

N-CHANNEL 24V - 0.0052 Ω - 60A D²PAK STripFET™ III POWER MOSFET

TYPE	V _{DSS}	R _{DS(on)}	ID
STB100NH02L	24 V	< 0.006 Ω	60 A (2)

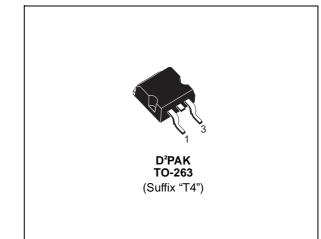
- TYPICAL R_{DS}(on) = 0.0052 Ω @ 10 V
- TYPICAL $R_{DS}(on) = 0.007 \ \Omega @ 5 V$
- R_{DS(ON)} * Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- SURFACE-MOUNTING D²PAK (TO-263)
 POWER PACKAGE IN TUBE (NO SUFFIX) OR
 IN TAPE & REEL (SUFFIX "T4")

DESCRIPTION

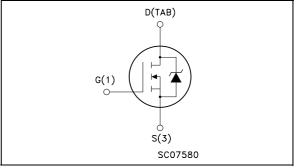
The STB100NH02L utilizes the latest advanced design rules of ST's proprietary STripFET[™] technology. This is suitable fot the most demanding DC-DC converter applications where high efficiency is to be achieved.

APPLICATIONS

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTERS



INTERNAL SCHEMATIC DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Symbol Parameter		Value	Unit
V _{spike(1)}	Drain-source Voltage Rating	30	V
V _{DS}	Drain-source Voltage (V _{GS} = 0)	24	V
Vdgr	Drain-gate Voltage (R_{GS} = 20 k Ω)	24	V
V_{GS}	Gate- source Voltage	± 20	V
I _D (2)	Drain Current (continuous) at T _C = 25°C	60	A
I _D (2)	Drain Current (continuous) at T _C = 100°C	60	A
I _{DM} (3)	Drain Current (pulsed)	240	A
P _{tot}	Total Dissipation at $T_C = 25^{\circ}C$	100	W
	Derating Factor	0.67	W/°C
E _{AS} (4)	Single Pulse Avalanche Energy	600	mJ
T _{stg} Storage Temperature		55 to 175	°C
T _i Max. Operating Junction Temperature		-55 to 175	

September 2003

THERMAL DATA

Rthj-case Rthj-amb	Thermal Resistance Junction-case Thermal Resistance Junction-ambient Maximum Lead Temperature For Soldering Purpose	Max Max	1.5 62.5 300	°C/W °C/W
11	Maximum Leau Temperature For Soldening Purpose		300	C

ELECTRICAL CHARACTERISTICS (T_{CASE} = 25 °C UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V _{(BR)DSS}	Drain-source Breakdown Voltage	I _D = 25 mA, V _{GS} = 0	24			V
I _{DSS}	Zero Gate Voltage Drain Current (V _{GS} = 0)	$V_{DS} = 20 V$ $V_{DS} = 20 V$ $T_{C} = 125^{\circ}C$			1 10	μΑ μΑ
I _{GSS}	Gate-body Leakage Current (V _{DS} = 0)	$V_{GS} = \pm 20 V$			±100	nA

ON (5)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
V _{GS(th)}	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I _D = 250 μA	1	1.8		V
R _{DS(on)}	Static Drain-source On Resistance				0.0052 0.007	0.006 0.011	Ω Ω

DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
g _{fs} ⁽⁵⁾	Forward Transconductance	V _{DS} = 10 V I _D = 30 A		40		S
C _{iss} C _{oss} C _{rss}	Input Capacitance Output Capacitance Reverse Transfer Capacitance	V_{DS} = 15V f = 1 MHz V_{GS} = 0		2850 800 120		pF pF pF
R _G	Gate Input Resistance	f=1 MHz Gate DC Bias=0 Test Signal Level =20 mV Open Drain		1		Ω



ELECTRICAL CHARACTERISTICS (continued)

SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{d(on)} t _r	Turn-on Delay Time Rise Time			13 75		ns ns
Q _g Q _{gs} Q _{gd}	Total Gate Charge Gate-Source Charge Gate-Drain Charge	V _{DD} =10 V I _D =60 A V _{GS} =10 V		47.5 10 7	64	nC nC nC

