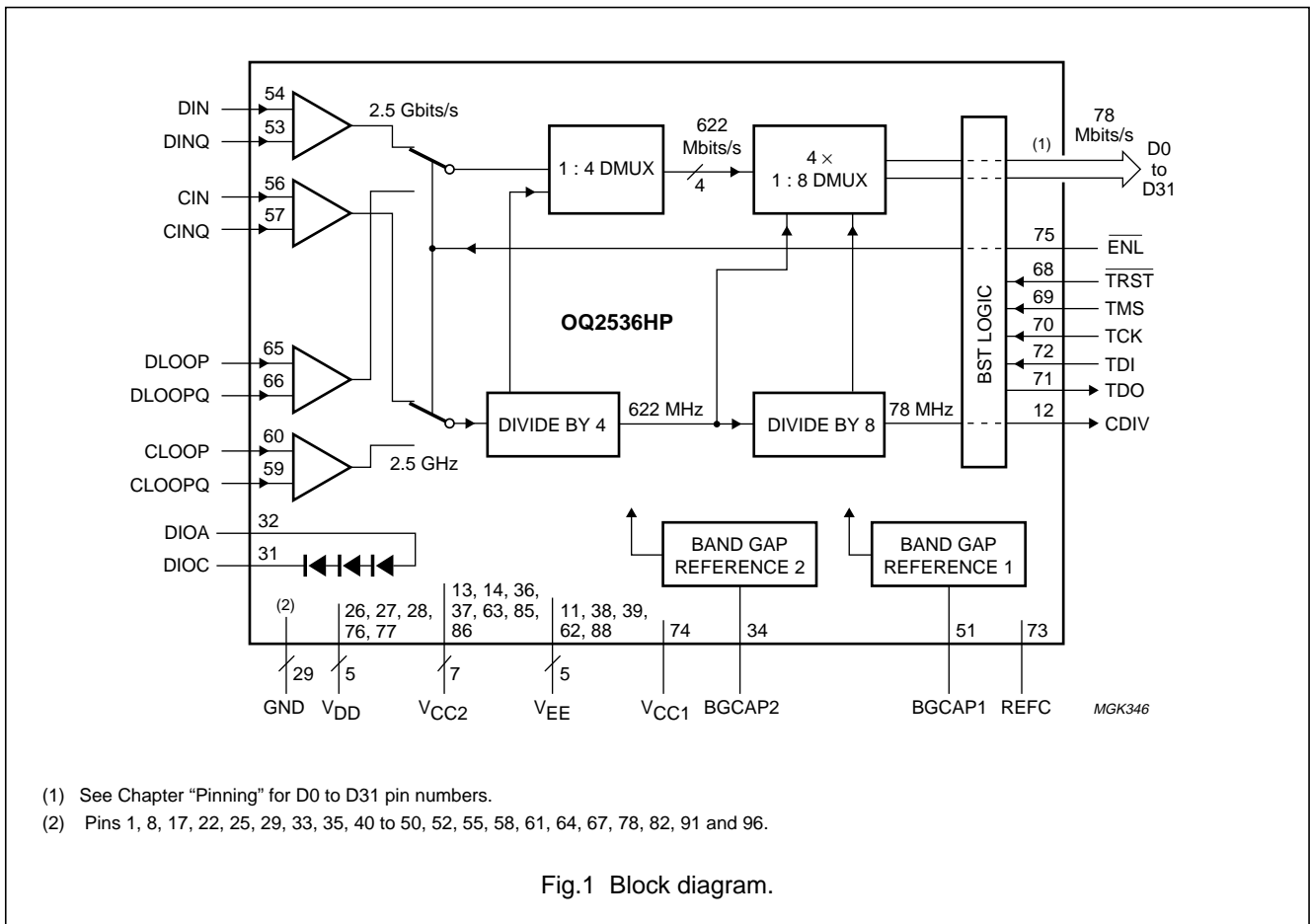
DESCRIPTION

The OQ2536HP is a 32-channel demultiplexer intended for use in STM16/OC48 applications. It demultiplexes a single 2.5 Gbits/s input channel to 32 × 78 Mbits/s output channels. The data and clock outputs on the low speed interface are GTL compatible, while the high speed data and clock inputs are CML compatible.

ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
OQ2536HP	HLQFP100	plastic heat-dissipating low profile quad flat package; 100 leads; body 14 × 14 × 1.4 mm	SOT470-1

BLOCK DIAGRAM



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PINNING

SYMBOL	PIN	TYPE ⁽¹⁾	DESCRIPTION
GND	1	S	ground
D29	2	O	78 Mbits/s data output channel for D29
D25	3	O	78 Mbits/s data output channel for D25
D21	4	O	78 Mbits/s data output channel for D21
D17	5	O	78 Mbits/s data output channel for D17
D13	6	O	78 Mbits/s data output channel for D13
D9	7	O	78 Mbits/s data output channel for D9
GND	8	S	ground
D5	9	O	78 Mbits/s data output channel for D5
D1	10	O	78 Mbits/s data output channel for D1
V _{EE}	11	S	supply voltage (−4.5 V)
CDIV	12	O	78 MHz clock output
V _{CC2}	13	S	supply voltage (+1.5 V)
V _{CC2}	14	S	supply voltage (+1.5 V)
D28	15	O	78 Mbits/s data output channel for D28
D24	16	O	78 Mbits/s data output channel for D24
GND	17	S	ground
D20	18	O	78 Mbits/s data output channel for D20
D16	19	O	78 Mbits/s data output channel for D16
D12	20	O	78 Mbits/s data output channel for D12
D8	21	O	78 Mbits/s data output channel for D8
GND	22	S	ground
D4	23	O	78 Mbits/s data output channel for D4
D0	24	O	78 Mbits/s data output channel for D0
GND	25	S	ground
V _{DD}	26	I	supply voltage (+3.3 V)
V _{DD}	27	I	supply voltage (+3.3 V)
V _{DD}	28	I	supply voltage (+3.3 V)
GND	29	S	ground
i.c.	30	–	internally connected, to be left open-circuit
DIOC	31	A	cathode of temperature diode array
DIOA	32	A	anode of temperature diode array
GND	33	S	ground
BGCAP2	34	A	pin for connecting external band gap decoupling capacitor (4 × 1 : 8 DMUX)
GND	35	S	ground
V _{CC2}	36	S	supply voltage (+1.5 V)
V _{CC2}	37	S	supply voltage (+1.5 V)
V _{EE}	38	S	supply voltage (−4.5 V)
V _{EE}	39	S	supply voltage (−4.5 V)
GND	40	S	ground

