Preliminary
RF2368

## Typical Applications

- DCS Handsets
- PCS Handsets
- General Purpose Amplification
- Commercial and Consumer Systems


## Product Description

The RF2368 is a DCS/PCS low noise amplifier with bypass switch designed for use as a front-end for DCS1800/PCS1900 applications. The LNA is a two-stage amplifier with bypass switch. This amplifier has low noise figure and high linearity in both high gain and bypass/low gain mode.

Optimum Technology Matching ${ }^{\circledR}$ Applied $\begin{array}{lll}\square \text { Si BJT } & \square \text { GaAs HBT } & \square \text { GaAs MESFET } \\ \square \text { Si Bi-CMOS } & \square \text { SiGe HBT } & \square \text { Si CMOS }\end{array}$


Functional Block Diagram


Package Style: SOT 8 Lead

## Features

- Low Noise and High Intercept Point
- Power Down Control
- Switchable Gain


## Ordering Information

| RF2368 | DCS/PCS 2.7V Low Noise Amplifier |
| :--- | :--- |
| RF2368 PCBA | Fully Assembled Evaluation Board |

## Absolute Maximum Ratings

| Parameter | Rating | Unit |
| :--- | :---: | :---: |
| Supply Voltage | -0.5 to +6.0 | $\mathrm{~V}_{\mathrm{DC}}$ |
| Input RF Level | +10 | dBm |
| Storage Temperature | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |



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| Parameter | Specification |  |  | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |  |  |
| Operating Range |  |  |  |  |  |
| Overall Frequency Range | 1800 |  | 2000 | MHz |  |
| Supply Voltage ( $\mathrm{V}_{\mathrm{CC}}$ ) | 2.7 | 2.78 | 2.86 | V | $\mathrm{V}_{\mathrm{CC} 1}, \mathrm{~V}_{\mathrm{CC} 2}$ |
| Power Down Voltage (V $\mathrm{V}_{\text {BIAS }}$ ) | 2.7 | 2.78 | 2.86 | V | BIAS |
| Logic Control Voltage Level | 0 |  | 2.86 | V | SELECT |
| Operating Ambient Temperature | -40 |  | +85 | ${ }^{\circ} \mathrm{C}$ |  |
| Input Impedance |  | 50 |  | $\Omega$ |  |
| Output Impedance |  | 50 |  | $\Omega$ |  |
| 1850MHz Performance High Gain Mode |  |  |  |  | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=1850 \mathrm{MHz}, \mathrm{~V}_{\mathrm{CC}}=\mathrm{BIAS}=2.78 \mathrm{~V}, \\ & \mathrm{SELECT}=0 \mathrm{~V}, \mathrm{Z}_{\mathrm{IN}}=\mathrm{Z}_{\mathrm{O}}=50 \Omega \end{aligned}$ |
| Gain | 17 | 18 | 19 | dB |  |
| Gain Variation Over Temperature Range |  |  | $\pm 0.5$ | dB |  |
| Gain Variation Over Frequency Band |  |  | $\pm 0.5$ | dB |  |
| Current Consumption |  | 9.0 | 9.5 | mA | $\mathrm{I}_{\mathrm{CC}}+\mathrm{I}_{\text {BIAS }}$ |
| Noise Figure |  | 1.6 | 1.7 | dB |  |
| Reverse Isolation | 15 | 20 |  | dB |  |
| Input IP3 | 0.0 | +1.0 |  | dBm |  |
| Input P1dB | -13 | -10 |  | dB |  |
| 1850MHz Performance Bypass Mode |  |  |  |  | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=1850 \mathrm{MHz}, \mathrm{VCC}=2.78 \mathrm{~V}, \\ & \mathrm{SELECT}=2.7 \mathrm{~V}, \mathrm{Z}_{\mathrm{IN}}=\mathrm{Z}_{\mathrm{O}}=50 \Omega \end{aligned}$ |
| Gain |  | -4.5 |  | dB |  |
| Gain Reduction | 21 | 22.5 | 24 | dBc |  |
| Power Down Current |  |  | 10 | $\mu \mathrm{A}$ |  |
| Input IP3 | 12 | 15.0 |  | dBm |  |
| Input P1dB | +5 | +8 |  | dB |  |
| 1960 MHz Performance High Gain Mode |  |  |  |  | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=1960 \mathrm{MHz}, \mathrm{~V}_{\mathrm{CC}}=\mathrm{BIAS}=2.78 \mathrm{~V}, \\ & \mathrm{SELECT}=0 \mathrm{~V}, \mathrm{Z}_{\mathrm{IN}}=\mathrm{Z}_{\mathrm{O}}=50 \Omega \end{aligned}$ |
| Gain | 15.5 | 16.5 | 17.5 | dB |  |
| Gain Variation Over Temperature Range |  |  | $\pm 0.5$ | dB |  |
| Gain Variation Over Frequency Band |  |  | $\pm 0.5$ | dB |  |
| Current Consumption |  | 9.0 | 9.5 | mA | $\mathrm{I}_{\mathrm{CC}}+\mathrm{I}_{\mathrm{BIAS}}$ |
| Noise Figure |  | 1.6 | 1.7 | dB |  |
| Reverse Isolation | 15 | 20 |  | dB |  |
| Input IP3 | +1 | +2 |  | dBm |  |
| Input P1dB | -13 | -10 |  | dB |  |
| 1960MHz Performance Bypass Mode |  |  |  |  | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{RF}=1960 \mathrm{MHz}, \mathrm{VCC}=2.78 \mathrm{~V}, \\ & \mathrm{SELECT}=2.7 \mathrm{~V}, \mathrm{Z}_{\mathrm{IN}}=\mathrm{Z}_{\mathrm{O}}=50 \Omega \end{aligned}$ |
| Gain |  | -5 |  | dB |  |
| Gain Reduction | 20 | 21.5 | 23 | dBc |  |
| Power Down Current |  |  | 10 | $\mu \mathrm{A}$ |  |
| Input IP3 | 14.0 | 17.0 |  | dBm |  |
| Input P1dB | +5 | +8 |  | dB |  |

