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# SA9102F/SA9102H

SINGLE PHASE BIDIRECTIONAL POWER/ENERGY METERING IC WITH INSTANTANEOUS PULSE OUTPUT

# FEATURES

- Performs bidirectional power and energy measurement
- Meets the IEC 521/1036 Specification requirements for Class 1 AC Watt hour meters
- Protected against ESD
- Power consumption rating typically 25mW

# DESCRIPTION

The SAMES SA9102F and SA9102H Single Phase bidirectional Power/Energy metering integrated circuits generate pulse rate outputs for positive and negative energy directions, the frequency of which is proportional to the power consumption. These devices perform the calculation for active power.

The method of calculation takes the power factor into account.

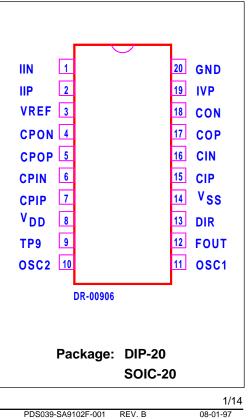
Energy consumption is determined by the power measurement being integrated over time.

These innovative universal power/energy metering integrated circuits are ideally suited for energy calculations in applications such as electricity dispensing systems (ED's), residential municipal metering and factory energy metering and control.

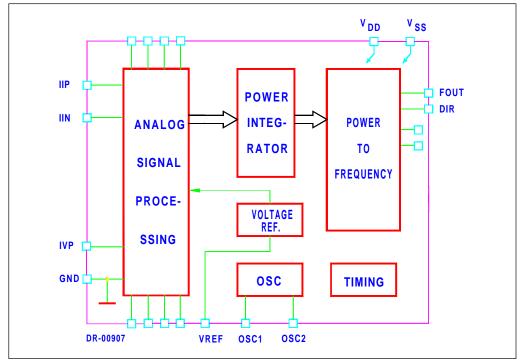
The SA9102F and SA9102H integrated circuits are available in both 20 pin dual-inline plastic (DIP-20), as well as 20 pin small outline (SOIC-20) package types.

- Adaptable to different types of current sensors
- Operates over a wide temperature range
- Precision voltage reference on-chip
- Two output signal formats available

## **PIN CONNECTIONS**



#### **BLOCK DIAGRAM**



#### **ABSOLUTE MAXIMUM RATINGS\***

Parameter	Symbol	Min	Max	Unit
Supply Voltage	V <sub>DD</sub> -V <sub>SS</sub>	-0.3	6.0	V
Current on any pin	I <sub>PIN</sub>	-150	+150	mA
Storage Temperature	T <sub>stg</sub>	-40	+125	°C
Operating Temperature	T <sub>o</sub>	-40	+85	С

\* Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only. Functional operation of the device at these or any other condition above those indicated in the operational sections of this specification, is not implied. Exposure to Absolute Maximum Ratings for extended periods may affect device reliability.

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# **ELECTRICAL CHARACTERISTICS**

 $(V_{DD} = 2.5V, V_{SS} = -2.5V)$ , over the temperature range -10°C to +70°C<sup>#</sup>, unless otherwise specified.)

Parameter	Symbol	Min	Тур	Max	Unit	Condition
Supply Voltage: Positive	V <sub>dd</sub>	2.25		2.75	V	
Supply Voltage: Negative	V <sub>ss</sub>	-2.75		-2.25	V	
Supply Current: Positive	I <sub>DD</sub>		5	6	mA	
Supply Current: Negative	I <sub>ss</sub>		5	6	mA	
Current Sensor Inputs (Dif	ferential)					
Input Current Range	I <sub>II</sub>	-25		+25	μA	Peak value
Voltage Sensor Input (Asymmetrical)						
Input Current Range	I <sub>IV</sub>	-25		+25	μA	Peak value
Pins FOUT, DIR Output Low Voltage Output High Voltage	V <sub>ol</sub> V <sub>oh</sub>	V <sub>DD</sub> -1		V <sub>ss</sub> +1	V V	I <sub>оL</sub> = 5mA I <sub>оH</sub> = -2mA
Pulse Rate FOUT	f <sub>P</sub>	0 0		64 180	Hz Hz	Specified linearity Min and max limits
Oscillator	Recommended crystal: TV colour burst crystal f = 3.5795 MHz					
Pin VREF Ref. Current Ref. Voltage	-I <sub>R</sub> V <sub>R</sub>	45 1.1	50	55 1.3	μA V	With R = $24k\Omega$ connected to V <sub>ss</sub> Referred to V <sub>ss</sub>

\* Extended Operating Temperature Range available on request.