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SYMBOL	PIN	TYPE ⁽¹⁾	DESCRIPTION
GND	41	S	ground
GND	42	S	ground
GND	43	S	ground
GND	44	S	ground
GND	45	S	ground
GND	46	S	ground
GND	47	S	ground
GND	48	S	ground
GND	49	S	ground
GND	50	S	ground
BGCAP1	51	A	pin for connecting external band gap decoupling capacitor (1 : 4 DMUX)
GND	52	S	ground
DINQ	53	I	inverted data input in normal mode
DIN	54	I	data input in normal mode
GND	55	S	ground
CIN	56	I	clock input in normal mode
CINQ	57	I	inverted clock input in normal mode
GND	58	S	ground
CLOOPQ	59	I	inverted clock input from multiplexer IC OQ2535 (loop mode)
CLOOP	60	I	clock input from multiplexer IC OQ2535 (loop mode)
GND	61	S	ground
V _{EE}	62	S	supply voltage (-4.5 V)
V _{CC2}	63	S	supply voltage (+1.5 V)
GND	64	S	ground
DLOOP	65	I	data input from multiplexer IC OQ2535 (loop mode)
DLOOPQ	66	I	inverted data input from multiplexer IC OQ2535 (loop mode)
GND	67	S	ground
$\overline{\text{TRST}}$	68	I	test RESET input for BST mode (active LOW)
TMS	69	I	test mode select input for BST
TCK	70	I	test clock input for BST mode
TDO	71	O	serial test data output for BST mode
TDI	72	I	serial test data input for BST mode
REFC	73	A	pin for connecting external reference decoupling capacitor (for standard TTL reference)
V _{CC1}	74	S	supply voltage (+5.0 V)
$\overline{\text{ENL}}$	75	I	loop mode enable input (active LOW)
V _{DD}	76	I	supply voltage (+3.3 V)
V _{DD}	77	I	supply voltage (+3.3 V)
GND	78	S	ground
D31	79	O	78 Mbits/s data output channel for D31
D27	80	O	78 Mbits/s data output channel for D27

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SYMBOL	PIN	TYPE ⁽¹⁾	DESCRIPTION
D23	81	O	78 Mbits/s data output channel for D23
GND	82	S	ground
D19	83	O	78 Mbits/s data output channel for D19
D15	84	O	78 Mbits/s data output channel for D15
V _{CC2}	85	S	supply voltage (+1.5 V)
V _{CC2}	86	S	supply voltage (+1.5 V)
D11	87	O	78 Mbits/s data output channel for D11
V _{EE}	88	S	supply voltage (-4.5 V)
D7	89	O	78 Mbits/s data output channel for D7
D3	90	O	78 Mbits/s data output channel for D3
GND	91	S	ground
D30	92	O	78 Mbits/s data output channel for D30
D26	93	O	78 Mbits/s data output channel for D26
D22	94	O	78 Mbits/s data output channel for D22
D18	95	O	78 Mbits/s data output channel for D18
GND	96	S	ground
D14	97	O	78 Mbits/s data output channel for D14
D10	98	O	78 Mbits/s data output channel for D10
D6	99	O	78 Mbits/s data output channel for D6
D2	100	O	78 Mbits/s data output channel for D2

Note

1. Pin type abbreviations: O = Output, I = Input, S = power Supply, A = Analog function.

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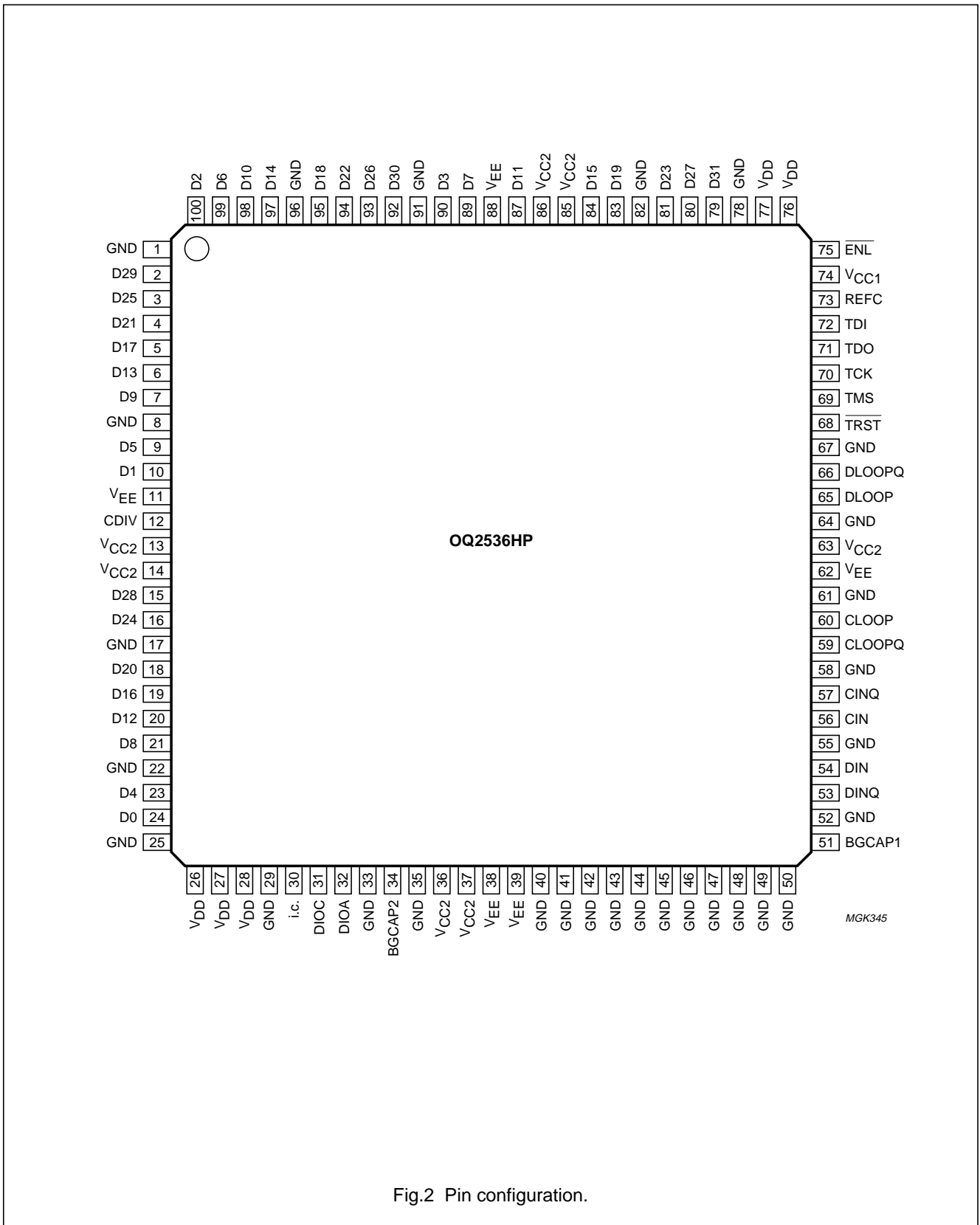


Fig.2 Pin configuration.

