

# TSB21LV03C IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

- Fully Supports Provisions of IEEE 1394-1995 Standard for High Performance Serial Bus
- Fully Interoperable with FireWire™ and i.LINK™ Implementation of IEEE 1394-1995
- Provides Three Fully Compliant Cable Ports at 100/200 Megabits per Second (Mbits/s)
- Cable Ports Monitor Line Conditions for Active Connection to Remote Node
- Device Power-Down Feature to Conserve Energy in Battery-Powered Applications
- Inactive Ports Disabled to Save Power
- Logic Performs System Initialization and Arbitration Functions
- Encode and Decode Functions Included for Data-Strobe Bit-Level Encoding
- Incoming Data Resynchronized to Local Clock
- Single 3.3-V Supply Operation
- Interface to Link-Layer Controller Supports Low Cost TI™ Bus-Holder Isolation
- Data Interface to Link-Layer Controller Provided Through 2/4 Parallel Lines at 49.152 MHz
- Low Cost 24.576-MHz Crystal Oscillator and PLL Provide Transmit/Receive Data at 100/200 Mbits/s, and Link-Layer Controller Clock at 49.152 MHz
- Interoperable with 1394 Link-Layer Controllers Using 5-V Supplies
- Interoperable Across 1394 Cable with 1394 Physical Layers (Phy) Using 5-V Supplies
- Node Power-Class Information Signaling for System Power Management
- Cable Power Presence Monitoring
- Separate Cable Bias and Driver Termination Voltage Supply for Each Port
- High Performance 64-Pin TQFP (PM) Package and 68-Pin CFP (HV) Package

## description

The TSB21LV03C provides the analog and digital physical layer functions needed to implement a three-port node in a cable-based IEEE 1394-1995 network. Each cable port incorporates two differential line transceivers. The transceivers include circuitry to monitor the line conditions as needed for determining connection status, for initialization and arbitration, and for packet reception and transmission. The TSB21LV03C is designed to interface with a link-layer controller (LLC), such as the TSB12LV21, TSB12LV31, TSB12C01, TSB12LV22, TSB12LV41, or TSB12LV01.

The TSB21LV03C requires either an external 24.576-MHz crystal or crystal oscillator. The internal oscillator drives an internal phase-locked loop (PLL), which generates the required 196.608-MHz reference signal. The 196.608-MHz reference signal is internally divided to provide the 49.152/98.304-MHz clock signals that control transmission of the outbound encoded strobe and data information. The 49.152-MHz clock signal is also supplied to the associated LLC for synchronization of the two chips and is used for resynchronization of the received data. For the TSB21LV03C, the 49.152 MHz clock output is active when  $\overline{\text{RESET}}$  is asserted low. The power-down function, when enabled by taking the PD terminal high, stops operation of the PLL and disables all circuitry except the cable-not-active signal circuitry.

The TSB21LV03C supports an optional isolation barrier between itself and its LLC. When  $\overline{\text{ISO}}$  is tied high, the link interface outputs behave normally. Also, when  $\overline{\text{ISO}}$  is tied high, the internal bus hold function is enabled for use with the TI Bus Holder isolation. TI bus holder isolation is implemented when  $\overline{\text{ISO}}$  is tied high.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

FireWire is a trademark of Apple Computer, Incorporated.  
i.LINK is a trademark of SONY.  
TI is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1999, Texas Instruments Incorporated  
On products compliant to MIL-PRF-38535, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

---

### description (continued)

Data bits to be transmitted through the cable ports are received from the LLC on two or four data lines (D0 – D3), and are latched internally in the TSB21LV03C in synchronization with the 49.152-MHz system clock. These bits are combined serially, encoded, and transmitted at 98.304 or 196.608 Mb/s as the outbound data-strobe information stream. During transmission, the encoded data information is transmitted differentially on the TPB cable pair(s), and the encoded strobe information is transmitted differentially on the TPA cable pair(s).

During packet reception the TPA and TPB transmitters of the receiving cable port are disabled, and the receivers for that port are enabled. The encoded data information is received on the TPA cable pair, and the encoded Strobe information is received on the TPB cable pair. The received data-strobe information is decoded to recover the receive clock signal and the serial data bits. The serial data bits are split into two or four parallel streams, resynchronized to the local system clock, and sent to the associated LLC. The received data is also transmitted (repeated) out of the other active (connected) cable ports.

Both the TPA and TPB cable interfaces incorporate differential comparators to monitor the line states during initialization and arbitration. The outputs of these comparators are used by the internal logic to determine the arbitration status. The TPA channel monitors the incoming cable common-mode voltage. The value of this common mode voltage is used during arbitration to set the speed of the next packet transmission. In addition, the TPB channel monitors the incoming cable common-mode voltage for the presence of the remotely supplied twisted-pair bias voltage. The presence or absence of this common-mode voltage is used as an indication of cable connection status. The cable connection status signal is internally debounced in the TSB21LV03C on a cable disconnect-to-connect. The debounced cable connection status signal initiates a bus reset. On a cable disconnect-to-connect a debounce delay is incorporated. There is no delay on a cable disconnect.

The TSB21LV03C provides a 1.86-V nominal bias voltage for driver load termination. This bias voltage, when seen through a cable by a remote receiver, indicates the presence of an active connection. The value of this bias voltage has been chosen to allow interoperability between transceiver chips operating from either 5-V or 3-V nominal supplies. This bias voltage source should be stabilized by using an external filter capacitor of approximately 1.0  $\mu$ F.

The transmitter circuitry is disabled under the following conditions: power down, cable not active, reset, or transmitter disable. The receiver circuitry is disabled under the following conditions: power down, cable not active, or receiver disable. The twisted-pair bias voltage circuitry is disabled under the following conditions: power down or reset. The power-down condition occurs when the PD input is high. The cable-not-active (CNA) condition occurs when the cable connection status indicates that no cable is connected. The reset condition occurs when the  $\overline{\text{RESET}}$  input terminal is low. The transmitter disable and receiver disable conditions are determined from the internal logic.

The line drivers in the TSB21LV03C operate in a high-impedance current mode and are designed to work with external 110- $\Omega$  line-termination resistor networks. One network is provided at each end of each twisted-pair cable. Each network is composed of a pair of series-connected 55- $\Omega$  resistors. The midpoint of the pair of resistors that is directly connected to the twisted-pair A (TPA) package terminals is connected to the TPBIAS voltage terminal. The midpoint of the pair of resistors that is directly connected to the twisted-pair B (TPB) package terminals is coupled to ground through a parallel RC network with recommended resistor and capacitor values of 5 k $\Omega$  and 220 pF respectively. The values of the external resistors are designed to meet the draft standard specifications when connected in parallel with the internal receiver circuits and are shown in Figure 3.

The driver output current, along with other internal operating currents, is set by an external resistor. This resistor is connected between the R0 and R1 terminals and has a value of 6.3 k $\Omega$ ,  $\pm 0.5\%$ . This might be accomplished by placing a 6.34 k $\Omega$ ,  $\pm 0.5\%$  resistor in parallel with a 1-M $\Omega$  resistor.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**description (continued)**

Four package terminals are used as inputs to set four configuration status bits in the self-identification (Self-ID) packet. These terminals are hardwired high or low as a function of the equipment design. PC0 – PC2 are the three terminals that indicate either the need for power from the cable or the ability to supply power to the cable. The fourth terminal, C/LKON, indicates whether a node is a contender for bus manager. When the C/LKON terminal is asserted, it means the node can be a contender for bus manager. When the terminal is not asserted, it means that the node is not a contender. The C bit corresponds to bit 20 in the Self-ID packet, PC0 corresponds to bit 21, PC1 corresponds to bit 22, and PC2 corresponds to bit 23 (see Table 4–29 of the IEEE 1394–1995 standard for additional details).

A power-down terminal, PD, is provided to allow a power-down mode where most of the TSB21LV03C circuits are powered down to conserve energy in battery-powered applications. A cable status terminal, CNA, provides a high output when all twisted-pair cable ports are disconnected. This output is not debounced. The CNA output can be used to determine when to power the TSB21LV03C down or up. In the power-down mode all circuitry is disabled except the CNA circuitry. It should be noted that when the device is powered-down it does not act in a repeater mode. When the TSB21LV03C is powered down using the PD terminal, the twisted-pair transmitter and receiver circuitry has been designed to present a high impedance to the cable to prevent loading the TPBias terminal voltage on the other end of the cable.

**NOTE:**

Reference suspend/resume section in the current 1394a specification for interoperability with PD implementation of power down.

If the TSB21LV03C is being used with one or more of the ports not being brought out to a connector, the TPB terminals must be terminated for reliable operation. For each unused port, the TPB+ and TPB– terminals must be connected to GND. This is done in the normal termination network. When a port does not have a cable connected, the normal termination network pulls TPB+ and TPB– to ground through a 5-k $\Omega$  resistor, thus disabling the port.

**NOTE:**

All gap counts on all nodes of a 1394 bus must be identical. This may only be accomplished by using phy configuration packets (see section 4.3.4.3 of IEEE 1394-1995 Standard) or by using two bus resets, which resets the gap counts to the maximum level (3 Fh).

The link power status (LPS) terminal works with the C/LKON terminal to manage the LLC power usage of the node. The LPS terminal indicates that the LLC of the node is powered down and powers down the phy-LLC interface to save power. If the phy then receives a link-on packet, the C/LKON terminal is activated to output a 6.114 MHz signal, which can be used by the LLC to power itself up. Once the LLC is powered up, the LPS signal communicates this to the TSB21LV03C and the C/LKON signal is turned off and the phy-link interface is enabled.

Two of the package terminals are used to set up various test conditions used in manufacturing. These terminals, TESTM1 and TESTM2, should be connected to V<sub>DD</sub> for normal operation.