SWITCHING OFF

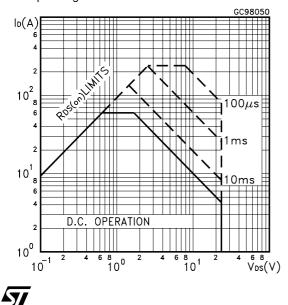
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{d(off)} t _f	Turn-off Delay Time Fall Time	$ \begin{array}{ll} V_{DD} = 10 \ V & I_D = 30 \ A \\ R_G = 4.7 \Omega, & V_{GS} = 10 \ V \\ (\text{Resistive Load, Figure 3}) \end{array} $		50 18	24.3	ns ns

SOURCE DRAIN DIODE

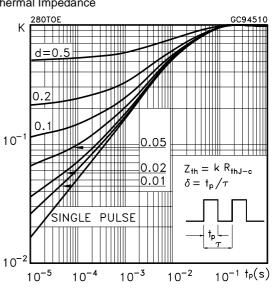
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{SD} I _{SDM}	Source-drain Current Source-drain Current (pulsed)				60 240	A A
V _{SD} ⁽⁵⁾	Forward On Voltage	$I_{SD} = 30 \text{ A}$ $V_{GS} = 0$			1.3	V
t _{rr} Q _{rr} I _{RRM}	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$ I_{SD} = 60 \mbox{ A} \qquad \mbox{di/dt} = 100 \mbox{A/} \mu \mbox{s} \\ V_{DD} = 16 \mbox{ V} \qquad \mbox{T}_j = 150^{\circ} \mbox{C} \\ (\mbox{see test circuit, Figure 5}) $		35 35 2		ns nC A

 $^{(5)}$ Pulsed: Pulse duration = 300 µs, duty cycle 1.5 %. $^{(6)}$ Q_{oss} = C_{oss}* Δ V_{in} , C_{oss} = C_{gd} + C_{ds} . See Appendix A $^{(7)}$ Gate charge for synchronous operation

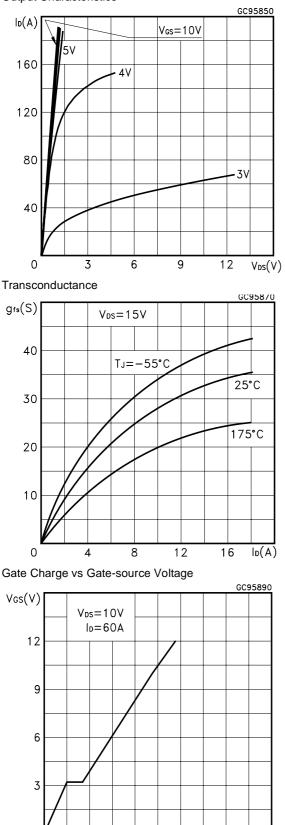
Safe Operating Area

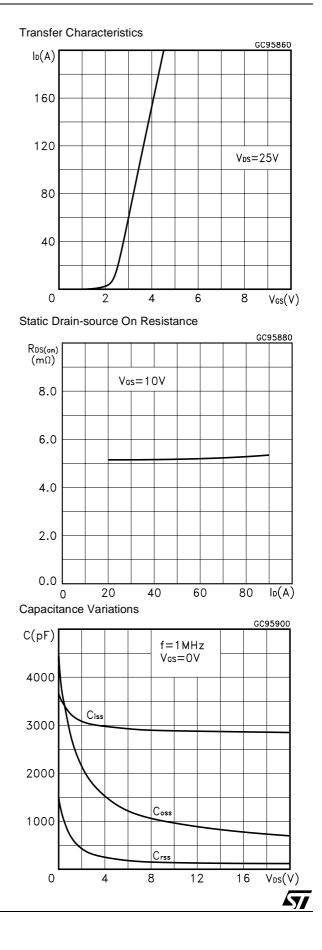


Thermal Impedance



Output Characteristics





0

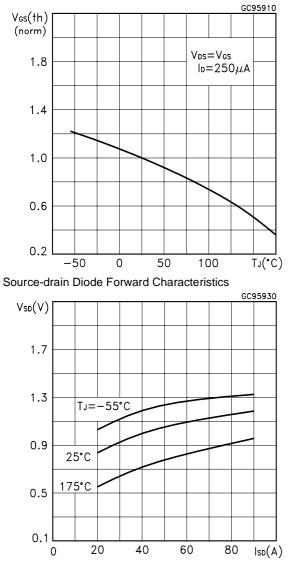
20

40

60

80

Q₀(nC)