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FUNCTIONAL DESCRIPTION

The OQ2536HP is a 32-channel demultiplexer, intended for use in STM16/OC48 applications. It demultiplexes a single 2.5 Gbits/s input channel to 32×78 Mbits/s output channels.

The demultiplexing is performed in two stages. The 2.5 Gbits/s data channel is first demultiplexed to four 622 Mbits/s data channels. Each of these channels is then fed to a 1 : 8 demultiplexer to generate 32×78 Mbits/s output channels.

The $\overline{\text{ENL}}$ control input is used for switching between normal and loop modes. When loop mode is enabled ($\overline{\text{ENL}} = \text{LOW}$), inputs DLOOP, DLOOPQ, CLOOP and CLOOPQ are selected. In normal mode ($\overline{\text{ENL}} = \text{HIGH}$), inputs DIN, DINQ, CIN and CINQ are selected.

The signal applied to CIN and CINQ is a 2.5 GHz recovered clock signal, e.g. coming from the OQ2541 data and clock recovery IC. The clock is divided down to 78 MHz, which is used for receive logic timing and is available as a GTL compatible output at pin CDIV.

High bit rate stage: 1 : 4 DMUX

The 2.5 Gbits/s data stream is fed into a 1 : 4 demultiplexer to generate four 622 Mbits/s channels.

The input pins DIN, DINQ, DLOOP, DLOOPQ, CIN, CINQ, CLOOP and CLOOPQ are terminated internally with 50Ω resistors to GND.

Low bit rate part: 4 × 1 : 8 DMUX

The four 622 Mbits/s output channels coming from the high bit rate stage are loaded into four 8-bit shift registers. The 622 MHz clock for these shift registers comes from the preceding stage.

The 32 bits contained in the shift registers are loaded into latches and made available on outputs D0 to D31. These outputs are 1.2 V GTL compatible and have internal 100Ω pull up resistors. The 78 MHz clock output, CDIV, has an internal 50Ω pull up resistor.

The first serial data bit coming in at DIN or DLOOP is given out at pin D31 (MSB) and so on.

The data outputs may not always represent four STM bytes. This is because the internal load pulse for the output latches is not synchronized to the STM16 frame.

Power supply connections

The power supply pins need to be individually decoupled using chip capacitors mounted as close as possible to the

IC. If multiple decoupling capacitors are used for a single supply node, large distance between the capacitances should be avoided in order to avoid resonance.

To minimize low frequency switching noise in the vicinity of the OQ2536HP, all power supply lines should be filtered once by an LC-circuit with a low cutoff frequency (as shown in the application diagram, Fig.7).

Ground connection

The ground connection on the PCB needs to be a large copper area fill connected to a common ground plane with low inductance.

RF connections

A coupled stripline or microstrip with an odd mode characteristic impedance of 50Ω (nominal value) should be used for the RF connections on the PCB.

The connections should be kept as short as possible. This applies to the CML differential line pairs CIN and CINQ, DIN and DINQ, CLOOP and CLOOPQ, and DLOOP and DLOOPQ. In addition, the following lines should not vary in length by more than 5 mm:

- CIN, CINQ, DIN and DINQ
- DLOOP, DLOOPQ, CLOOP and CLOOPQ.

Interface to receive logic

The 78 Mbits/s interface lines, CDIV and D0 to D31, should not exceed 50 mm in length. The parasitic capacitance of these lines should be as small as possible (less than 3 pF is desirable).

ESD protection

All pads are protected by ESD protection diodes, with the exception of the high frequency inputs DIN, DINQ, DLOOP, DLOOPQ, CIN, CINQ, CLOOP and CLOOPQ.

Cooling

In many cases it is necessary to mount a special cooling device on the package. The thermal resistance from junction to case, $R_{\text{th}j-c}$ and from junction to ambient, $R_{\text{th}j-a}$, are given in Chapter "Thermal characteristics". Since the heat-slug in the package is connected to the die, the cooling device should be electrically isolated.

To calculate if a heatsink is necessary, the maximum allowed total thermal resistance R is calculated as:

$$R_{\text{th}} = \frac{T_j - T_{\text{amb}}}{P_{\text{tot}}} \quad (1)$$

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where:

R_{th} = total thermal resistance from junction to ambient in the application

T_j = junction temperature

T_{amb} = ambient temperature.

As long as R_{th} is greater than R_{thj-a} of the OQ2536HP including environmental conditions like air flow and board layout, no heatsink is necessary. For example if $T_j = 120\text{ °C}$, $T_{amb} = 70\text{ °C}$ and $P_{tot} = 1.45\text{ W}$, then:

$$R_{th} = \frac{(120 - 70)}{1.45} = 34.4 \quad [\text{K/W}] \quad (2)$$

which is more than the worst case $R_{thj-a} = 33\text{ K/W}$, so no heatsink is necessary.

Another example; if for safety reasons T_j should stay as low as 110 °C , while $T_{amb} = 85\text{ °C}$ and $P_{tot} = 2\text{ W}$, then:

$$R_{th} = \frac{(110 - 85)}{2.0} = 12.5 \quad [\text{K/W}] \quad (3)$$

In this case extra cooling is needed. The thermal resistance of the heatsink is calculated as follows:

$$R_{thh-a} \leq \left(\frac{1}{R_{th}} - \frac{1}{R_{thj-a}} \right)^{-1} - R_{thj-c} - R_{thc-h} \quad (4)$$

where:

R_{thh-a} = thermal resistance from heatsink to ambient

R_{thc-h} = thermal resistance from case to heatsink

R_{thj-c} = thermal resistance from junction to case, see Chapter "Thermal characteristics".