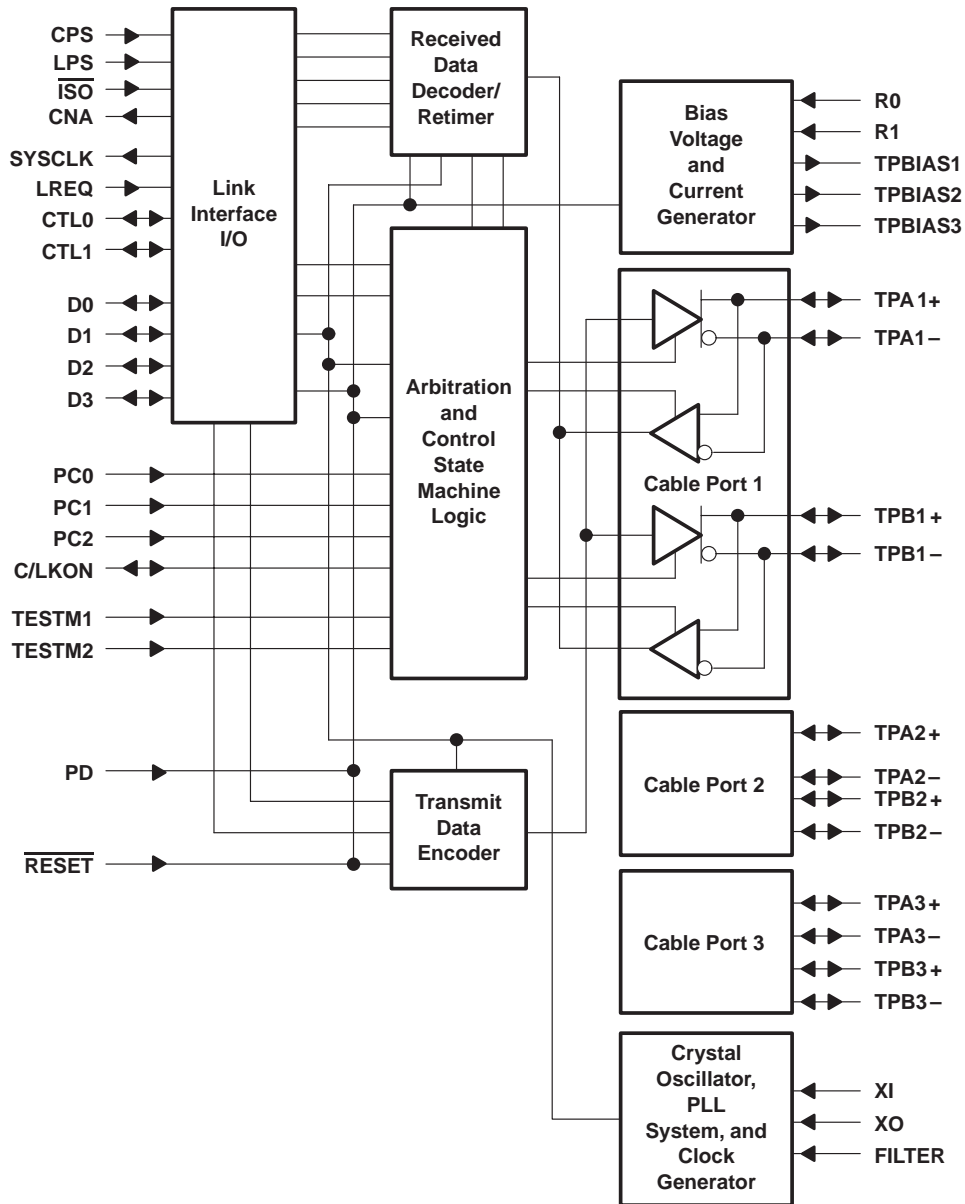
The TSB21LV03C is characterized for operation from 0°C to 70°C. The TSB21LV03CI is characterized for operation from –40°C to 85°C. The TSB21LV03CM is characterized for operation over the full military temperature range of –55°C to 125°C.

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### functional block diagram



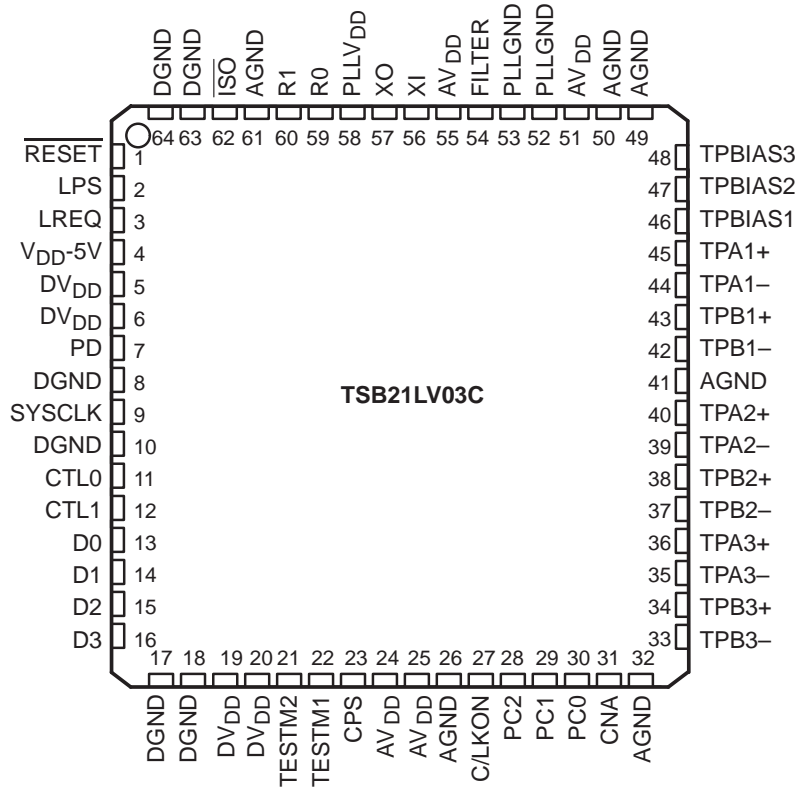
# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### package outline

**PLASTIC QUAD FLATPACK (PM)  
(TOP VIEW)**

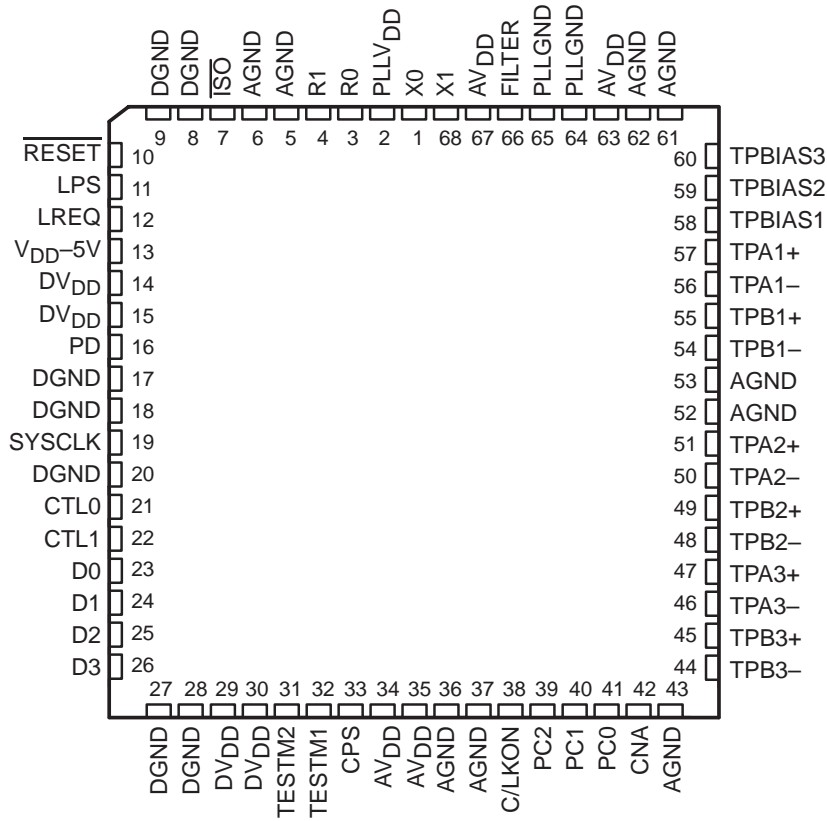


# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### CERAMIC QUAD FLATPACK (HV) (TOP VIEW)



### AVAILABLE OPTIONS

| T <sub>A</sub> | PACKAGE                     |                             |
|----------------|-----------------------------|-----------------------------|
|                | PLASTIC QUAD FLAT PACK (PM) | CERAMIC QUAD FLAT PACK (HV) |
| 0°C to 70°C    | TSB21LV03CPM                | —                           |
| -40°C to 85°C  | TSB21LV03CIPM               | —                           |
| -55°C to 125°C | —                           | TSB21LV03CMHVB              |



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### Terminal Functions

| TERMINAL     |  |                           | TYPE   | I/O | DESCRIPTION   |
|--------------|--|---------------------------|--------|-----|---|
| NAME         | NO.                                    |                           |        |     |   |
|              | HV                                     | PM                        |        |     |   |
| AGND         | 5, 6, 36,<br>37, 43, 52,<br>53, 61, 62 | 26, 32, 41,<br>49, 50, 61 | Supply | —   | Analog circuit ground. All AGND terminals should be tied together to the low-impedance circuit-board ground plane. External to the device, AGND should be tied to DGND and PLLGND.  |
| AVDD         | 34, 35,<br>63, 67                      | 24, 25,<br>51, 55         | Supply | —   | Analog circuit power. A combination of high frequency decoupling capacitors near each AVDD terminal is suggested, such as 0.1- $\mu$ F and 0.001- $\mu$ F capacitors. Lower frequency 10- $\mu$ F filtering capacitors are also recommended. AVDD terminals are separated from DVDD terminals internally from the other supply terminals to provide noise isolation. They should be tied together to a power plane on the circuit board. Each supply source should be individually filtered.  |
| C/LKON       | 38                                     | 27                        | CMOS   | I/O | Bus manager capable (input). When set as an input, C/LKON specifies in the Self-ID packet that the node is bus manager capable. The bit value programming is done by tying the terminal through a 10-k $\Omega$ resistor to VDD (high, bus manager capable) or to GND (low, not bus manager capable). Using either the pullup or pulldown resistor allows the LINK ON output to override the input bit value when necessary.<br><br>Link-on (output). When set as an output, C/LKON indicates the reception of a link-on message by asserting a 6.114-MHz signal. |
| CNA          | 42                                     | 31                        | CMOS   | O   | Cable-not-active output. CNA is asserted high when none of the TSB21LV03C ports are connected to another active port. This circuit remains active during the power-down mode.   |
| CPS          | 33                                     | 23                        | CMOS   | I   | Cable power status. CPS is normally connected to the cable power through a 400-k $\Omega$ resistor. This circuit drives an internal comparator that detects the presence of cable power. This information is maintained in two internal registers and is available to the LLC by way of a register read (see the Phy-Link Interface Annex in the IEEE 1394-1995 standard).  |
| CTL0<br>CTL1 | 21<br>22                               | 11<br>12                  | CMOS   | I/O | Control I/O. The CTLn terminals are bidirectional communications control signals between the TSB21LV03C and the LLC. These signals control the passage of information between the two devices. Control I/O terminals are 5-V tolerant. The CTLn terminals have an internal bus-holder function built-in.  |
| D0 – D3      | 23, 24,<br>25, 26                      | 13, 14,<br>15, 16         | CMOS   | I/O | Data I/O. The D terminals are bidirectional and pass data between the TSB21LV03C and the LLC. Data I/O terminals are 5-V tolerant. The D terminals have an internal bus-holder function built-in.   |
| DGND         | 8, 9, 17,<br>18, 20,<br>27, 28         | 8, 10, 17,<br>18, 63, 64  | Supply | —   | Digital circuit ground. The DGND terminals should be tied to the low-impedance circuit-board ground plane. External to the device, AGND should be tied to DGND and PLLGND.  |
| DVDD         | 14, 15,<br>29, 30                      | 5, 6,<br>19, 20           | Supply | —   | Digital circuit power. DVDD supplies power to the digital portion of the device. It is recommended that a combination of high-frequency decoupling capacitors be connected to DVDD (i.e., paralleled 0.1 $\mu$ F and 0.001 $\mu$ F). Lower frequency 10- $\mu$ F filtering capacitors can also be used. These supply terminals are separated from AVDD internally in the device to provide noise isolation. These terminals should also be tied together to a power plane on the circuit board. Individual filtering networks for each is desired.                |
| FILTER       | 66                                     | 54                        | CMOS   | I/O | PLL filter. FILTER is connected to a 0.1- $\mu$ F capacitor and then to PLLGND to complete the internal lag-lead filter. This filter is required for stable operation of the frequency multiplier PLL running off of the crystal oscillator.  |



# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### Terminal Functions (Continued)

| NAME                      | TERMINAL   |            | TYPE   | I/O | DESCRIPTION   |
|---------------------------|------------|------------|--------|-----|---|
|                           | NO.        |            |        |     |   |
|                           | HV         | PM         |        |     |   |
| $\overline{\text{ISO}}$   | 7          | 62         | CMOS   | I   | Link interface isolation input. $\overline{\text{ISO}}$ is normally tied high both to implement TI bus-holder isolation or no isolation. The TSB21LV03C does not support Annex J isolation.   |
| LPS                       | 11         | 2          | CMOS   | I   | Link power status. LPS is connected to either the $V_{DD}$ supplying the LLC through a 1-k $\Omega$ resistor or directly to a pulsed output that is active when the LLC is powered for the purpose of monitoring the LLC power status. The pulsed signal must be between 220 kHz and 5.5 MHz to be sensed as active. If LPS is inactive, the phy-LLC interface is disabled, and the TSB21LV03C performs only the basic repeater functions required for network initialization and operation. LPS is 5-V tolerant and has an internal bus-holder function built-in. If this terminal is tied through a resistor to a fixed state, the resistor must be 1 k $\Omega$ or less. |
| LREQ                      | 12         | 3          | CMOS   | I   | Link request. LREQ is an input from the LLC that requests the TSB21LV03C to perform some service. LREQ is 5-V tolerant and has an internal bus-holder function built-in. If this terminal is tied through a resistor to a fixed state, the resistor must be 1 k $\Omega$ or less.   |
| PC2 – PC0                 | 39, 40, 41 | 28, 29, 30 | CMOS   | I   | Power class indicators. The PC signals set the bit values of the three power-class bits in the Self-ID packet (bits 21, 22, and 23). These bits can be programmed by tying the terminals to $V_{DD}$ (high) or to GND (low).  |
| PD                        | 16         | 7          | CMOS   | I   | Power down. When asserted high, PD turns off all internal circuitry except the CNA monitor circuits that drive the CNA terminal. PD is 5-V tolerant. The PD terminal may be tied directly to $V_{DD}$ or to DGND. If this terminal is tied through a resistor to a fixed state, the resistor must be 1 k $\Omega$ or less. The PD terminal has an internal bus-holder function built in to it.  |
| PLLGND                    | 64, 65     | 52, 53     | Supply | —   | PLL circuit ground. The PLLGND terminals should be tied to the low-impedance circuit-board ground plane. External to the device, AGND should be tied to DGND and PLLGND.  |
| PLL $V_{DD}$              | 2          | 58         | Supply | —   | PLL circuit power. PLL $V_{DD}$ supplies power to the PLL portion of the device. It is recommended that a combination of high-frequency decoupling capacitors be connected to PLL $V_{DD}$ (i.e., paralleled 0.1 $\mu\text{F}$ and 0.001 $\mu\text{F}$ ). Lower frequency 10- $\mu\text{F}$ filtering capacitors can also be used. The PLL $V_{DD}$ supply terminals are separated from $AV_{DD}$ and $DV_{DD}$ internally in the device to provide noise isolation. The PLL $V_{DD}$ , $AV_{DD}$ , and $DV_{DD}$ terminals should also be tied together to a power plane on the circuit board. Individual filtering networks for each is recommended.                      |
| R0<br>R1                  | 3<br>4     | 59<br>60   | —      | —   | Current setting resistor. An internal reference voltage is applied to a resistor connected between R0 and R1 to set the operating current and the cable driver output current. A resistance of 6.3 k $\Omega$ $\pm$ 0.5% should be used to meet the IEEE 1394-1995 standard requirements for output voltage limits.   |
| $\overline{\text{RESET}}$ | 10         | 1          | CMOS   | I   | Reset. When $\overline{\text{RESET}}$ is asserted low (active), a bus reset condition is set on the active cable ports and the the internal logic is reset to the reset start state. An internal pullup resistor, which is connected to $V_{DD}$ , is provided so only an external delay capacitor is required. This input is a standard logic buffer and can also be driven by an open-drain logic output buffer. The minimum hold time for $\overline{\text{RESET}}$ is listed in the recommended operating characteristics table.  |
| SYSCLK                    | 19         | 9          | CMOS   | O   | System clock. SYSCLK provides a 49.152-MHz clock signal, which is synchronized with the data transfers to the LLC.  |
| TESTM1<br>TESTM2          | 32<br>31   | 22<br>21   | CMOS   | I   | Test mode control. TESTM1 and TESTM2 are used during the manufacturing test and should be tied to $V_{DD}$ .  |



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### Terminal Functions (Continued)

| TERMINAL            |         |          | TYPE   | I/O | DESCRIPTION   |
|---------------------|---------|----------|--------|-----|---|
| NAME                | NO.     |          |        |     |   |
|                     | HV      | PM       |        |     |   |
| TPA1+               | 57      | 45       | Cable  | O   | Portn, port cable pair A. TPA <sub>n</sub> is the port A connection to the twisted-pair cable. Board traces from these terminals should be kept matched and as short as possible to the external load resistors and to the cable connector.   |
| TPA2+               | 51      | 40       |        |     |   |
| TPA3+               | 47      | 36       |        |     |   |
| TPA1–               | 56      | 44       |        |     |   |
| TPA2–               | 50      | 39       |        |     |   |
| TPA3–               | 46      | 35       |        |     |   |
| TPB1+               | 55      | 43       | Cable  | O   | Portn, port cable pair B. TPB <sub>n</sub> is the port B connection to the twisted-pair cable. Board traces from these terminals should be kept matched and as short as possible to the external load resistors and to the cable connector.   |
| TPB2+               | 49      | 38       |        |     |   |
| TPB3+               | 45      | 34       |        |     |   |
| TPB1–               | 54      | 42       |        |     |   |
| TPB2–               | 48      | 37       |        |     |   |
| TPB3–               | 44      | 33       |        |     |   |
| TPBIAS1             | 58      | 46       | Cable  | O   | Portn, twisted-pair bias. TPBIAS <sub>n</sub> provides the 1.86-V nominal bias voltage needed for proper operation of the twisted-pair cable drivers and receivers and for sending a valid cable connection signal to the remote nodes.   |
| TPBIAS2             | 59      | 47       |        |     |   |
| TPBIAS3             | 60      | 48       |        |     |   |
| V <sub>DD</sub> –5V | 13      | 4        | Supply | —   | 5-V V <sub>DD</sub> supply. V <sub>DD</sub> –5V should be connected to the LLC V <sub>DD</sub> supply when a 5-V LLC is connected to the phy, and it should be connected to the phy DV <sub>DD</sub> when a 3-V LLC is used.  |
| XI<br>XO            | 68<br>1 | 56<br>57 | —      | —   | Crystal oscillator. XO and XI connect to a 24.576-MHz parallel resonant fundamental mode crystal. Although, when a 24.576-MHz crystal oscillator is used, it can be connected to XI with XO left unconnected. The optimum values for the external shunt capacitors are dependent on the specifications of the crystal used. See application note on crystal oscillator. |

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

|  |                                  |
|--|----------------------------------|
| Supply voltage range, V <sub>DD</sub> .....                        | –0.3 V to 4 V                    |
| Input voltage range, V <sub>I</sub> .....                          | –0.5 V to V <sub>DD</sub> +0.5 V |
| Output voltage range at any output, V <sub>O</sub> .....           | –0.5 V to V <sub>DD</sub> +0.5 V |
| Continuous total power dissipation .....                           | See Dissipation Rating Table     |
| Operating free-air temperature, T <sub>A</sub> , TSB21LV03C .....  | 0°C to 70°C                      |
| TSB21LV03CI .....  | –40°C to 85°C                    |
| TSB21LV03CM .....  | –55°C to 125°C                   |
| Junction temperature, T <sub>J</sub> , PM package .....            | 150°C                            |
| HV package .....   | 165°C                            |
| Storage temperature range, T <sub>stg</sub> .....                  | –65°C to 150°C                   |
| Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds ..... | 220°C                            |

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

**DISSIPATION RATING TABLE**

| PACKAGE | T <sub>A</sub> ≤ 25°C<br>POWER RATING | DERATING FACTOR‡<br>ABOVE T <sub>A</sub> = 25°C | T <sub>A</sub> = 70°C<br>POWER RATING | T <sub>A</sub> = 85°C<br>POWER RATING | T <sub>A</sub> = 125°C<br>POWER RATING |
|---------|---------------------------------------|---|---------------------------------------|---------------------------------------|--|
| PM      | 1866 mW                               | 14.9 mW/°C                                      | 1194 mW                               | 972 mW                                | —                                      |
| HV      | 2943 mW                               | 21.02 mW/°C                                     | 1997 mW                               | 1681 mW                               | 841 mW                                 |

‡ This is the inverse of the traditional junction-to-ambient thermal resistance (R<sub>θJA</sub>) and uses a board-mounted 67°C/W for PM package and 47.57°C/W for HV package.



# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### recommended operating conditions

|                                      |  | MIN                  | NOM           | MAX    | UNIT |
|--------------------------------------|--|----------------------|---------------|--------|------|
| Supply voltage, $V_{DD}$             | Source power node  | 3                    | 3.3           | 3.6    | V    |
|                                      | Nonsource power node <sup>§</sup>  | 2.7                  | 3             | 3.6    |      |
| High-level input voltage, $V_{IH}$   | CMOS inputs  | Commercial           | 0.7 $V_{DD}$  |        | V    |
|                                      |  | Industrial, Military | 0.85 $V_{DD}$ |        |      |
| Low-level input voltage, $V_{IL}$    | CMOS inputs  | 0.2 $V_{DD}$         |               | V      |      |
| Differential input voltage, $V_{ID}$ | Cable inputs, 100-Mbit operation   | 142                  | 260           |        | mV   |
|                                      | Cable inputs, 200-Mbit operation   | 132                  | 260           |        |      |
|                                      | Cable inputs, during arbitration   | 171                  | 262           |        |      |
| Common-mode input voltage, $V_{IC}$  | TPB cable inputs, 100-Mbit or speed signaling off, Source power node                 | 1.165                | 2.515         |        | V    |
|                                      | TPB cable inputs, 100-Mbit or speed signaling off, Nonsource power node <sup>§</sup> | 1.165                | 2.015         |        |      |
|                                      | TPB cable inputs, 200-Mbit speed signaling, Source power node                        | 0.935                | 2.515         |        |      |
|                                      | TPB cable inputs, 200-Mbit speed signaling, Nonsource power node <sup>§</sup>        | 0.935                | 2.015         |        |      |
| Receive input jitter                 | TPA, TPB cable inputs, 100-Mbit operation  |                      |               | *±1.08 | ns   |
|                                      | TPA, TPB cable inputs, 200-Mbit operation  |                      |               | *±0.5  |      |
| Receive input slew                   | Between TPA and TPB cable inputs, 100-Mbit operation                                 |                      |               | *±0.8  | ns   |
|                                      | Between TPA and TPB cable inputs, 200-Mbit operation                                 |                      |               | *±0.55 |      |
| Output current, $I_{OL}/I_{OH}$      | Control, Data, CNA and C/LKON outputs, SYSCLK  | -12                  | 12            |        | mA   |
| Output current, $I_O$                | TPBIAS outputs   | -3                   | 1.3           |        | mA   |
| Hold time, power-up reset (RESET)    |  | *2                   |               |        | ms   |

<sup>§</sup> For a node that does not source power (see Section 4.2.2.2 in IEEE 1394-1995 Standard).

