## Typical Applications

- GSM/DCS Dual-Band Handsets
- Cellular/PCS Dual-Band Handsets
- General Purpose Amplification
- Commercial and Consumer Systems


## Product Description

The RF2363 is a dual-band Low Noise Amplifier designed for use as a front-end for 950 MHz GSM/ 1850 MHz DCS applications and may be used for dualband cellular/PCS applications. The 900 MHz LNA is a single-stage amplifier; the 1900 MHz LNA is a 2 -stage amplifier. The part may also be tuned for applications in other frequency bands. The device has an excellent combination of low noise figure and high linearity at a very low supply current. It is packaged in a very small industry standard SOT 8-lead plastic package.

Optimum Technology Matching ${ }^{\circledR}$ A pplied $\begin{array}{lll}\square \text { Si BJT } & \square \text { GaAs HBT } & \square \text { GaAs MESFET } \\ \square \text { Si Bi-CMOS } & \square \text { SiGe HBT } & \square \text { Si CMOS }\end{array}$


Functional Block Diagram


Package Style: SOT, 8-Lead

## Features

- Low Noise and High Intercept Point
- 18dB Gain at 900 MHz
- 21 dB Gain at 1900 MHz
- Low Supply Current
- Single 2.5V to 5.0V Power Supply
- Very Small SOT-23-8 Plastic Package


## Ordering Information

| RF2363 | Dual-Band 3V Low Noise Amplifier |
| :--- | :--- |
| RF2363 PCBA | Fully Assembled Evaluation Board |

Absolute Maximum Ratings

| Parameter | Rating | Unit |
| :--- | :---: | :---: |
| Supply Voltage | -0.5 to +6.0 | $\mathrm{~V}_{\mathrm{DC}}$ |
| Input RF Level | +10 | $\mathrm{dBm}^{\circ} \mathrm{C}$ |
| Operating Ambient Temperature | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |



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| Parameter | Specification |  |  | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |  |  |
| Overall <br> RF Frequency Range |  | $\begin{gathered} 800 \text { to } 1000 \\ 1800 \text { to } 2000 \end{gathered}$ |  | $\begin{aligned} & \mathrm{MHz} \\ & \mathrm{MHz} \end{aligned}$ |  |
| 950MHz Performance <br> Gain <br> Isolation <br> Gain Step <br> Noise Figure <br> Output IP3 <br> Input P1dB <br> Reverse Isolation <br> Input VSWR <br> Output VSWR | 16 +17 | $\begin{gathered} 18 \\ 16 \\ 34 \\ 1.3 \\ +24 \\ -10 \\ 20 \\ 1.8: 1 \\ 1.8: 1 \end{gathered}$ | 20 2:1 2:1 | dB dB dB dB dBm dBm dB | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=950 \mathrm{MHz}, \mathrm{~V}_{\mathrm{CC}}=2.8 \mathrm{~V}, \\ & \mathrm{EN} 1=2.8 \mathrm{~V}, \mathrm{EN} 2=0 \mathrm{~V} \\ & \\ & \mathrm{EN} 1=0 \mathrm{~V} \\ & \text { Gain - Isolation } \end{aligned}$ <br> No external matching With external match as per GSM/DCS Application Schematic |
| 1850 MHz Performance <br> Gain <br> Isolation <br> Gain Step <br> Noise Figure <br> Output IP3 <br> Input P1dB <br> Reverse Isolation <br> Input VSWR <br> Output VSWR | 20 +16 | $\begin{gathered} 21.5 \\ 10 \\ 31.5 \\ 1.4 \\ +22 \\ -12 \\ 30 \\ 1.7: 1 \\ 1.7: 1 \end{gathered}$ | 24 <br> 2:1 $2: 1$ | dB <br> dB <br> dB <br> dB <br> dBm <br> dBm <br> dB | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=1850 \mathrm{MHz}, \mathrm{~V}_{\mathrm{CC}}=2.8 \mathrm{~V}, \\ & \mathrm{EN} 2=2.8 \mathrm{~V}, \mathrm{EN} 1=0 \mathrm{~V} \\ & \mathrm{EN} 2=0 \mathrm{~V} \\ & \text { Gain - Isolation } \end{aligned}$ <br> No external matching With external match as per GSM/DCS Application Schematic |
| LNA Select "Enable" Voltage "Disable" Voltage |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}} \\ 0 \end{gathered}$ |  | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \end{aligned}$ |  |
| Power Supply <br> Voltage <br> Current Consumption |  | $\begin{gathered} 2.8 \\ 2.5 \text { to } 5.0 \\ 5 \\ 7.5 \end{gathered}$ | 1 | V <br> V <br> mA <br> mA <br> $\mu \mathrm{A}$ | $\mathrm{T}=25^{\circ} \mathrm{C}$ <br> Specifications <br> Operating limits <br> 900MHz LNA Enabled, 1900MHz LNA Disabled; total DC current <br> 1900MHz LNA Enabled, 900 MHz LNA Disabled; total DC current $\mathrm{EN} 1=\mathrm{EN} 2=0 \mathrm{~V}$ |