If for instance $R_{thc-h} = 0.5\text{ K/W}$ and $R_{thj-a} = 33\text{ K/W}$ then:

$$R_{thh-a} \leq \left(\frac{1}{12.5} - \frac{1}{33} \right)^{-1} - 3.1 \leq 17.0 \quad [\text{K/W}] \quad (5)$$

Built in temperature sensor

Three series-connected diodes have been integrated for measuring junction temperature. The diode array, accessed by means of the DIOA (anode) and DIOC (cathode) pins, has a temperature dependency of approximately -6 mV/°C . With a diode current of 1 mA , the voltage will be somewhere in the range 1.7 to 2.5 V , depending on temperature.

Boundary Scan Test (BST) interface

Boundary scan test logic has been implemented for all digital inputs and outputs on the low frequency interface, in accordance with "IEEE Std 1149.1-1990". All scan tests other than SAMPLE mode are available. The boundary scan test logic consists of a TAP controller, a BYPASS register, a 2-bit instruction register, a 32-bit identification register and a 36-bit boundary scan register (the last two are combined). The architecture of the TAP controller and the BYPASS register is in accordance with IEEE recommendations.

The four command modes, selected by means of the instruction register, are: EXTEST (00), PRELOAD (01), IDCODE (10) and BYPASS (11).

All boundary scan test inputs, TDI, TMS, TCK and $\overline{\text{TRST}}$, have internal pull up resistors. The maximum test clock frequency at TCK is 12 MHz .

Table 1 BST identifier code

VERSION	OQ	2536 (BINARY)	PHILIPS SEMICONDUCTORS	LSB ⁽¹⁾
0001	01	00 1001 1110 1000	0000 0011 101 ⁽²⁾	1

Notes

1. LSB is shifted out first on the TDO pin.
2. The manufacturer's code was implemented incorrectly. It should have been 0000 0010 101.

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Table 2 BST bit order

BIT NUMBER	SYMBOL	PIN
33 (MSB)	D31	79
32	D27	80
31	D23	81
30	D19	83
29	D15	84
28	D11	87
27	D7	89
26	D3	90
25	D30	92
24	D26	93
23	D22	94
22	D18	95
21	D14	97
20	D10	98
19	D6	99
18	D2	100
17	D29	2
16	D25	3
15	D21	4
14	D17	5
13	D13	6
12	D9	7
11	D5	9
10	D1	10
9	CDIV	12
8	D28	15
7	D24	16
6	D20	18
5	D16	19
4	D12	20
3	D8	21
2	D4	23
1	D0	24
0 (LSB) ⁽¹⁾	$\overline{\text{ENL}}$	75

Note

1. LSB is shifted out first on the TDO pin.

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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
V _{CC1}	supply voltage	-0.5	+6.0	V
V _{EE}	supply voltage	-6.0	+0.5	V
V _{DD}	supply voltage	-0.5	+5.0	V
V _{CC2}	supply voltage	-0.5	+2.0	V
V _n	DC voltage			
	pins 2 to 7, 9, 10, 15, 16, 18 to 21, 23, 24, 79, 80, 81, 83, 84, 87, 89, 90, 92 to 95 and 97 to 100	0.0	2.0	V
	pins 53, 54, 56, 57, 59, 60, 65 and 66	-1.0	+0.5	V
	pins 68, 69, 70, 72, 73 and 75	-0.5	V _{CC1} + 0.5	V
	pins 30, 34 and 51	V _{EE} - 0.5	0.5	V
	pins 31 and 32	V _{EE} - 0.5	V _{CC1} + 0.5	V
I _n	DC current			
	pins 2 to 7, 9, 10, 15, 16, 18 to 21, 23, 24, 79, 80, 81, 83, 84, 87, 89, 90, 92 to 95, and 97 to 100	-	15	mA
	pin 12	-	30	mA
	pins 31 and 32	-	10	mA
	pin 71	-	50	mA
P _{tot}	total power dissipation	-	2.6	W
T _j	junction temperature	-	120	°C
T _{stg}	storage temperature	-65	+150	°C

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
R _{th j-c}	thermal resistance from junction to case		2.6	K/W
R _{th j-a}	thermal resistance from junction to ambient	see note 1		
		airflow = 0 ft/min	33	K/W
		airflow = 100 ft/min	28	K/W
		airflow = 200 ft/min	25	K/W
		airflow = 400 ft/min	22	K/W
		airflow = 600 ft/min	20	K/W

Note

- The thermal resistance from junction to ambient is strongly depending on the board design and airflow. The values given in the table are typical values and are measured on a single sided test board with dimensions of 76 × 114 × 1.6 mm. Better values can be obtained when mounted on multilayer boards with large ground planes.

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DC CHARACTERISTICS

Typical values at $T_{amb} = 25\text{ °C}$ and at typical supply voltages; minimum and maximum values are valid over the entire ambient temperature range and supply voltage range.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
General						
V_{CC1}	supply voltage		4.75	5.0	5.25	V
V_{EE}	supply voltage		-4.75	-4.5	-4.25	V
V_{DD}	supply voltage		3.14	3.3	3.47	V
V_{CC2}	supply voltage		1.1	1.5	1.6	V
I_{CC1}	supply current		-	14	22	mA
I_{EE}	supply current		-	170	215	mA
I_{DD}	supply current		-	100	185	mA
I_{CC2}	supply current	note 1	-	190	525	mA
P_{tot}	total power dissipation	note 1	-	1.45	2.6	W
T_j	junction temperature		-	-	+120	°C
T_{amb}	ambient temperature		-40	-	+85	°C
TTL input: \overline{ENL}						
V_{IL}	LOW-level input voltage		-	-	0.8	V
V_{IH}	HIGH-level input voltage		2.0	-	-	V
I_{IL}	LOW-level input current		-90	-	-	μA
I_{IH}	HIGH-level input current		-	-	210	μA
TTL inputs: TDI, TCK, TMS and \overline{TRST}; note 2						
V_{IL}	LOW-level input voltage		-	-	0.4	V
V_{IH}	HIGH-level input voltage		2.0	-	-	V
CML inputs: CIN, Cinq, DIN, Dinq, CLOOP, CLOOPQ, DLOOP and DLOOPQ; note 3						
$V_{i(p-p)}$	input voltage (peak-to-peak value)	50 Ω measurement system	100	250	500	mV
V_{IO}	permitted input offset voltage		-25	-	+25	mV
$V_{I,IQ}$	input voltages		-600	-	+250	mV
Z_i	single ended input impedance	for DC signal	-	50	-	Ω
TTL output: TDO; note 4						
V_{OL}	LOW-level output voltage	$I_{OL} = 4\text{ mA}$	-	0.3	0.5	V