\* These parameters are not production tested for the HV package.

### electrical characteristics over recommended operating conditions (unless otherwise noted)

#### driver

| PARAMETER   | TEST CONDITION                      | MIN                | TYP | MAX                | UNIT |
|---|-------------------------------------|--------------------|-----|--------------------|------|
| $V_{OD}$ Differential output voltage                        | 55- $\Omega$ load                   | 172                |     | 265                | mV   |
| $V_{(OFF)}$ Off-state differential output voltage           | Drivers disabled                    |                    |     | 20                 | mV   |
| $I_{O(diff)}$ Differential current (TPA+, TPA-, TPB+, TPB-) | Driver enabled, Speed signaling off | -1.05 <sup>†</sup> |     | 1.05 <sup>†</sup>  | mA   |
| $I_{(SP)}$ Common-mode speed signaling current (TPB+, TPB-) | 200-Mbit speed signaling enabled    | -2.53 <sup>‡</sup> |     | -4.84 <sup>‡</sup> | mA   |

<sup>†</sup> Limits are defined as the algebraic sum of TPA+ and TPA- driver currents. Limits also apply to TPB+ and TPB- as the algebraic sum of driver currents.

<sup>‡</sup> Limits are defined as the absolute limit of each of TPB+ and TPB- driver currents.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### electrical characteristics over recommended operating conditions (unless otherwise noted) (continued)

#### receiver

| PARAMETER |  | TEST CONDITION  | MIN | TYP | MAX | UNIT       |
|-----------|--|-----------------|-----|-----|-----|------------|
| $V_{IT}$  | Input threshold voltage                                      |                 | -30 |     | 30  | mV         |
| $V_{IT}$  | Cable bias-detect input threshold voltage, TPBn cable inputs | Driver disabled | 0.6 |     | 1   | V          |
| $I_{IC}$  | Common-mode input current                                    | Driver disabled | -40 |     | 40  | $\mu$ A    |
| $Z_{ID}$  | Differential input impedance                                 | Driver disabled | 15  |     |     | k $\Omega$ |
|           |  |                 |     |     | *6  | pF         |
| $Z_{IC}$  | Common-mode impedance  | Driver disabled | 20  |     |     | k $\Omega$ |
|           |  |                 |     |     | *24 | pF         |

\* These parameters are not production tested for the HV package.

#### device

| PARAMETER |   | TEST CONDITIONS                 |                                | MIN             | TYP | MAX     | UNIT    |
|-----------|---|---------------------------------|--------------------------------|-----------------|-----|---------|---------|
| $I_{DD}$  | Supply current  | $V_{DD} = 3.3$ V                | Node transmitting or repeating |                 | 114 |         | mA      |
|           |   |                                 | Node receiving                 |                 | 140 |         | mA      |
|           |   | $V_{DD} = 3.6$ V                |                                |                 |     | *175    | mA      |
|           |   | $V_{DD} = 3.6$ V                | Power-down mode                |                 | 20  |         |         |
| $V_{IT}$  | Power status input threshold voltage (CPS)                    | $R_L = 400$ k $\Omega$          |                                | 4.7             |     | 7.5     | V       |
| $V_{OH}$  | High-level output voltage                                     | $V_{DD} = \text{min}$ ,         | $I_{OH}$ max                   | $V_{DD} - 0.55$ |     |         | V       |
| $V_{OL}$  | Low-level output voltage                                      | $V_{DD} = \text{max}$ ,         | $I_{OL}$ min                   |                 |     | 0.5     | V       |
| $I_I$     | Input current (TESTM1, TESTM2, PC0, PC1, PC2)                 | $V_I = V_{DD}$ or 0             |                                |                 |     | $\pm 1$ | $\mu$ A |
| $I_{off}$ | Off-state output current (CTL0, CTL1, D0, D1, D2, D3, C/LKON) | $V_O = V_{DD}$ or 0             |                                |                 |     | $\pm 5$ | $\mu$ A |
|           | Pullup current (RESET)  | $V_I = 0$ or 1.5 V              | Commercial                     | -90             | -45 | -10     | $\mu$ A |
|           | Industrial, Military  |                                 | -110                           | -45             | -10 |         |         |
| $V_{TH+}$ | Positive arbitration comparator-input threshold voltage       |                                 |                                | 89              |     | 168     | mV      |
| $V_{TH-}$ | Negative arbitration comparator-input threshold voltage       |                                 |                                | -168            |     | -89     | mV      |
| $V_{IT}$  | Speed-signal input threshold voltage                          | TPBIAS –TPA common-mode voltage |                                | 49              |     | 131     | mV      |
| $V_O$     | Output voltage (TPBIAS1, TPBIAS2, TPBIAS3)                    | At rated $I_O$ current          |                                | 1.665           |     | 2.015   | V       |
|           | Bus holding current (LPS, LREQ, CTLn, Dn, PD)                 | $V_I = 1/2(V_{DD})$             |                                |                 | 725 |         | $\mu$ A |

\* These parameters are not production tested for the HV package.

#### thermal characteristics

| PARAMETER       |   | TEST CONDITIONS           | MIN | TYP   | MAX | UNIT           |
|-----------------|---|---------------------------|-----|-------|-----|----------------|
| $R_{\theta JA}$ | Junction-to-free-air thermal resistance | EIA/JESD51-3, No air flow |     | 67    |     | $^{\circ}$ C/W |
|                 |   |                           |     | 47.57 |     |                |
| $R_{\theta JC}$ | Junction-to-case thermal resistance     |                           |     | 10.4  |     | $^{\circ}$ C/W |
|                 |   |                           |     | 3     |     |                |



# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### switching characteristics

| PARAMETER   | MEASURED            | TEST CONDITIONS                        | MIN | TYP | MAX    | UNIT |
|---|---------------------|--|-----|-----|--------|------|
| Jitter, transmit  | TPA, TPB            |  |     |     | ±0.25  | ns   |
| Skew rate, transmit   | Between TPA and TPB |  |     |     | *±0.15 | ns   |
| $t_r$ Rise time, transmit   | 10% to 90%          | $R_L = 55 \Omega, C_L = 10 \text{ pF}$ |     |     | *2.2   | ns   |
| $t_f$ Fall time, transmit   | 90% to 10%          | $R_L = 55 \Omega, C_L = 10 \text{ pF}$ |     |     | *2.2   | ns   |
| $t_{su}$ Setup time, Dn, CTLn, LREQ $\uparrow\downarrow$ to SYSCLK $\uparrow$ | 50% to 50%          | See Figure 1                           |     | *5  |        | ns   |
| $t_h$ Hold time, Dn, CTLn, LREQ $\uparrow\downarrow$ before SYSCLK $\uparrow$ | 50% to 50%          | See Figure 1                           |     | *2  |        | ns   |
| $t_d$ Delay time, SYSCLK $\uparrow$ to Dn, CTLn $\uparrow\downarrow$          | 50% to 50%          | See Figure 2                           | 2   |     | 11     | ns   |

\* These parameters are not production tested for the HV package.

### PARAMETER MEASUREMENT INFORMATION

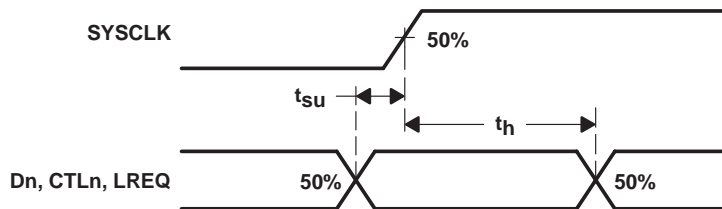


Figure 1. Dn, CTLn, LREQ Input Setup and Hold Timing Waveforms

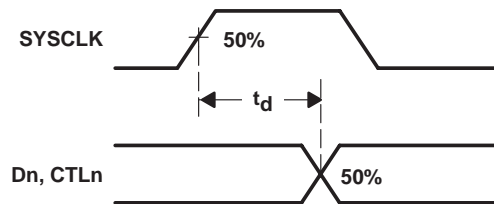


Figure 2. Dn and CTLn Output-Delay Timing Waveforms

## APPLICATION INFORMATION

### internal register configuration

The accessible internal registers of this device are listed in Table 1.

**Table 1. Internal Register Configuration**

| Address | 0           | 1      | 2      | 3  | 4        | 5    | 6        | 7   |
|---------|-------------|--------|--------|----|----------|------|----------|-----|
| 0000    | Physical ID |        |        |    |          |      | R        | CPS |
| 0001    | RHB         | IBR    | GC     |    |          |      |          |     |
| 0010    | SPD         |        | Rev    |    | NP       |      |          |     |
| 0011    | AStat1      |        | BStat1 |    | Ch1      | Con1 | Reserved |     |
| 0100    | AStat2      |        | BStat2 |    | Ch2      | Con2 | Reserved |     |
| 0101    | AStat3      |        | BStat3 |    | Ch3      | Con3 | Reserved |     |
| 0110    | LoopInt     | CPSInt | CPS    | IR | Reserved |      |          | C   |
| 0111    | Reserved    |        |        |    |          |      |          |     |
| 1000    | Reserved    |        |        |    |          |      |          |     |

**Table 2. Internal Register Field Descriptions**

| FIELD    | SIZE | TYPE      | DESCRIPTION  |
|----------|------|-----------|--|
| AStat(n) | 2    | Read only | AStat contains the line state of TPA <sub>n</sub> . The status is indicated by the following:<br>11 = high-impedance state<br>01 = 1<br>10 = 0<br>00 = Invalid data state. Power-up reset initializes to this line state. This line state is also output during transmit and receive operations, including data-end signaling. The line state outputs are generally valid during arbitration and idle conditions on the bus.   |
| BStat(n) | 2    | Read only | BStat contains the line state of TPB <sub>n</sub> . The status is indicated by the following:<br>11 = high-impedance state<br>01 = 1<br>10 = 0<br>00 = Invalid data state. Power-up reset initializes to this line state. This line state is also output during transmit and receive operations. The line state outputs are generally valid during arbitration and idle conditions on the bus.   |
| C        | 1    | R         | Bus manager capable. C indicates the state of the Bus Manager Capable input. When set, this bit is used by the TSB21LV03C to specify in the Self-ID packet that the node is Bus Manager Capable.   |
| Ch(n)    | 1    | Read only | When Ch = 1, the port is a child, otherwise it is a parent. This bit is invalid after a hardware reset or a bus reset until tree-ID processing is completed.   |
| Con(n)   | 1    | Read only | Con indicates the connection status of the port. When Con = 1, the port is connected, otherwise it is disconnected. This bit is set to 1 by a hardware reset and is updated to reflect the actual cable connection status of the port during bus reset. The TSB21LV03C contains connection debounce circuitry that prevents a new cable connection on a port from initiating a bus reset until the connection status has been stable for at least 335 ms. A cable disconnect initiates a bus reset immediately. After a hardware reset, the TSB21LV03C sets the connection status of all ports to 0. The TSB21LV03C proceeds with the bus reset, tree-ID, and Self-ID, but with all ports considered to be disconnected child ports. The TSB21LV03C can not transmit any signals on the serial bus ports during this time. The TSB21LV03C does report itself as root with a physical address of 00h at the completion of Self-ID. If any port is actually connected, after the debounce delay, the TSB21LV03C initiates another bus reset, which proceeds normally with interaction between the TSB21LV03C and its peer nodes. |