#### **PIN DESCRIPTION**

Pin	Designation	Description
20	GND	Ground
8	V <sub>DD</sub>	Positive Supply Voltage
14	V <sub>ss</sub>	Negative Supply Voltage
19	IVP	Analog input for Voltage
1	IIN	Inputs for current sensor
2	IIP	-
11	OSC1	Connections for crystal or ceramic resonator
10	OSC2	(OSC1 = Input ; OSC2 = Output)
12	FOUT	Pulse rate output
13	DIR	Direction indication output
4	CPON	Connections for outer loop capacitor of
5	CPOP	A/D converter (Voltage)
6	CPIN	Connections for inner loop capacitor of
7	CPIP	A/D converter (Voltage)
15	CIP	Connections for inner loop capacitor of
16	CIN	A/D converter (Current)
17	COP	Connections for outer loop capacitor of
18	CON	A/D converter (Current)
3	VREF	Connection for current setting resistor
9	TP9	Test Pin. Connect to V <sub>ss</sub>

#### FUNCTIONAL DESCRIPTION

The SA9102F/SA9102H are CMOS mixed signal Analog/Digital integrated circuits, which perform bidirectional power/energy calculations across a power range of 1000:1, to an overall accurancy of better than Class 1.

These integrated circuits includes all the required functions such as two oversampling A/D converters for the voltage and current sense inputs, power calculation and energy integration. Internal offsets are eliminated through the use of cancellation procedures. These devices generate pulses, the frequency of which is proportional to the power consumption. The pulse rate follows the instataneous power measured. Direction information is also provided.

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#### 1. Power calculation

In the Application Circuit (Figure 1), the voltage drop across the shunt will be between 0 and 16mV (0 to 80A through a shunt resistor of  $200\mu\Omega$ ). This voltage is converted to a current of between 0 and 16µA, by means of resistors R<sub>1</sub> and R<sub>2</sub>.

The current sense input saturates at an input current of ±25µA peak.

For the voltage sensor input, the mains voltage (230VAC) is divided down through a divider to 14V. The current into the A/D converter input is set at 14 $\mu$ A at nominal mains voltage, via resistor R4 (1M $\Omega$ ).

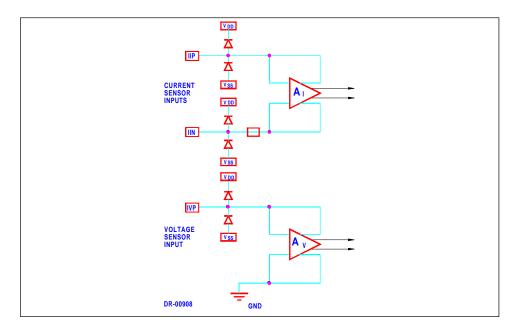
In this configuration, with a mains voltage of 230V and a current of 80A, the output frequency of the SA9102F and SA9102H power meter chips at FOUT (Pin 12) is 64Hz. In this case 1 pulse will correspond to an energy consumption of 18.4kW/ 64Hz = 287.5Ws.

#### 2. Analog Input configuration

The input circuitry of the current and voltage sensor inputs are illustrated below.

These inputs are protected against electrostatic discharge through clamping diodes.

The feedback loops from the outputs of the amplifiers  $A_1$  and  $A_2$  generate virtual shorts on the signal inputs. Exact duplications of the input currents are generated for the analog signal processing circuitry.



#### 3. Electrostatic Discharge (ESD) Protection

The SA9102F/SA9102H integrated circuits inputs/outputs are protected against ESD.

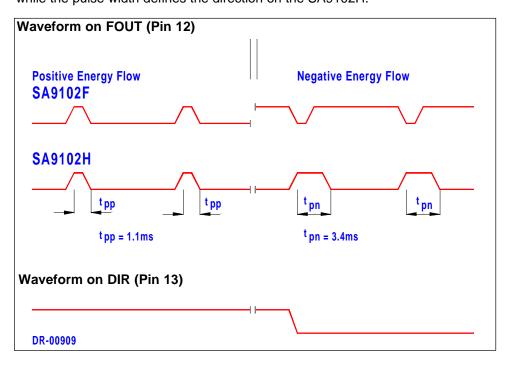
# 4. Power Consumption The power consumption rating of the SA9102F and SA9102H integrated circuits is

# less than 30mW.

#### 5. Pulse Output Signals

The calcualted power is divided down to a pulse rate of 64Hz, for rated conditions on FOUT (Pin 12), for both the SA9102F and SA9102H.