## Preliminary

RF2363

| Pin | Function | Description | Interface Schematic |
| :---: | :---: | :---: | :---: |
| 1 | RF OUT1 | RF output pin for $\sim 900 \mathrm{MHz}$ LNA. This pin is an open-collector output. It must be biased to either $\mathrm{V}_{\mathrm{CC}}$ or pin 4 through a choke or matching inductor. It is typically matched to $50 \Omega$ with a shunt bias/matching inductor and series blocking/matching capacitor. Refer to application schematics. |  |
| 2 | GND | Ground connection. <br> NOTE: Ground traces on pins 2 and 7 are equivalent to a small amount of inductance $(\sim 0.75 \mathrm{nH})$. The dimensions of these lines are as follows. Pin 2: $L=56$ mils, $W=15$ mils, $H=31$ mils Pin 7: $L=56$ mils, $W=15$ mils, $H=31$ mils Dielectric is FR-4. |  |
| 3 | RF OUT2 | RF output pin for $\sim 1900 \mathrm{MHz}$ LNA. This pin is an open-collector output. It must be biased to either $\mathrm{V}_{\mathrm{CC}}$ or pin 4 through a choke or matching inductor. It is typically matched to $50 \Omega$ with a shunt bias/matching inductor and series blocking/matching capacitor. Refer to application schematics. |  |
| 4 | EN1 | Enable pin for $\sim 900 \mathrm{MHz}$ LNA. A voltage equal to the supply voltage LNA. This pin should be disabled ( 0 V ) when the $\sim 1900 \mathrm{MHz} \mathrm{LNA}$ is in use. |  |
| 5 | EN2 | Enable pin for $\sim 1900 \mathrm{MHz}$ LNA. A voltage equal to the supply voltage LNA. This pin should be disabled ( 0 V ) when the $\sim 900 \mathrm{MHz}$ LNA is in use. See package drawing for description of pin orientation. | See pin 3. |
| 6 | RF IN2 | RF input pin for $\sim 1900 \mathrm{MHz}$. This pin is matched to approximately $50 \Omega$ at DCS/PCS frequencies. An external AC coupling capacitor is required at this pin. | See pin 3. |
| 7 | GND | Same as pin 2. | See pin 2. |
| 8 | RF IN1 | RF input pin for $\sim 900 \mathrm{MHz}$. This pin is matched to approximately $50 \Omega$ at GSM/Cellular frequencies. An external AC coupling capacitor is required at this pin. | See pin 1. |

## RF2363 Theory of Operation and Application Information

The RF2363 contains two independent low noise amplifiers which have been optimized for dual-band applications in the GSM ( 905 MHz to 960 MHz ) and DCS ( 1805 MHz to 1880 MHz ) frequency bands. Fabricated using heterojunction bipolar transistor (HBT) technology, the RF2363 delivers high linear gain at a very low noise figure and low power consumption. Internal temperature compensation keeps the gain tightly controlled over temperature extremes (typically less than 1 dB of gain variation from $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ at 2.8 V ). A $50 \Omega$ input impedance allows the part to be connected to standard receiver front end filters without additional matching components.

## MODE CONTROL

The RF2363 incorporates two enable pins (EN1 and EN2) for biasing the desired LNA according to the table below.

| EN1 | EN2 | Mode |
| :--- | :--- | :--- |
| GND | GND | Power Down |
| GND | VCC | 1900 MHz LNA On |
| VCC | GND | 900 MHz LNA On |

## 900MHz LNA

The 900 MHz LNA is a single-stage, common emitter amplifier. Since the input pin contains a DC bias, an AC coupling capacitor is required at this pin. An external bias inductor from the output pin (RF OUT1) to VCC provides DC biasing for the amplifier transistor and assists in matching the output impedance to the next receiver stage. A capacitor having a good RF bypass characteristic at the frequency of operation should be placed as close as possible to the supply voltage side of the bias inductor; a low frequency bypass capacitor should also be included. The EN1 pin supplies VCC to the bias circuits of the LNA and should also be effectively bypassed with both low and high frequency capacitors.

## 1900MHz LNA

The 1900 MHz LNA is implemented by two common emitter stages in cascade. The first stage is biased through an external inductor at the EN2 pin. This inductor also acts as an interstage match; a resistor in parallel with the inductor is recommended to 'de-Q' the inductor, thus providing a broader band interstage match. An external bias inductor from the output pin (RF OUT2) to VCC provides DC biasing for the second stage transistor and assists in matching the output impedance to the next receiver stage. Low and high frequency bypass capacitors should be used on the supply side of both the EN2 and RF OUT2 bias inductors. An AC coupling capacitor is required at the RF IN2 pin.

## LAYOUT CONSIDERATIONS

To provide optimal balance of gain and linearity, a small amount of inductance is required in the ground traces of the PCB. The recommended inductance is between 0.5 and 1.0 nH , with 0.75 nH used on the Evaluation Board. Depending on the application, more gain with less linearity or more linearity with less gain may be desired. Appropriate adjustment of the ground inductance can accomplish these objectives. Minimizing the ground inductance will maximize the gain at the expense of linearity while increasing the ground inductance will increase the linearity at the expense of gain. It is important to remember that the pin 7 ground inductance affects the performance of both LNAs, while the pin 2 ground inductance affects only the 1900 MHz LNA.

## Application Schematic (GSM/DCS)



## Evaluation Board Schematic

(Download Bill of Materials from www.rfmd.com.)


Pin 7: $\mathrm{L}=56$ mils, $W=15$ mils, $\mathrm{H}=31$ mils
Dielectric is FR-4

Evaluation Board Layout<br>Board Size 1.0" x 1.0"<br>Board Thickness 0.031", Board Material FR-4




Noise Figure versus Frequency 900 MHz LNA


Input 1 dB Compression Point versus Frequency



Noise Figure versus Frequency 1900 MHz LNA


Input 1 dB Compression Point versus Frequency



Current versus Supply Voltage


Output 3rd Order Intercept Point versus Frequency 1900 MHz LNA


Current versus Supply Voltage



1900MHz LNA