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### APPLICATION INFORMATION

**Table 2. Internal Register Field Descriptions (continued)**

| FIELD       | SIZE | TYPE       | DESCRIPTION   |
|-------------|------|------------|---|
| CPS         | 1    | Read only  | Cable power status (CPS) contains the status of the CPS input terminal. When cable power voltage has dropped too low for reliable operation, CPS is reset (0). CPS is included twice in the internal registers to expedite handling of the CPSInt.  |
| CPSInt      | 1    | Read/Write | CPSInt indicates that a cable power status interrupt has occurred. This interrupt occurs whenever the CPS input goes low. The interrupt indicates that the cable power voltage has dropped too low to ensure reliable operation. This bit is cleared (0) by a hardware reset or by writing a 0 to this register. However, if the CPS input is still low, another cable-power status interrupt immediately occurs.   |
| GC          | 6    | Read/Write | The gap count (GC) register sets the fair and arb-reset gap times. The gap count may be set to a particular value to optimize bus performance. Typically, the gap count should be set to 2 times the maximum number of hops on the bus and <b>must</b> be set to the same value for all nodes on the bus. The gap count can be set by either a write to this register or by reception or transmission of a PHY_CONFIG packet. The gap count is reset to 3Fh after a hardware reset or after two consecutive bus resets without an intervening write to the gap count register (either a write to the gap count register by the LLC or a PHY_CONFIG packet). |
| IBR         | 1    | Read/Write | When set, initiate bus reset (IBR) causes the current node to immediately initiate a bus reset. IBR is cleared (0) after a hardware reset or a bus reset.   |
| IR          | 1    | Read/Write | IR indicates that the last bus reset was initiated in this TSB21LV03C phy. This bit is also included in the self-ID packet.   |
| LoopInt     | 1    | Read/Write | LoopInt indicates that a configuration loop timeout has occurred. This interrupt occurs when the arbitration controller waits for too long a period of time during tree-ID. This interrupt can indicate that the bus is configured in a loop. This bit is cleared (0) by a hardware reset or by writing a 0 to this register bit.   |
| NP          | 4    | Read only  | NP contains the number of ports implemented in the core logic (not the number of ports actually on the device). For the TSB21LV03C, NP is set to 0011b.   |
| Physical ID | 6    | Read only  | Physical ID contains the physical address of the local node. The physical ID is valid after a hardware reset or a bus reset until the Self-ID process has been completed. A complete Self-ID is indicated by an unsolicited status transfer of the register 0 contents to the LLC.  |
| R           | 1    | Read only  | R indicates whether the current node is the root node or not. This bit is cleared (0) on a hardware reset or a bus reset. This bit is set during tree-ID when the current node is root.   |
| Rev         | 2    | Read only  | The revision (Rev) bits indicate the design revision of the core logic. For the TSB21LV03C, Rev is set to 00.   |
| RHB         | 1    | Read/Write | When set, the root hold-off bit (RHB) instructs the local node to try to become the root node during the next bus reset. RHB is reset (0) during a hardware reset and is not affected by a bus reset.   |
| SPD         | 2    | Read only  | The speed (SPD) bits indicates the top signaling speed of the local port and for the TSB21LV03C is set to 01b.  |



APPLICATION INFORMATION

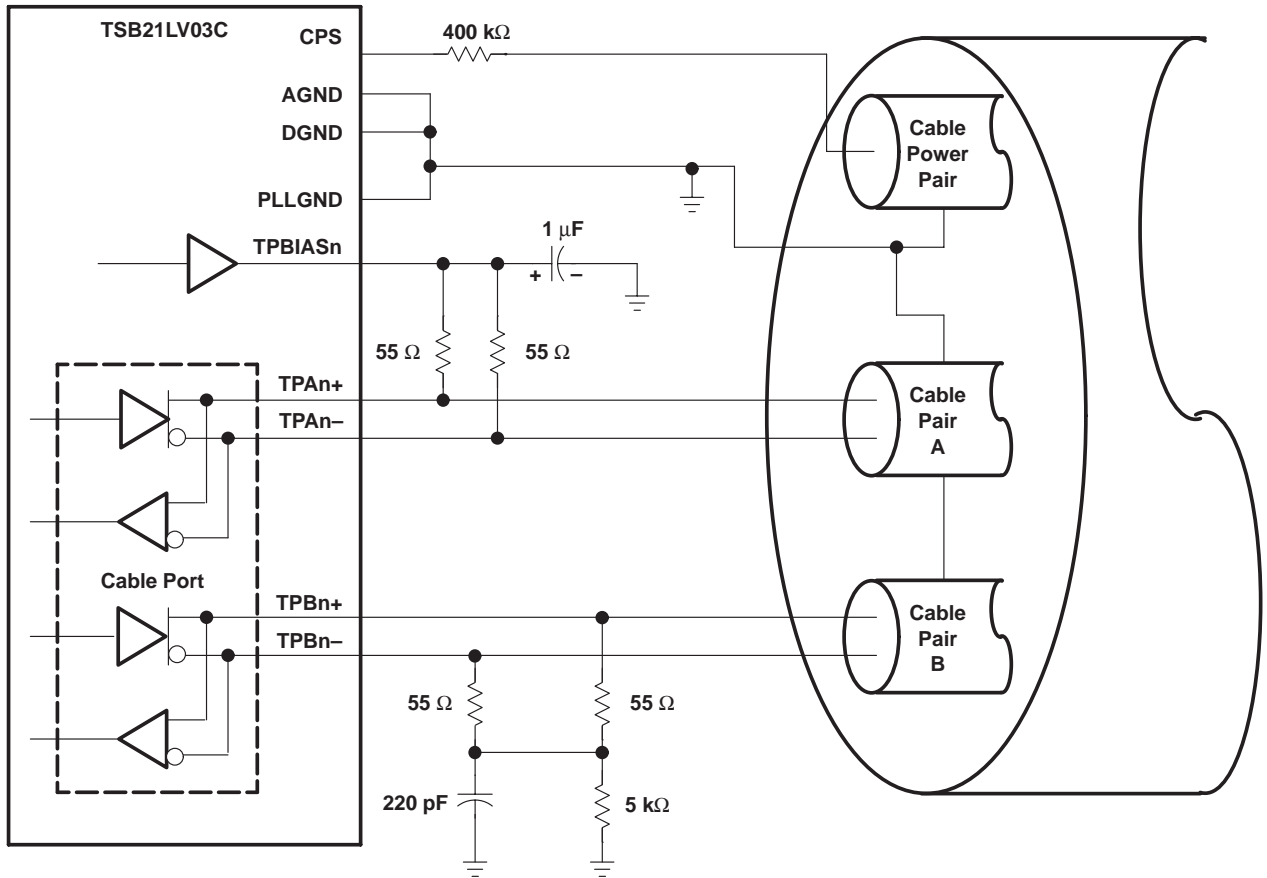


Figure 3. Twisted-Pair Cable Interface Connections

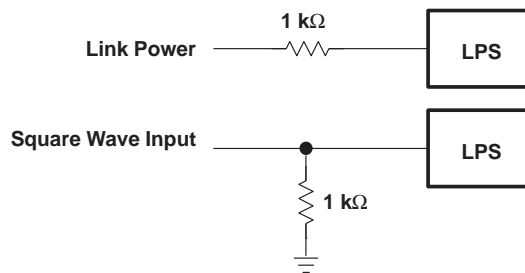


Figure 4. Nonisolated Connection Variations for LPS

APPLICATION INFORMATION

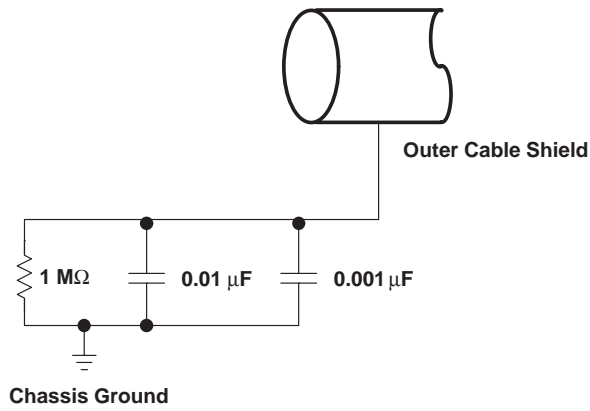


Figure 5. Compliant DC Isolated Outer Shield Termination

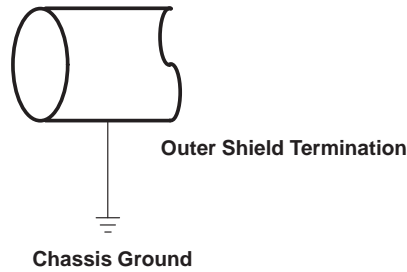


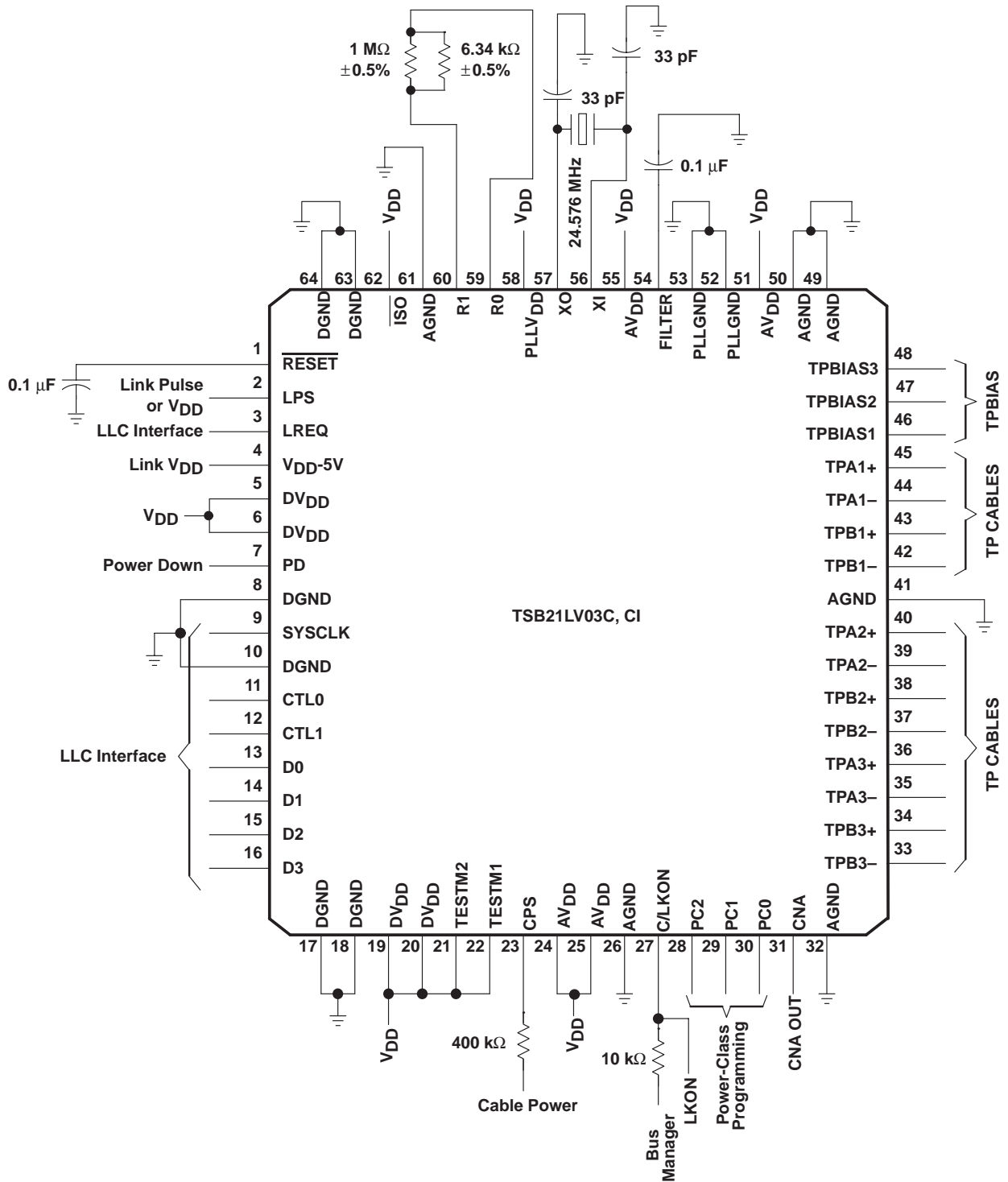
Figure 6. Nonisolated Outer Shield Termination



# TSB21LV03C IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

## APPLICATION INFORMATION (PM PACKAGE)



NOTE A: For more information see the application note.