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{OH}	HIGH-level output voltage	$I_{OH} = -400 \mu A$	2.4	4.0	–	V
I_{OZ}	output current in high impedance state		–	–	1	μA
Outputs: CDIV and D0 to D31; notes 5 and 6						
V_{OL}	LOW-level output voltage	Open outputs	–	0.3	0.4	V
V_{OH}	HIGH-level output voltage; note 7		1.1	1.5	1.6	V
Temperature diode array						
$\Delta V_{DIOA-DIOC}$	diode voltage range ⁽⁸⁾	$I_{I(d)} = 1 \text{ mA}$	–	2.1	–	V

Notes

1. Maximum current I_{CC2} and maximum power dissipation P_{tot} are worst case figures i.e. data outputs D0 to D31 remain in LOW state.
2. TDI, TCK, TMS and \overline{TRST} are connected via 90 k Ω to V_{DD} .
3. See Fig.3 for symbol definitions.
4. TDO is switched to high impedance state if BST is inactive.
5. Output CDIV has an internal pull-up resistor of 50 Ω to V_{CC2} . Outputs D0 to D31 have internal pull-up resistors of 100 Ω to V_{CC2} .
6. The first serial data bit coming in at DIN or DLOOP is given out at D31 (MSB) and so on.
7. The HIGH-level output voltage depends on the supply voltage V_{CC2} .
8. The temperature diode array can be used to measure the temperature of the die. The temperature dependency of this voltage is approximately –6 mV/K.

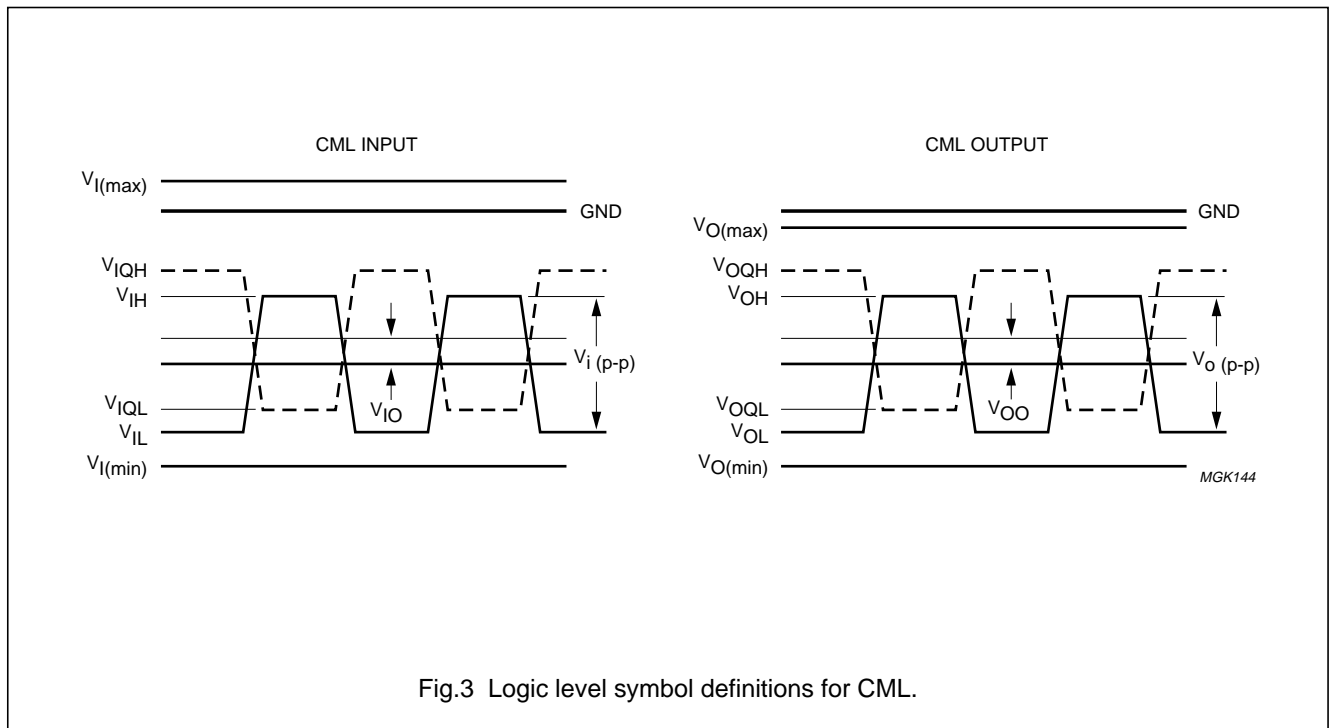


Fig.3 Logic level symbol definitions for CML.

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TIMING

Typical values at $T_{amb} = 25\text{ }^{\circ}\text{C}$ and at typical supply voltages; minimum and maximum values are valid over the entire ambient temperature range and supply voltage range.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
CML input timing; note 1; Fig.5						
$f_{clk(CIN)}$	input clock frequency		2.488	–	–	GHz
t_{su}	input data set-up time		140	–	–	ps
t_h	input data hold time		80	–	–	ps
SR_{CIN}	clock slew rate		1	–	–	V/ns
TTL output timing; note 2; Fig.6						
$f_{clk(CDIV)}$	output clock frequency	$f_{clk(CDIV)} = 2.488\text{ GHz}$	–	77.76	–	MHz
δ_{CDIV}	output clock duty factor		–	50	–	%
$t_{r(CDIV)}$	output clock rise time	Measured between 10% and 90% levels of full output swing	–	–	2700	ps
$t_{f(CDIV)}$	output clock fall time		–	–	1000	ps
$t_{r(D0\text{ to }D31)}$	data out rise time		–	–	5100	ps
$t_{f(D0\text{ to }D31)}$	data out fall time		–	–	1000	ps
t_{CDV}	clock edge to data valid time		–	–	2700	ps
t_{DI}	data invalid time		–	–	2850	ps

Notes

1. The specified timing characteristics are applicable in both normal and loop modes.
2. A capacitive load of 15 pF was connected at all outputs. An input reference level of 1 V was used.