Normalized Gate Threshold Voltage vs Temperature

Ros(on) (norm) 1.6 1.4 1.2 1.0 0.8 -50 0 50 100 TJ(°C)



Normalized Breakdown Voltage vs Temperature.

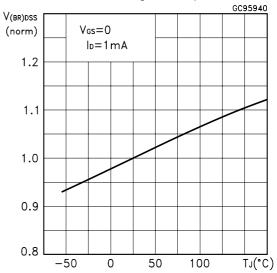


Fig. 1: Unclamped Inductive Load Test Circuit

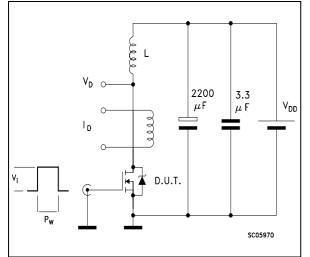


Fig. 3: Switching Times Test Circuits For Resistive Load

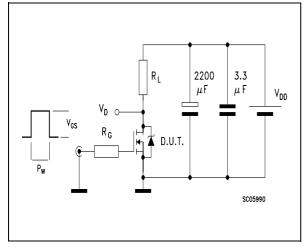


Fig. 5: Test Circuit For Inductive Load Switching And Diode Recovery Times

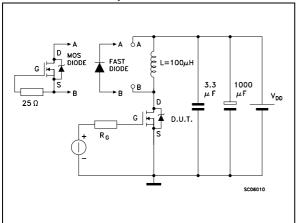


Fig. 2: Unclamped Inductive Waveform

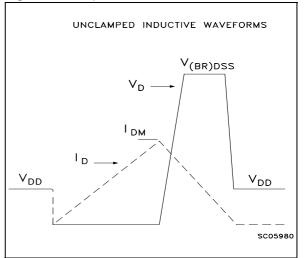
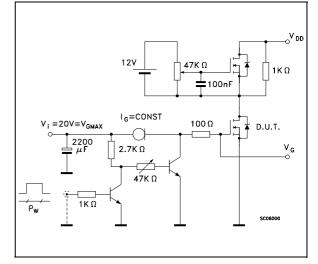
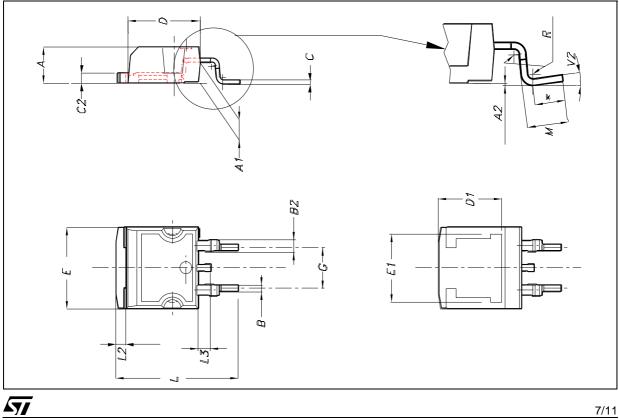