**Figure 7. External Component Connections**





---

## APPLICATION INFORMATION

### crystal selection

TI PHYs may use an external 24.576 MHz crystal connected between the XI and XO pins on the PHY to provide the PHY clock. The following are some typical specifications for the crystals used with the Physical Layers from TI. The clock resulting from the input from the crystal must be within the tolerance of  $\pm 100$  parts per million for the PHYs to function correctly. This is required by the 1394 standard. This frequency tolerance for the PHY clocks on each node must be maintained over the variation introduced over production runs of boards and environment the machines operate in. Every board must have an SYSCLK (clock generated by the PHY) within  $\pm 100$  ppm of 49.152 MHz to be compliant to the 1394 standard. If adjacent nodes are more than 200 ppm away from one another then long packets sent across the 1394 bus may be corrupted, with the final bits of the packet being lost. TI PHYs are designed with a maximum of margin, but the limits imposed by 1394 must still be adhered to.

1. Crystal Mode of operation:  
Fundamental
2. Frequency Tolerance at 25°C:  
Total variation specification for the complete circuit is 100ppm. The crystal is specified at less than 100 ppm.
3. Frequency stability (over temperature):  
Total variation specification for the complete circuit is 100 ppm. The crystal is specified at less than 100 ppm.

**NOTE:**

The total variation must be kept below 100 ppm with some allowance for variation introduced by variations in board builds and device tolerances. So the sum of the frequency tolerance and the frequency stability must be less than 100 ppm. This can be traded off between the two, for example the frequency tolerance may be specified at 50 ppm and the temperature may be specified at 30 ppm to give a total of 80 ppm possible variation just due to the crystal.

4. Load capacitance: [Parallel (pF)]  
Parallel mode crystal circuits should be used for optimum precision. Load capacitance will be a function of your board layout and circuit. The total load capacitance ( $C_L$ ) will affect the frequency of oscillation. Consult with the crystal vendor on design to get an SYSCLK supplied by the PHY to less than 100 ppm from 49.152 MHz. A tolerance of  $\pm 5\%$  is recommend for load capacitors. For TI's TSBKOHCI403 Designer Kit with a crystal specified for 20-pF loading, a value of 33 pF for each load capacitor ( $C_9 = C_{10}$  below) is appropriate with the layout used for the board. The load specified for the crystal includes the load capacitors ( $C_9, C_{10}$ ), the loading of the PHY pins ( $C_{PHY}$ ), and the loading of the board itself ( $C_{BD}$ ). To summarize:  $C_L = [(C_9 \times C_{10}) / (C_9 + C_{10})] + C_{PHY} + C_{BD}$ . Representative values for  $C_{PHY}$  are  $\sim 1$  pF and for  $C_{BD}$  are about 0.8 pF per centimeter of board etch, a typical board can have from 3 pF to 6 pF or more. The capacitance of load capacitors  $C_9$  and  $C_{10}$  combine as capacitors in series.

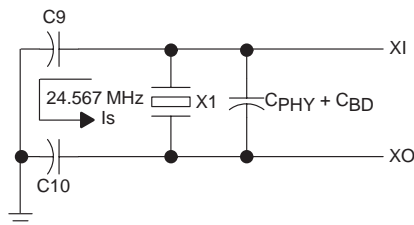
# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### APPLICATION INFORMATION

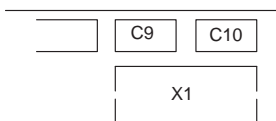
#### crystal selection (continued)



**Figure 9. Load Capacitance for the TSB21LV03C PHY**

**NOTE:**

The layout of the crystal portion of the PHY circuit is important for getting the correct frequency from the crystal, minimizing the noise introduced into the PHY Phase Lock Loop, and minimizing any emissions from the circuit. The crystal and the two load capacitors should be considered a unit during layout. The crystal and the load capacitors should be placed as close as possible to one another while minimizing the loop area created by the combination of the three components. Varying the size of the capacitors may help in this. Minimizing the loop area minimizes the effect of the resonant current ( $I_s$ ) that flows in this resonant circuit. This layout unit (crystal and load capacitors) should then be placed as close as possible to the PHY XI and XO pins to minimize etch lengths.



**Figure 10. Recommended Crystal and Capacitor Layout for the TSB21LV03C PHY**

Part of the verification process for the design should be to measuring the frequency of the SYSCLK output of the PHY. This should be done with a frequency counter with an accuracy of 6 digits or better. If the SYSCLK is more than the crystal tolerance away from 49.152 MHz, the load capacitance of the crystal may be varied to reduce total variation to below 100 ppm. Changes should be done to both load capacitors (C9 and C10 above) at the same time to the same value. Consult crystal vender for detailed understanding of requirements. In order for a 1394 bus to operate correctly each SYSCLK on each node on the bus must be within 200 ppm of the adjacent SYSCLK on the bus. The 1394 standard requires this by specifying a center frequency of 49.152 MHz and a  $\pm 100$  ppm tolerance around 49.152 MHz.

### PRINCIPLES OF OPERATION

The TSB21LV03C is designed to operate with a LLC such as the TI TSB12LV22, TSB12LV41, TSB12LV01, TSB12LV21, TSB12LV31, and TSB12C01. Details of how the LLC devices operate are described in the LLC data sheets. The following paragraphs describe the operation of the phy-LLC interface.

The TSB21LV03C supports 100-/200-Mbit/s data transfer and has four bidirectional data lines, D0 – D3, crossing the interface. In 100-Mbit/s operation only D0 and D1 terminals are used. In 200 Mbit/s operation, all Dn terminals are used for data transfer. The unused Dn terminals are driven low. In addition, there are two bidirectional control lines CTL0 and CTL1, the 49.152-MHz SYSCLK line from the phy to the LLC, and the LLC request terminal LREQ from the LLC to the phy. The TSB21LV03C has control of all bidirectional terminals. The LLC is allowed to drive these terminals only after it has been given permission by the phy. The dedicated LREQ request terminal is used by the LLC for any activity that it wishes to initiate.

There are four operations that may occur in the phy-LLC interface: request, status, transmit, and receive. With the exception of the request operation, all actions are initiated by the phy.

When the phy has control of the bus the CTL0 and CTL1 lines are encoded as shown in Table 3.

**Table 3. CTLn Status When Phy Has Control of the Bus**

| CTL0 | CTL1 | STATUS NAME | DESCRIPTION   |
|------|------|-------------|---|
| 0    | 0    | Idle        | No activity is occurring (this is the default mode).                  |
| 0    | 1    | Status      | Status information is being sent from the phy to the LLC.             |
| 1    | 0    | Receive     | An incoming packet is being sent from the phy to the LLC.             |
| 1    | 1    | Transmit    | The LLC has been given control of the bus to send an outgoing packet. |

When the LLC has control of the bus (phy permission) the CTL0 and CTL1 terminals are encoded as shown in Table 4.

**Table 4. CTLn Status When LLC Has Control of the Bus**

| CTL0 | CTL1 | STATUS NAME | DESCRIPTION  |
|------|------|-------------|--|
| 0    | 0    | Idle        | The LLC releases the bus (transmission has been completed).  |
| 0    | 1    | Hold        | The LLC is holding the bus while data is being prepared for transmission or is sending another packet without arbitrating. |
| 1    | 0    | Transmit    | An outgoing packet is being sent from the LLC to the phy.  |
| 1    | 1    | Reserved    | None   |

#### request

When the LLC requests the bus or accesses a register that is located in the TSB21LV03C, a serial stream of information is sent across the LREQ line. The length of the stream varies depending on whether the transfer is a bus request, a read command, or a write command. Regardless of the type of transfer, a start bit of 1 is required at the beginning of the stream, and a stop bit of 0 is required at the end of the stream. Bit 0 is the most significant bit, and is transmitted first. The LREQ terminal is required to idle low (logic level 0).

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### PRINCIPLES OF OPERATION

**Table 5. LLC Bus-Request or Register-Access-Request Bit Length**

| REQUEST TYPE           | NUMBER OF BITS |
|------------------------|----------------|
| Bus request            | 7              |
| Read register request  | 9              |
| Write register request | 17             |

For a Bus Request the length of the LREQ data stream is 7 bits as shown in Table 6.

**Table 6. LLC Bus Request**

| BIT(S) | NAME          | DESCRIPTION   |
|--------|---------------|---|
| 0      | Start Bit     | Indicates the beginning of the transfer (always 1).                             |
| 1–3    | Request Type  | Indicates the type of bus request (see Table 9 for the encoding of this field). |
| 4–5    | Request Speed | Should be 00 for TSB21LV03C 100-Mbit/s speed and 01 for 200-Mbit/s speed.       |
| 6      | Stop Bit      | Indicates the end of the transfer (always 0).                                   |

For a Read Register Request the length of the LREQ data stream is 9 bits as shown in Table 7.

**Table 7. LLC Read Register Access**

| BIT(S) | NAME         | DESCRIPTION   |
|--------|--------------|---|
| 0      | Start Bit    | Indicates the beginning of the transfer (always 1).           |
| 1–3    | Request Type | Always a 100 indicating that this is a read register request. |
| 4–7    | Address      | Identifies the address of the phy register to be read.        |
| 8      | Stop Bit     | Indicates the end of the transfer (always 0).                 |

For a Write Register Request the Length of the LREQ data stream is 17 bits as shown in Table 8.

**Table 8. LLC Write Register Access**

| BIT(S) | NAME         | DESCRIPTION   |
|--------|--------------|---|
| 0      | Start Bit    | Indicates the beginning of the transfer (always 1).                     |
| 1–3    | Request Type | Always a 101 indicating that this is a write register request.          |
| 4–7    | Address      | Identifies the address of the phy register to be written to.            |
| 8–15   | Data         | Gives the data that is to be written to the specified register address. |
| 16     | Stop Bit     | Indicates the end of the transfer (always 0).                           |

The 3-bit Request Type field has the values shown in Table 9.