The format of the pulse output signal, which provides power/energy and direction information, is the only difference between the SA9102F and SA9102H devices. The direction of the energy flow is defined by the mark/space ratio in the SA9102F, while the pulse width defines the direction on the SA9102H.



An integrated anticreep function ensures no metering at zero line currents.

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The formula for calculating the output frequency (f) is given below:

 $f = 11.16 * FOUTX * \frac{FOSC}{3.58MHz} * \frac{I_1 \cdot I_V}{I_R^2}$ Where FOUTX = Nominal rated frequency (64Hz) FOSC = Oscillator frequency (2MHz ...... 4MHz)  $I_1$  = Input currents for current inputs (16µA at rated)  $I_V$  = Input currents for voltage inputs (14µA at rated)  $I_R$  = Reference current (typically 50µA)

XTAL is a colour burst TV crystal (f = 3.5795MHz) for the oscillator. The oscillator frequency is divided down to 1.7897MHz on-chip, to supply the digital circuitry and the A/d converters.

#### **TYPICAL APPLICATIONS**

In the Application Circuits (Figures 1 and 2), the components required for power metering applications are shown.

In Figure 1 a shunt resistor is used for current sensing. In this application, the circuitry requires a +2.5V, 0V, -2.5V DC supply.

In the case of Figure 2, when using a current transformer for current sensing, a +5V, 0V DC supply is sufficient.

The most important external components for the SA9102F and SA9102H integrated circuits are:

 $C_1$  and  $C_2$  are the outer loop capacitors for the two integrated oversampling A/D converters. The value of these capacitors is 560pF.

The actual values determine signal to noise and stability performance. The tolerances should be within  $\pm 10\%$ .

 $C_3$  and  $C_4$  are the inner loop capacitors of the A/D converters. The optimum value is 3.3nF. The actual values are uncritical. Values smaller than 0.5nF and larger than 5nF should be avoided.

 $R_2$ ,  $R_1$  and RSH are the resistors defining the current level into the current sense input. The values should be selected for an input current of 16µA into the SA9102F/SA9102H at maximum line current.

Values for RSH of less than  $200\mu\Omega$  should be avoided.

 $R_1 = R_2 = (I_L/16\mu A) * R_{SH}/2$ 

Where  $I_{L} = Line current$ 

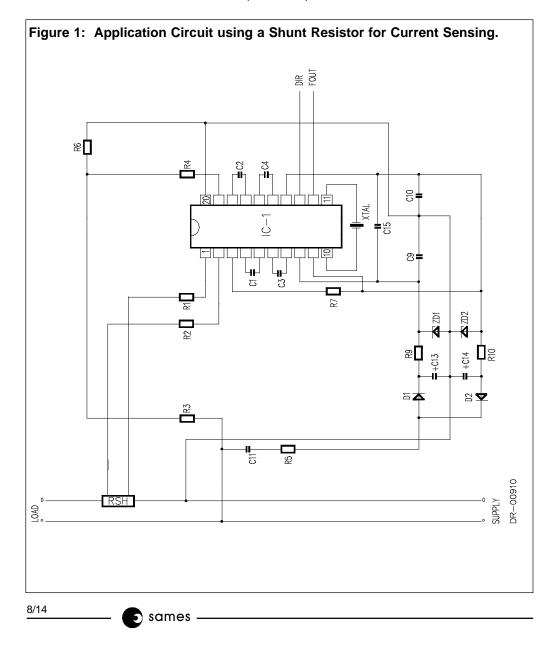
RSH = Shunt resistor/termination resistor

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 $R_3$ ,  $R_6$  and  $R_4$  set the current for the voltage sense input. The values should be selected so that the input current into the voltage sense input (virtual ground) is set to 14µA.

 $R_{\gamma}$  defines all on-chip bias and reference currents. With  $R_{\gamma} = 24k\Omega$ , optimum conditions are set.  $R_{\gamma}$  may be varied within ±10% for calibration purposes. Any change to  $R_{\gamma}$  will affect the output quadratically (i.e.:  $R_{\gamma} = +5\%$ ,  $f_{p} = +10\%$ ).