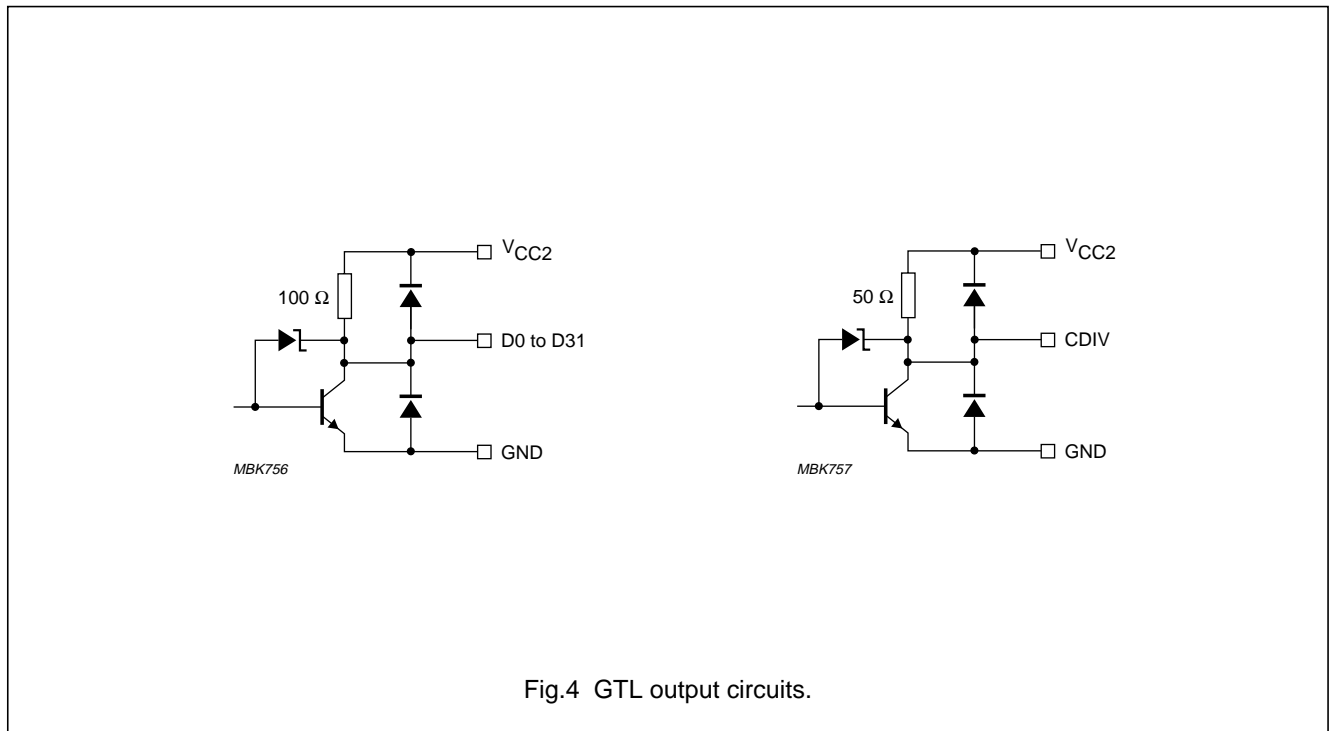
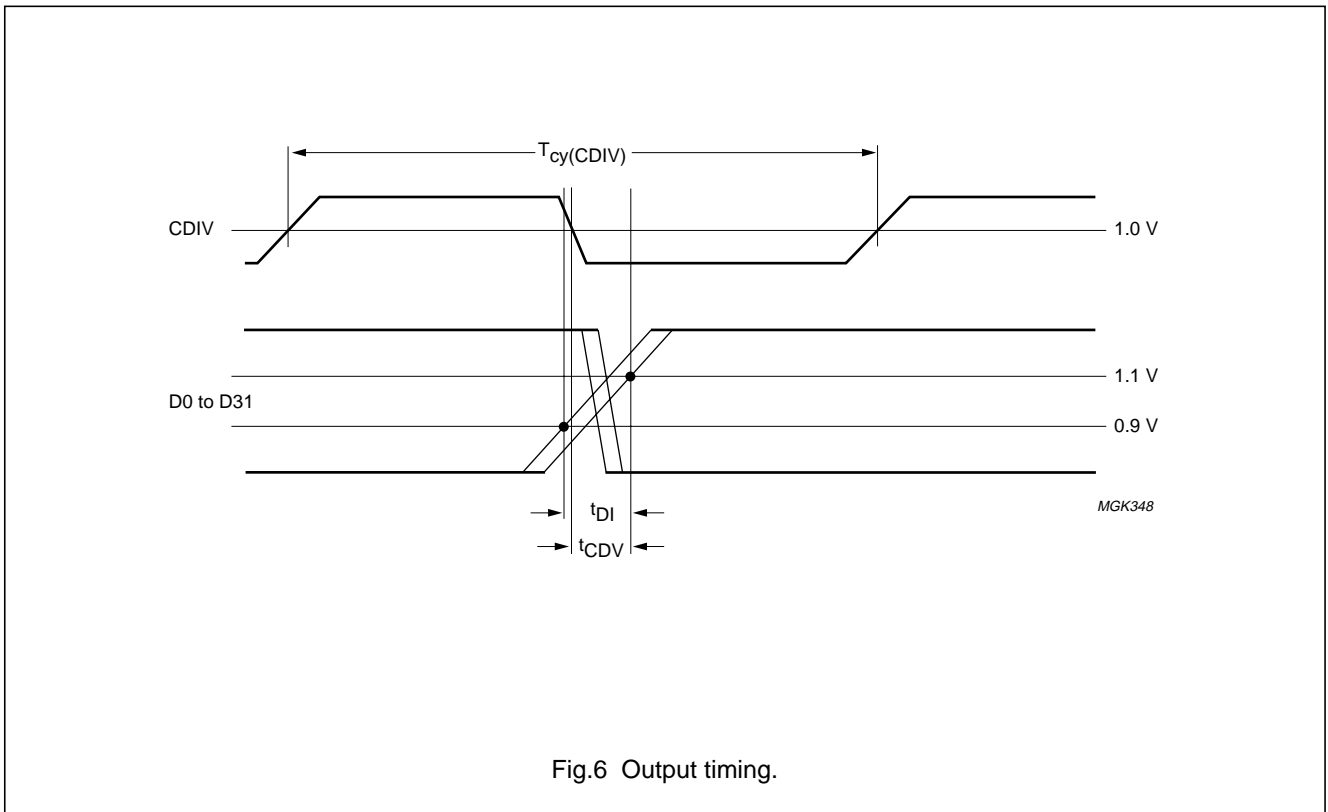
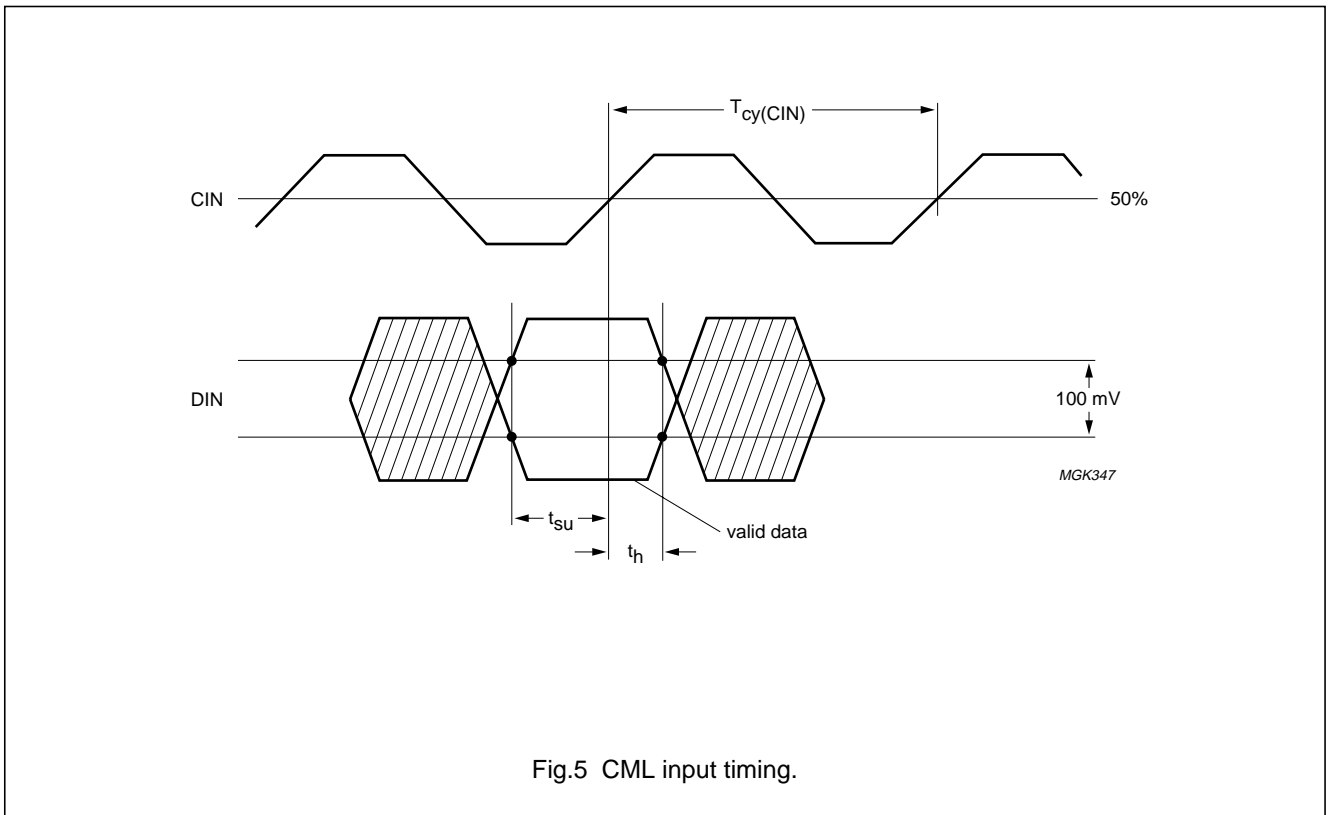