**Table 9. LLC Bus Request Type**

| LREQ1 | LREQ2 | LREQ3 | NAME     | DESCRIPTION  |
|-------|-------|-------|----------|--|
| 0     | 0     | 0     | ImmReq   | Immediate request. Upon detection of an idle, the LLC takes control of the bus immediately (no arbitration). |
| 0     | 0     | 1     | IsoReq   | Isochronous request: the LLC arbitrates for the bus, no gaps.  |
| 0     | 1     | 0     | PriReq   | Priority request: the LLC arbitrates after a subaction gap, ignores fair protocol.                           |
| 0     | 1     | 1     | FairReq  | Fair request: the LLC arbitrates after a subaction gap, follows fair protocol.                               |
| 1     | 0     | 0     | RdReg    | The LLC returns the specified register contents through a status transfer.                                   |
| 1     | 0     | 1     | WrReg    | The LLC writes to the specified register.  |
| 1     | 1     | 0     | Reserved | Reserved   |
| 1     | 1     | 1     | Reserved | Reserved   |



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

## PRINCIPLES OF OPERATION

LREQ timing (each cell represents one clock sample time):



NOTE B: Each cell represents one clock sample time.

**Figure 11. LREQ Timing**

For fair or priority access, the LLC requests control of the bus at least one clock after the phy-LLC interface becomes idle. If the LLC senses that the CTLn terminals are in a receive state (CTL0 = 1, CTL1 = 0), this indicates that its request has been lost. This is true anytime during or after the LLC sends the bus request transfer. Additionally, the phy ignores any fair or priority requests if it asserts the receive state while the LLC is requesting the bus. The LLC then reissues the request one clock after the next interface idle.

The cycle master uses a normal priority request to send a cycle-start message. After receiving a cycle-start message, the LLC can issue an isochronous bus request. When arbitration is won, the LLC proceeds with the isochronous transfer of data. The isochronous request register is cleared in the phy once the LLC sends another type of request or when the isochronous transfer has been completed. The isochronous request must be issued during a packet reception. Generally this request would be during reception of a cycle-start packet.

The ImmReq request is issued when the LLC needs to send an acknowledgment after reception of a packet addressed to it. This request must be issued during packet reception. This is done to minimize the delays that a phy would have to wait between the end of a packet and the transmittal of an acknowledgment. As soon as the packet ends, the phy immediately grants access of the bus to the LLC. The LLC sends an acknowledgment to the sender unless the header CRC of the packet turns out to be bad. In this case, the LLC releases the bus immediately; it is not allowed to send another type of packet on this grant. To guarantee this, the LLC is forced to wait 160 ns after the end of the packet is received. The phy then gains control of the bus and the acknowledgement with the CRC error is sent. Then the bus is released and allowed to proceed with another requests.

Although highly improbable, it is conceivable that two separate nodes can believe that an incoming packet is intended for them. The nodes then issue a ImmReq request before checking the CRC of the packet. Since both phys seize control of the bus at the same time, a temporary, localized collision of the bus occurs somewhere between the competing nodes. This collision would be interpreted by the other nodes on the network as being a high-impedance line state, not a bus reset. As soon as the two nodes check the CRC, the mistaken node drops its request and the false line state is removed. The only side effect would be the loss of the intended acknowledgment packet (this is handled by the higher-layer protocol).

### read/write requests

When the LLC requests to read the specified register contents, the phy sends the contents of the register to the LLC through a status transfer. When an incoming packet is received while the phy is transferring status information to the LLC, the phy continues to attempt to transfer the contents of the register until it is successful.

For write requests, the phy loads the data field into the appropriately addressed register as soon as the transfer has been completed. The LLC is allowed to request read or write operations at any time.

# TSB21LV03C

## IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

### PRINCIPLES OF OPERATION

#### status

A status transfer is initiated by the phy when it has status information to transfer to the LLC. The phy waits until the interface is idle before starting the transfer. The transfer is initiated by asserting the following on the control terminals: CTL0 – CTL1 = 01 along with the first two bits of status information on the D0 – D3 terminals. The phy maintains CTL0 – CTL1 = 01 for the duration of status transfer. The phy may prematurely end a status transfer by asserting something else other than CTL0 – CTL1 = 01 on the control terminals. This could be caused by an incoming packet from another node. The phy continues to attempt to complete the transfer until the information has been successfully transmitted. There must be at least one idle cycle in between consecutive status transfers.

The phy normally sends just the first 4 bits of status to the LLC. These bits are status flags that are needed by the LLC state machines. The phy sends an entire status packet to the LLC after a request transfer that contains a read request, or when the phy has pertinent information to send to the LLC or transaction layers. The only defined condition where the phy automatically sends a register to the LLC is after self-ID, when it sends the physical-ID register, which contains the new node address. After a power-on reset, the TSB21LV03C sends two self-ID status transfers. The first transfer is invalid (a status of not connected); later, during the same bus reset, a second, correct root, node number, and connection status self-ID is transferred. During all other bus resets, only one Self-ID status is transmitted.

The definition of the bits in the status transfer are shown in Table 10 and the timing is shown in Figure 7.

**Table 10. 16-Bit Stream Status Request**

| BIT(S) | NAME                  | DESCRIPTION   |
|--------|-----------------------|---|
| 0      | Arbitration Reset Gap | Bit 0 indicates that the phy has detected that the bus has been idle for an arbitration reset gap time (this time is defined in the IEEE 1394–1995 standard). Bit 0 is used by the LLC in its busy/retry state machine.         |
| 1      | Subaction Gap         | Bit 1 indicates that the phy has detected that the bus has been idle for a subaction gap time (this time is defined in the IEEE 1394–1995 standard). Bit 1 is used by the LLC to detect the completion of an isochronous cycle. |
| 2      | Bus Reset             | Bit 2 indicates that the phy has entered the bus reset state.   |
| 3      | State Timeout or CPS  | Bit 3 indicates that the phy stayed in a particular state for too long a period, which is usually the effect of a loop in the cable topology, or that the cable power has dropped below the threshold for reliable operation.   |
| 4–7    | Address               | Bits 4 – 7 hold the address of the phy register whose contents are transferred to the LLC.  |
| 8–15   | Data                  | Bits 8 – 15 contain the data that is to be sent to the LLC.   |



**Figure 12. Status Transfer Timing**



**PRINCIPLES OF OPERATION**

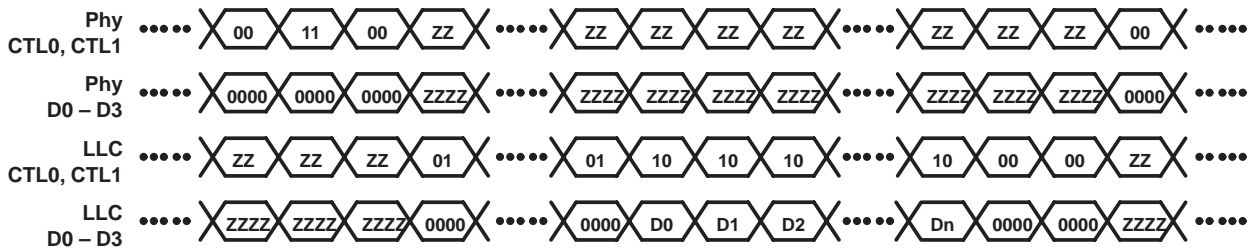
**transmit**

When the LLC wants to transmit information, it first requests access to the bus through the LREQ terminal. Once the phy receives this request, it arbitrates to gain control of the bus. When the phy wins ownership of the serial bus, it grants the bus to the LLC by asserting the transmit state on the CTLn terminals for at least one SYSClk cycle, followed by idle for one clock cycle. The LLC takes control of the bus by asserting either hold or transmit on the CTLn terminals. Hold is used by the LLC to keep control of the bus when it needs some time to prepare the data for transmission. The phy keeps control of the bus for the LLC by asserting a data-on state on the bus. It is not necessary for the LLC to use hold when it is ready to transmit as soon as bus ownership is granted.

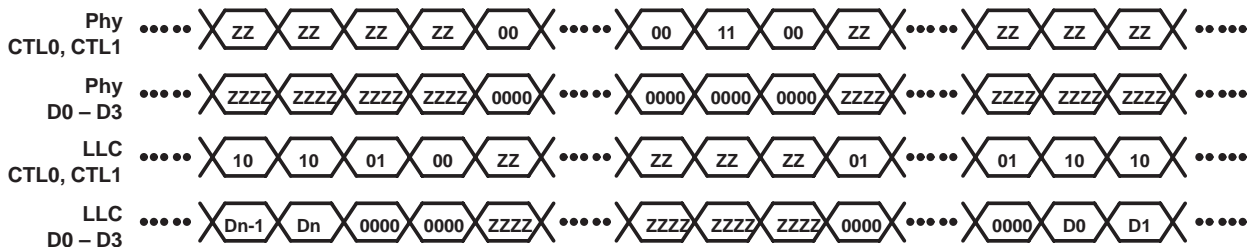
When the LLC is prepared to send data, it asserts the transmit state on the CTLn terminals as well as sending the first bits of the packet on the D0 – D3 lines (assuming 200 Mbits/s). The transmit state is held on the CTLn terminals until the last bits of data have been sent. The LLC then asserts an idle state on the CTLn terminals for one clock cycle after which it releases control of the interface.

However, there are times when the LLC needs to send another packet without releasing the bus. For example, the LLC may want to send consecutive isochronous packets or it may want to attach a response to an acknowledgment. To do this, the LLC asserts a hold state instead of an idle state when the first packet of data has been completely transmitted. In this case, hold informs the phy that the LLC needs to send another packet without releasing control of the bus. The phy then waits a set amount of time before asserting a transmit state. The LLC can then proceed with the transmittal of the second packet. After all data has been transmitted and the LLC has asserted an idle state on the CTLn terminals, the phy asserts its own idle state on the CTLn terminals. When sending multiple packets in this fashion, it is required that all data be transmitted at the same speed. This is required because the transmission speed is set during arbitration and since the arbitration step is skipped, there is no way of informing the network of a change in speed.

**Single Packet**



**Continued Packet**



NOTE A: ZZ = High-impedance state  
D0 => Dn = Packet data

**Figure 13. Transmit Timing Waveforms**

# TSB21LV03C IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

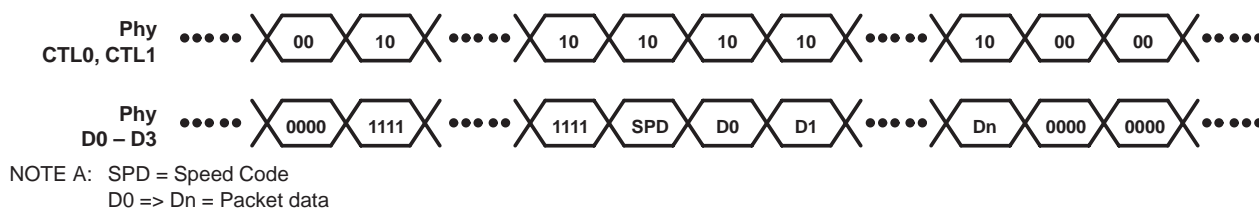
## PRINCIPLES OF OPERATION

### receive

When data is received by the phy from the serial bus, the phy transfers the data to the LLC for further processing. The phy asserts a receive state on the CTLn terminals and asserts a 1 on each Dn terminal. The phy indicates the start of the packet by placing the speed code on the data bus. The phy then proceeds with the transmittal of the packet to the LLC on the Dn terminals while still keeping the receive status on the CTLn terminals. Once the packet has been completely transferred, the phy asserts an idle state on the CTLn terminals, which completes the receive operation.