Fig. 4: Gate Charge test Circuit

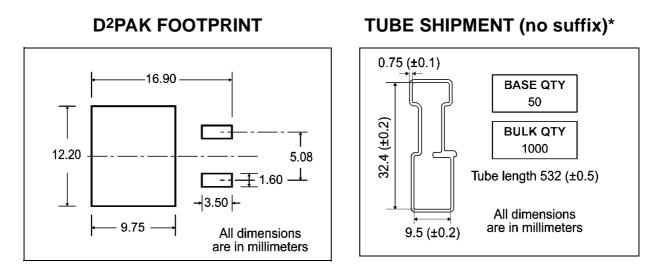


Γ

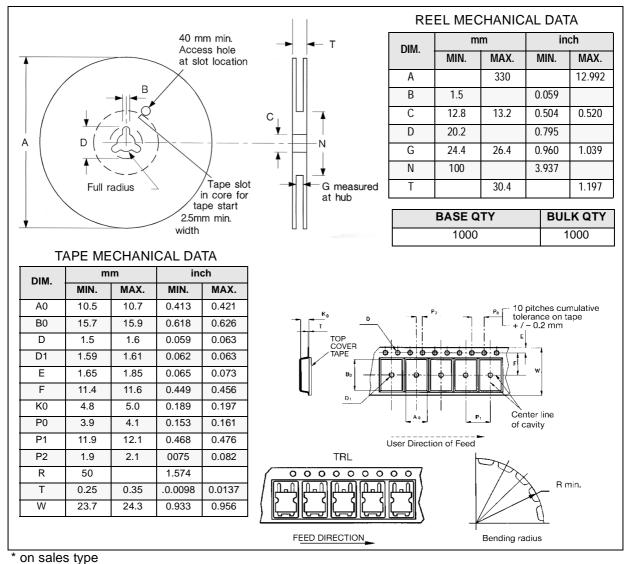
		D ² PA	K MECHANIC	AL DATA		
DIM.		mm.			inch.	
	MIN.	TYP.	MAX.	MIN.	TYP.	TYP.
Α	4.4		4.6	0.173		0.181
A1	2.49		2.69	0.098		0.106
A2	0.03		0.23	0.001		0.009
В	0.7		0.93	0.028		0.037
B2	1.14		1.7	0.045		0.067
С	0.45		0.6	0.018		0.024
C2	1.21		1.36	0.048		0.054
D	8.95		9.35	0.352		0.368
D1		8			0.315	
E	10		10.4	0.394		0.409
E1		8.5			0.334	
G	4.88		5.28	0.192		0.208
L	15		15.85	0.591		0.624
L2	1.27		1.4	0.050		0.055
L3	1.4		1.75	0.055		0.069
М	2.4		3.2	0.094		0.126
R		0.4		1	0.015	
V2	0°		4°	0°		4°



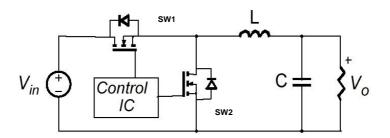
7/11



TAPE AND REEL SHIPMENT (suffix "T4")*



APPENDIX A Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is **e**moved to allow for a safer working junction temperature.

The low side (SW2) device requires:

- Very low R_{DS(on)} to reduce conduction losses
- Small Q_{gls} to reduce the gate charge losses
- Small C_{oss} to reduce losses due to output capacitance
- Small Q_{rr} to reduce losses on SW₁ during its turn-on
- The C_{gd}/C_{gs} ratio lower than V_{th}/V_{gg} ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- Small R_g and L_s to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Qg to have a faster commutation and to reduce gate charge losses
- Low R_{DS(on)} to reduce the conduction losses.

A7/

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduct	ion	$R_{\rm DS(on)SW1}*I_L^2*d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitching		$V_{in} * (Q_{gsth(SW1)} + Q_{gd(SW1)}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P _{diode}	Recovery	Not Applicable	$^{1}V_{in} * Q_{rr(SW2)} * f$
	Conduction	Not Applicable	$V_{\rm f(SW2)} * I_{\rm L} * t_{\rm deadtime} * f$
Pgate(Q _G)	$Q_{g(SW1)} * V_{gg} * f$	$\mathbf{Q}_{\mathbf{gls}(\mathrm{SW2})} * \mathbf{V}_{\mathbf{gg}} * \mathbf{f}$
P _{Qoss}		$\frac{V_{in} * Q_{oss(SW1)} * f}{2}$	$\frac{V_{in} * Q_{oss(SW2)} * f}{2}$

Parameter	Meaning
d	Duty-cycle
Qgsth	Post threshold gate charge
Q _{gls}	Third quadrant gate charge
Pconduction	On state losses
Pswitching	On-off transition losses
Pdiode	Conduction and reverse recovery diode losses
Pgate	Gate drive losses
PQoss	Output capacitance losses

¹ Dissipated by SW1 during turn-on

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