Fig.4 GTL output circuits.

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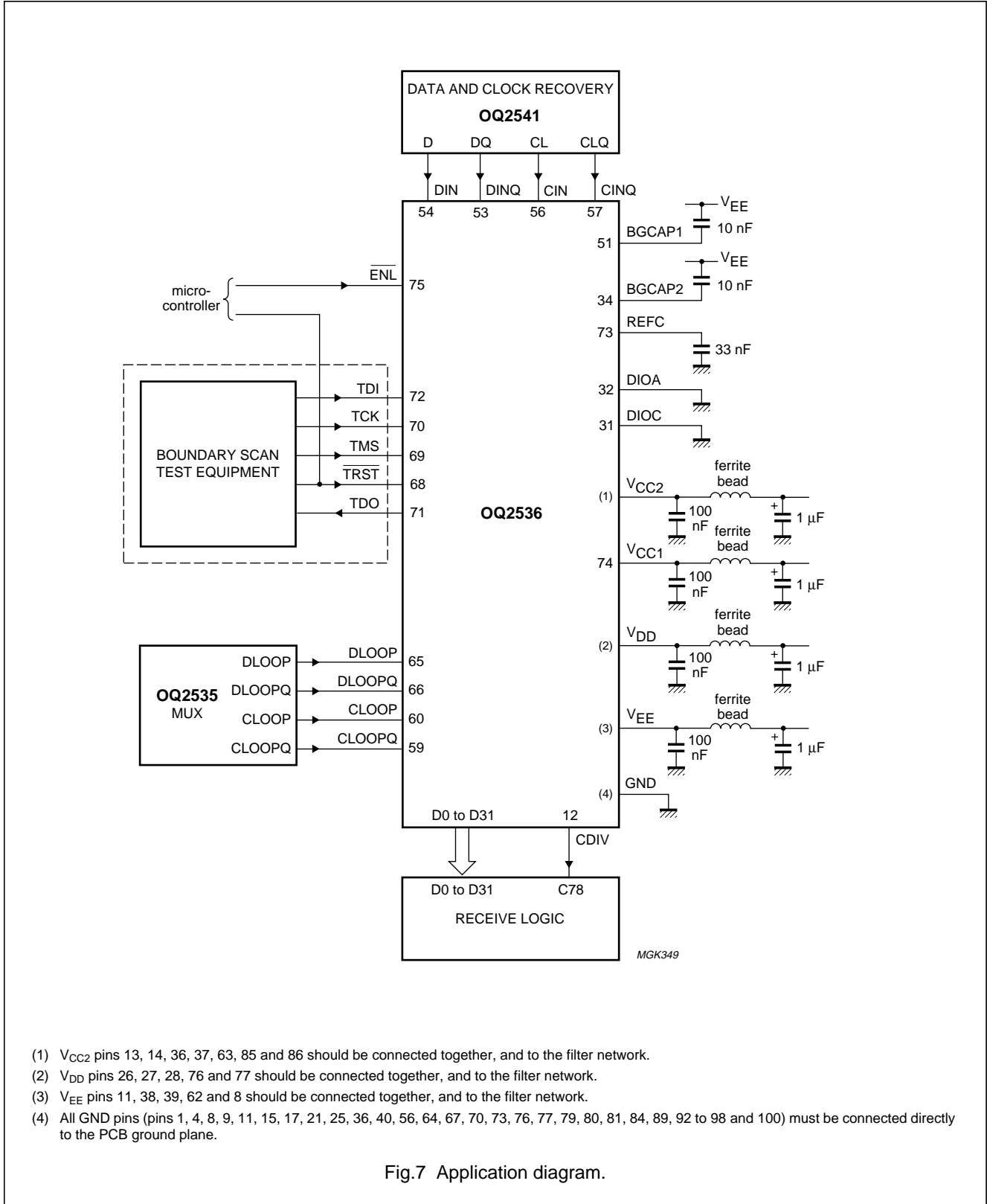
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APPLICATION INFORMATION