**NOTE:**

The speed is a phy-LLC protocol and not included in the CRC.



NOTE A: SPD = Speed Code  
D0 => Dn = Packet data

**Figure 14. Receive Timing Waveforms**

**Table 11. Speed Code for the Receiver**

| D0 – D3 | Data Rate  |
|---------|------------|
| 00YY†   | 100 Mbit/s |
| 0100    | 200 Mbit/s |

† Y = Transmitted as 0, ignored on receive.

### power class bits in self-ID packet

Table 12 describes the meaning of the power-class bits in the pwr field of the Self-ID packet. Bit 21 is transmitted first, followed by bit 22 and then bit 23. This power-field bit description complies with the IEEE 1394-1995 standard.

**Table 12. Self-ID Packet Pwr-Field Bit Description**

| PC0-PC2 | DESCRIPTION  |
|---------|--|
| 000     | Node does not need power and does not repeat power.  |
| 001     | Node is self powered, and provides a minimum of 15 W to the bus.   |
| 010     | Node is self powered, and provides a minimum of 30 W to the bus.   |
| 011     | Node is self powered, and provides a minimum of 45 W to the bus.   |
| 100     | Node may be powered from the bus, and is using up to 1 W.  |
| 101     | Node may be powered from the bus, and is using up to 1 W. An additional 2 W is needed to enable the LLC and higher layers. |
| 110     | Node may be powered from the bus, and is using up to 1 W. An additional 5 W is needed to enable the LLC and higher layers. |
| 111     | Node may be powered from the bus, and is using up to 1 W. An additional 9 W is needed to enable the LLC and higher layers. |



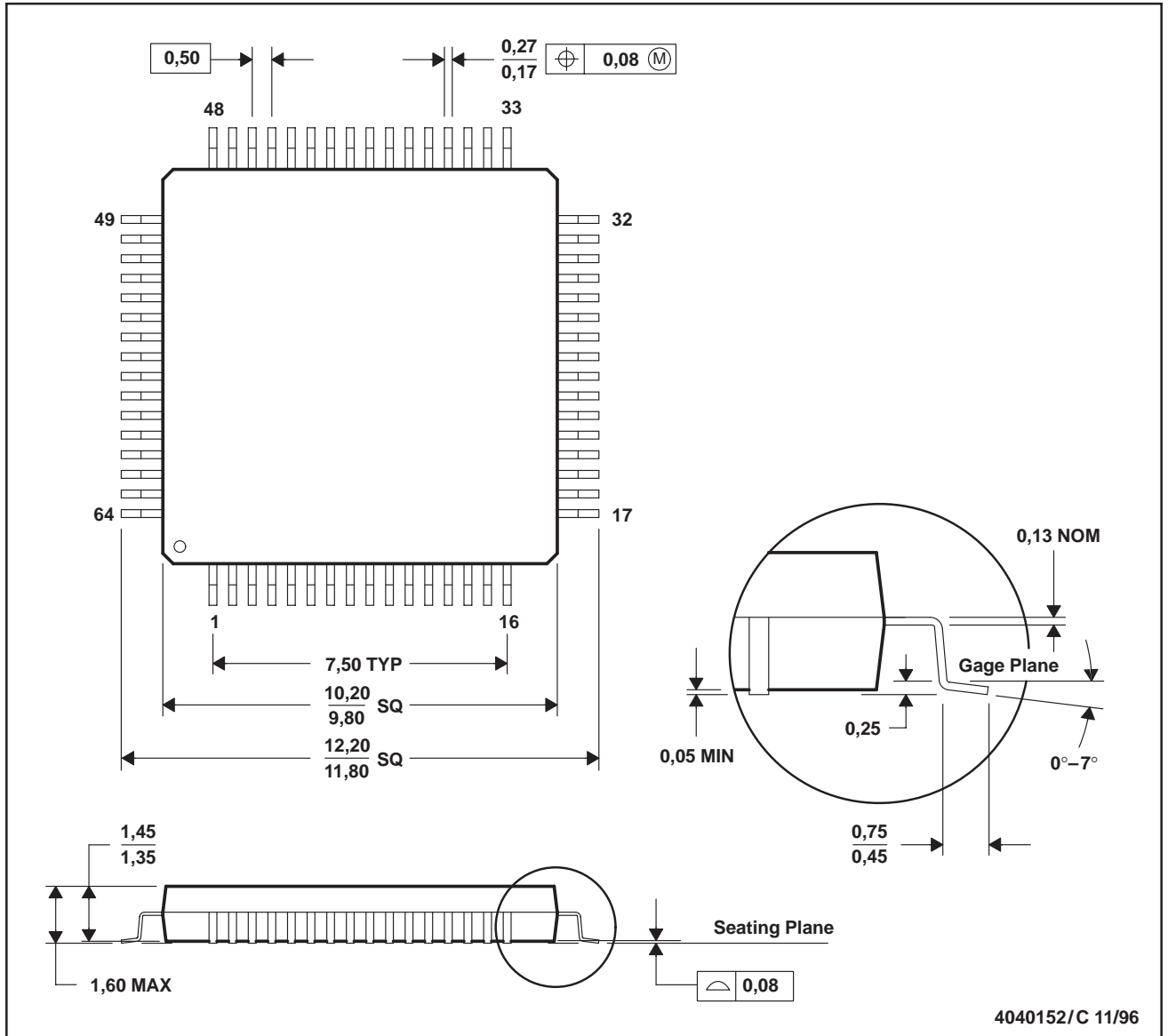
# TSB21LV03C IEEE 1394-1995 TRIPLE-CABLE TRANSCEIVER/ARBITER

SLLS331A – FEBRUARY 1999 – REVISED OCTOBER 1999

## MECHANICAL INFORMATION

PM (S-PQFP-G64)

PLASTIC QUAD FLATPACK



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Falls within JEDEC MS-026
  - D. May also be thermally enhanced plastic with leads connected to the die pads.

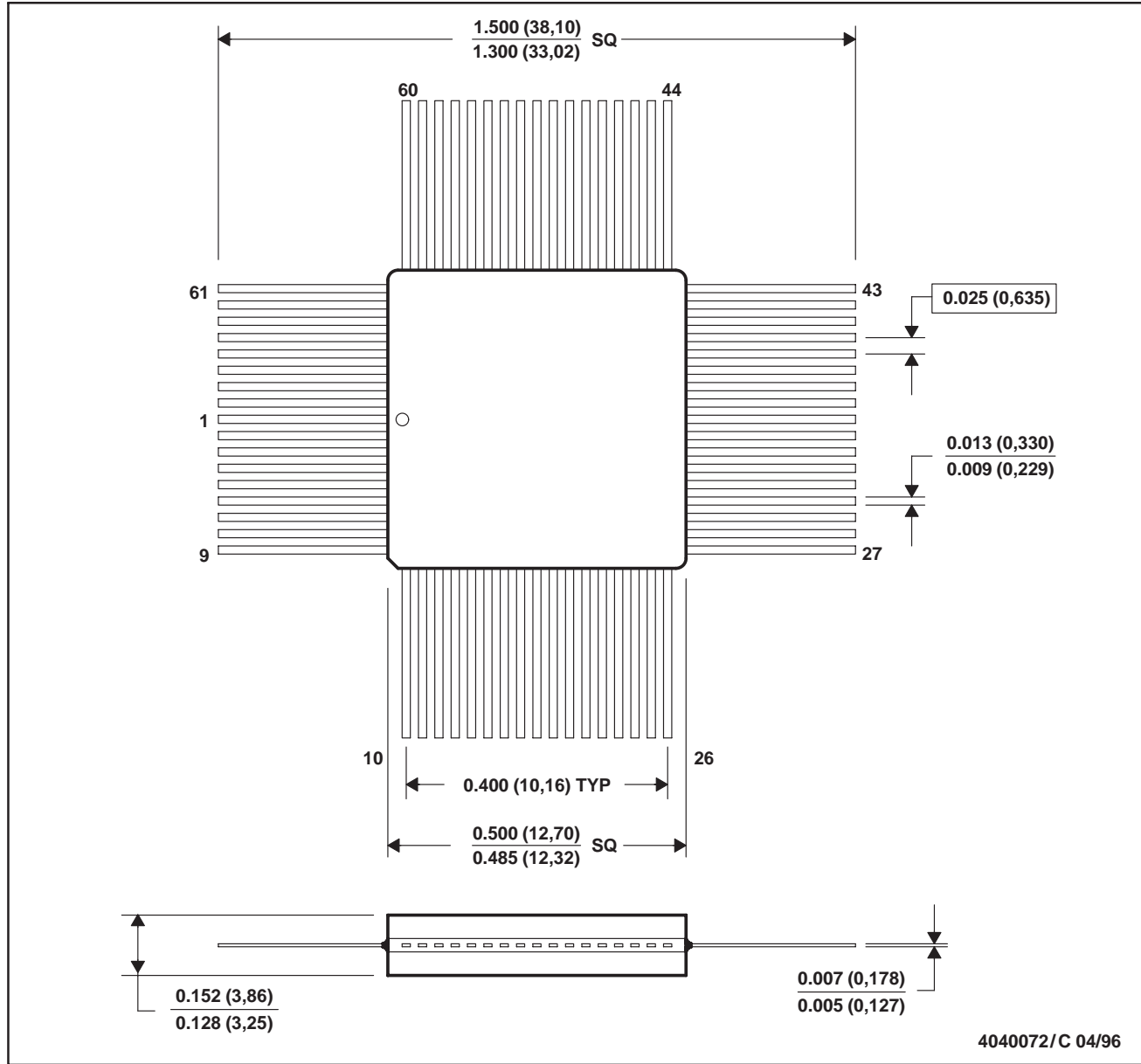
# MECHANICAL DATA

MCFP013A – JANUARY 1995 – REVISED OCTOBER 1996

## MECHANICAL INFORMATION

HV (S-GQFP-F68)

CERAMIC QUAD FLATPACK



- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

## **IMPORTANT NOTICE**

Texas Instruments and its subsidiaries (TI) reserve the right to make changes to their products or to discontinue any product or service without notice, and advise customers to obtain the latest version of relevant information to verify, before placing orders, that information being relied on is current and complete. All products are sold subject to the terms and conditions of sale supplied at the time of order acknowledgement, including those pertaining to warranty, patent infringement, and limitation of liability.

TI warrants performance of its semiconductor products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

CERTAIN APPLICATIONS USING SEMICONDUCTOR PRODUCTS MAY INVOLVE POTENTIAL RISKS OF DEATH, PERSONAL INJURY, OR SEVERE PROPERTY OR ENVIRONMENTAL DAMAGE ("CRITICAL APPLICATIONS"). TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS. INCLUSION OF TI PRODUCTS IN SUCH APPLICATIONS IS UNDERSTOOD TO BE FULLY AT THE CUSTOMER'S RISK.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards must be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance or customer product design. TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used. TI's publication of information regarding any third party's products or services does not constitute TI's approval, warranty or endorsement thereof.