## Preliminary

RF2368

| Pin | Function | Description | Interface Schematic |
| :---: | :---: | :---: | :---: |
| 1 | BIAS | BIAS is set to the supply voltage at high gain mode. For bypass mode see "Application Notes". | $\}_{-1}^{\text {Bias }}$ |
| 2 | GND |  |  |
| 3 | IN | DCS1800/PCS1900 RF input pin. |  |
| 4 | GND1 | LNA1 emittance inductance. Total inductance is comprised of package + bondwire + L2 on PCB. |  |
| 5 | VCC | Open collector for first stage LNA of DCS1800/PCS1900. It must be biased to $\mathrm{V}_{\mathrm{CC}}$ through a choke or matching inductor. | $\underbrace{\mathrm{vcc} 1}_{\mathrm{GND} 1}$ |
| 6 | GND2 | LNA2 emittance inductance. Total inductance is comprised of package+bondwire +L4 on PCB. |  |
| 7 | OUT | DCS1800 Amplifier Output pin. This pin is an open-collector output. It must be biased to $V_{C C}$ through a choke or matching inductor. This pin is typically matched to $50 \Omega$ with a shunt bias/matching inductor and series blocking/matching capacitor. Refer to application schematics. |  |
| 8 | SELECT | This pin selects high gain. Select $\leq 0.8 \mathrm{~V}$, high gain. Select $\geq 1.8 \mathrm{~V}$, low gain. |  |

## Application Notes

## Bypass Mode Configurations

The RF2368 may be placed into either high gain or bypass mode via the GAIN SELECT pin (pin 8). The high gain state is selected by asserting the GAIN SELECT pin to a voltage level of less than 0.8 V . For Bypass operation, there are two possible methods for placing the RF2368 into this low gain state. The table below shows the two possible Bypass configurations.

Bypass Mode Possibilities

| Gain Select | BIAS (V) | VCC1 and VCC2 (V) | Current (mA) |
| :---: | :---: | :---: | :---: |
| 2.7 | 0 | 2.78 | 1.4 |
| 2.7 | 2.7 | 2.78 | 2.2 |

For both Bypass configurations, the GAIN SELECT pin must be placed at a level greater than or equal to 1.8 V . The difference between the Bypass possibilities is determined by the specific application's ability to change the voltage of the BIAS pin (pin 1) independently of the $\mathrm{V}_{\mathrm{CC}}$ supply voltage. The advantage of the ability to assert the power down pin to 0 V when in Bypass mode is shown by the decreased current draw when in this Bypass configuration.

## BIAS Pin Resistor

The BIAS pin (pin1) of the RF2368 should be maintained at 2.7 V to 2.86 V for proper high gain operation. This voltage range ensures the correct bias current will be present at the BIAS pin of the device. However, an external series resistor may be used to allow various operating voltages at this pin (see R1 of the evaluation board schematic). The required value for this resistor may be roughly calculated by using the operating input voltage to the BIAS pin, the desired voltage at the device, and the typical current consumption for the BIAS pin, along with Ohm's law.

For example, assume the design will supply 5.0 V to the BIAS pin of the device, but the biasing circuitry internal to the RF2368 requires 2.78 V typical, and the BIAS current is known to typically be 0.25 mA , then the required value for R1 would be found as follows.

$$
\frac{5.0 \mathrm{~V}-2.78 \mathrm{~V}}{0.25 \mathrm{~mA}}=8.88 \mathrm{k} \Omega
$$

## Preliminary <br> RF2368

## Application Schematic



## Evaluation Board Schematic - PCS/DCS

(Download Bill of Materials from www.rfmd.com.)

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1
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$\qquad$

Evaluation Board Layout Board Size 1" x 1"

Board Thickness 0.032", Board Material FR-4


High Gain Mode Input Impedance (S11)


Bypass Mode Input Impedance (S11)


High Gain Mode Output Impedance (S22)


Bypass Mode Output Impedance (S22)


## S-Parameter Conditions:

All plots shown were taken at $\mathrm{VCC}=2.78 \mathrm{~V}$ and Ambient Temperature $=25^{\circ} \mathrm{C}$.

## Note:

All S11 and S22 plots shown were taken from an RF2368 while on a 2368310 evaluation board. The data was captured without the external input or output tuning components in place, and the reference points at the RF IN and RF OUT pins of the device.