# Parts List for Application Circuit: Figure 1

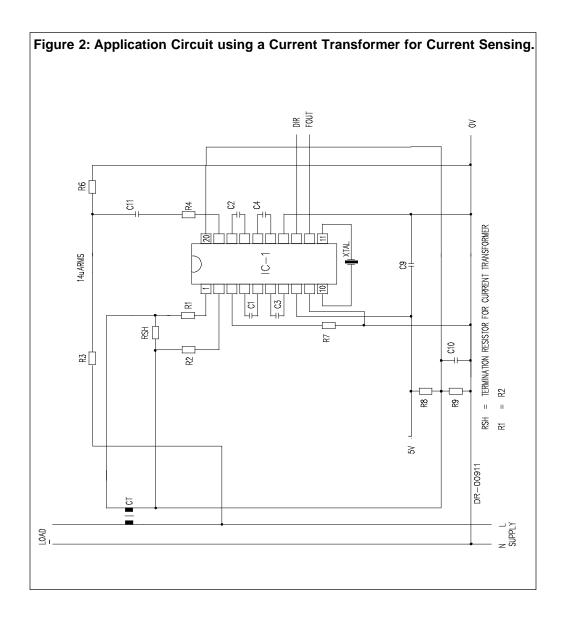
Item	Symbol	Description	Detail
1	IC-1	SA9102F/SA9102H	DIP-20/SOIC-20
2	D1	Diode, Silicon, 1N4148	
3	D2	Diode, Silicon, 1N4148	
4	ZD1	Diode, Zener, 2.4V, 200mW	
5	ZD2	Diode, Zener, 2.4V, 200mW	
6	XTAL	Crystal, 3.5795MHz	Colour burst TV
7	R1	Resistor, 1% metal	Note 1
8	R2	Resistor, 1% metal	Note 1
9	R3	Resistor, 390k, (230VAC) 1%, metal	
10	R4	Resistor, 1M, 1/4W, 1%, metal	
11	R5	Resistor, 470Ω, 2W, 1%, carbon	
12	R6	Resistor, 24k, 1/4W, 1%, metal	
13	R7	Resistor, 24k, 1/4W, 1%, metal	
14	R9	Resistor, 680Ω, 1/4W, 1%	
15	R10	Resistor, 680Ω, 1/4W, 1%	
16	C1	Capacitor, 560pF	
17	C2	Capacitor, 560pF	
18	C3	Capacitor, 3.3nF	
19	C4	Capacitor, 3.3nF	
20	C9	Capacitor, 100nF	
21	C10	Capacitor, 100nF	
22	C11	Capacitor, 0.47µF, 250VAC, polyester	
23	C13	Capacitor, 100µF	
24	C14	Capacitor, 100µF	
25	C15	Capacitor, 820nF	Note 2
26	RSH	Shunt Resistor	Note 3

Note 1: Resistor (R1 and R2) values are dependant upon the selected value of RSH.

Note 2: Capacitor (C15) to be positioned as close to Supply Pins (V\_{\_{DD}} \& V\_{\_{SS}}) of IC-1 as possible.

Note 3: See TYPICAL APPLICATIONS when selecting the value of RSH.

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Parts List for Application Circuit: Figure 2

Item	Symbol	Description	Detail
1	IC-1	SA9102F/SA9102H	DIP-20/SOIC-20
2	XTAL	Crystal, 3.5795MHz	Colour burst TV
3	RSH	Resistor	Note 1
4	R1	Resistor, 1%, metal	Note 2
5	R2	Resistor, 1%, metal	Note 2
6	R3	Resistor, 390k, (230VAC), 1%, metal	
7	R4	Resistor, 1M, 1/4W, 1%, metal	
8	R6	Resistor, 24k, 1/4W, metal	
9	R7	Resistor, 24k, 1/4W, 1%, metal	
10	R8	Resistor, 2.2k, 1/4W, 1%, metal	
11	R9	Resistor, 2.2k, 1/4W, 1%, metal	
12	C1	Capacitor, 560pF	
13	C2	Capacitor, 560pF	
14	C3	Capacitor, 3.3nF	
15	C4	Capacitor, 3.3nF	
16	C9	Capacitor, 820nF	Note 3
17	C10	Capacitor, 100nF	
18	C11	Capacitor	Note 4
19	СТ	Current Transformer	

Note 1: See TYPICAL APPLICATIONS when selecting the value of RSH.

Note 2: Resistor (R1and R2) values are dependent upon the selected value of RSH.

Note 3: Capacitor (C9) to be positioned as close to Supply Pins ( $V_{_{DD}} \& V_{_{SS}}$ ) of IC-1, as possible.

Note 4: Capacitor (C11) selected for DC blocking and to minimize phase error introduced by current transformer (typically 1.5μF).

# ORDERING INFORMATION

Part Number	Package
SA9102FPA	DIP-20
SA9102FSA	SOIC-20
SA9102HPA	DIP-20
SA9102HSA	SOIC-20

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