- (1) V_{CC2} pins 13, 14, 36, 37, 63, 85 and 86 should be connected together, and to the filter network.
- (2) V_{DD} pins 26, 27, 28, 76 and 77 should be connected together, and to the filter network.
- (3) V_{EE} pins 11, 38, 39, 62 and 8 should be connected together, and to the filter network.
- (4) All GND pins (pins 1, 4, 8, 9, 11, 15, 17, 21, 25, 36, 40, 56, 64, 67, 70, 73, 76, 77, 79, 80, 81, 84, 89, 92 to 98 and 100) must be connected directly to the PCB ground plane.

Fig.7 Application diagram.

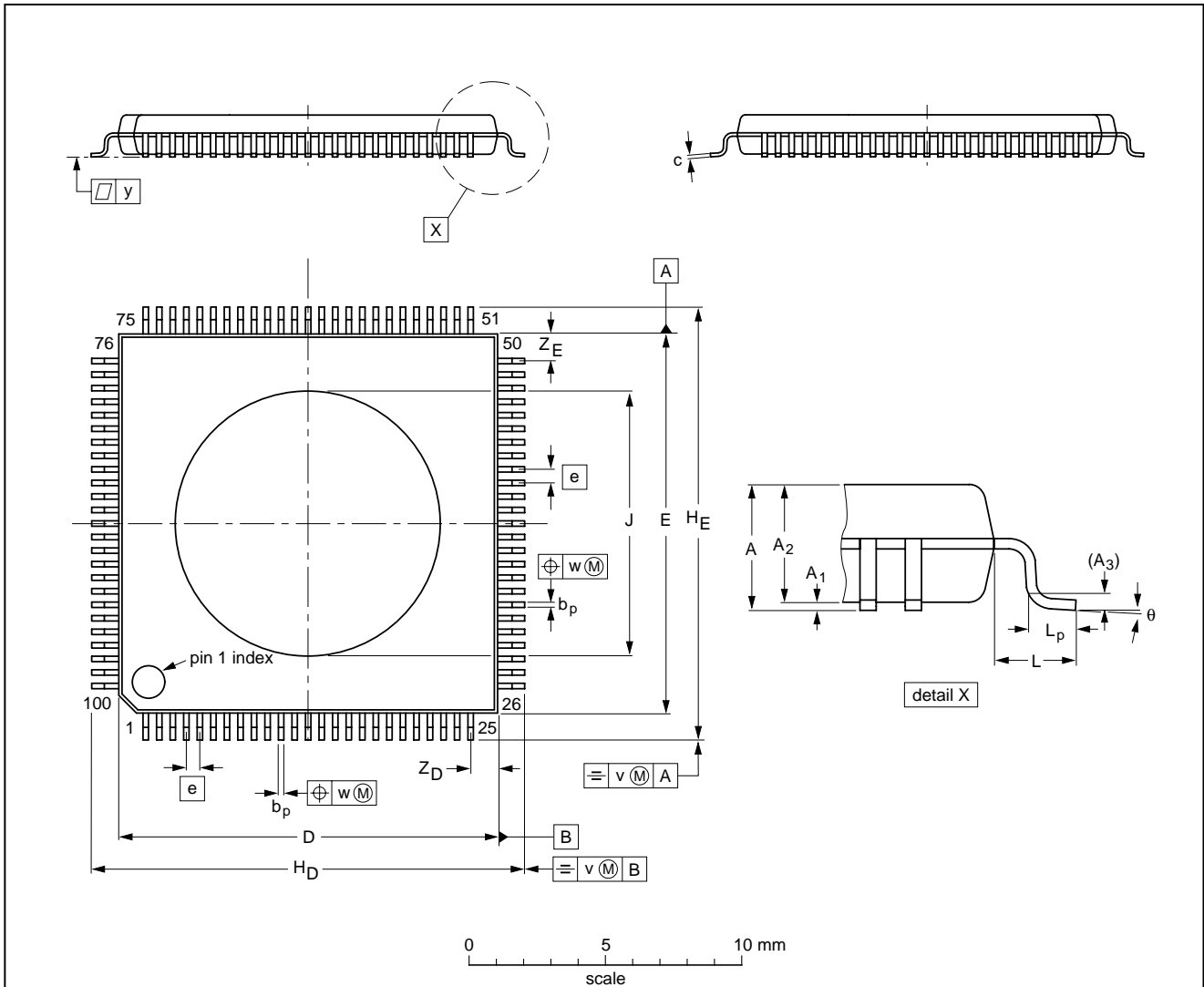
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PACKAGE OUTLINE

HLQFP100: plastic heat-dissipating low profile quad flat package;
100 leads; body 14 x 14 x 1.4 mm

SOT470-1



DIMENSIONS (mm are the original dimensions)

UNIT	A max.	A ₁	A ₂	A ₃	b _p	c	D ⁽¹⁾	E ⁽¹⁾	e	H _D	H _E	J ⁽²⁾	L	L _p	v	w	y	Z _D ⁽¹⁾	Z _E ⁽¹⁾	θ
mm	1.6	0.20 0.05	1.5 1.3	0.25	0.28 0.16	0.18 0.12	14.1 13.9	14.1 13.9	0.5	16.25 15.75	16.25 15.75	10.15 9.15	1.0	0.75 0.45	0.2	0.12	0.1	1.15 0.85	1.15 0.85	7° 0°

Notes

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.
2. Heatsink intrusion 0.0127 maximum.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT470-1						97-01-13

SDH/SONET STM16/OC48 demultiplexer**OQ2536HP**

SOLDERING**Introduction**

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Reflow soldering

Reflow soldering techniques are suitable for all LQFP packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

Wave soldering

Wave soldering is **not** recommended for LQFP packages. This is because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering cannot be avoided, the following conditions must be observed:

- **A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.**
- **The footprint must be at an angle of 45° to the board direction and must incorporate solder thieves downstream and at the side corners.**

Even with these conditions, do not consider wave soldering LQFP packages LQFP48 (SOT313-2), LQFP64 (SOT314-2) or LQFP80 (SOT315-1).

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

SDH/SONET STM16/OC48 demultiplexer

OQ2536HP

DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

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