### JNITRODE



UC1851 UC2851 UC3851

# Programmable, Off-Line, PWM Controller

#### **FEATURES**

- All Control, Driving, Monitoring, and Protection **Functions Included**
- Low-Current Off Line Start Circuit
- Voltage Feed Forward or Current Mode Control
- High Current Totem Pole Output
- 50% Absolute Max Duty Cycle
- PWM Latch for Single Pulse Per Period
- Pulse-by-Pulse Current Limiting plus Shutdown for Over-Current Fault
- No Start-Up or Shutdown Transients
- Slow Turn-On Both Initially and After Fault Shutdown
- Shutdown Upon Over or Under Voltage Sensing
- Latch Off or Continuous Retry After Fault
- 1% Reference Accuracy
- 500kHz Operation
- 18 Pin DIL or 20 Pin PLCC Package

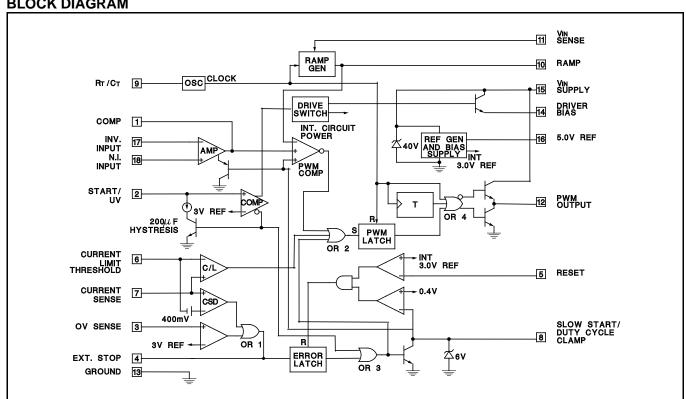
#### **DESCRIPTION**

The UC1851 family of PWM controllers are optimized for offline primary side control. These devices include a high current totem pole output stage and a toggle flip-flop for absolute 50% duty cycle limiting. In all other respects this line of controllers is pin for pin compatible with the UC1841 series. Inclusion of all major housekeeping functions in these high performance controllers makes them ideal for use in cost sensitive applications.

Important features of these controllers include low current start-up, linear feed-forward for constant volt-second operation, and compatibility with both voltage or current mode control. In addition, these devices include a programmable start threshold, as well as programmable over-voltage, under-voltage, and over current fault thresholds. The fault latch on these devices can be configured for automatic restart, or latched off response to a fault.

These devices are packaged in 18-pin plastic or ceramic dualin-line packages, or for surface mount applications, a 20 Pin PLCC. The UC1851 is characterized for -55°C to +125°C operation while the UC2851 and UC3851 are designed for -40°C to +85°C and 0°C to +70°C, respectively.

### **BLOCK DIAGRAM**



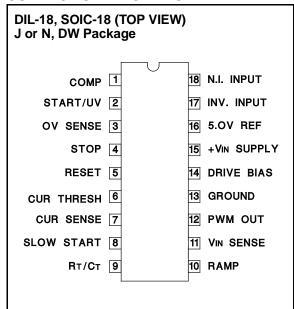
### **ABSOLUTE MAXIMUM RATINGS** (Note 1)

Supply Voltage, +VIN (Pin 15)	
Voltage Driven	. +32V
Current Driven, 100mA maximum Self-	-limiting
PWM Output Voltage (Pin 12)	40V
PWM Output Current, Steady-State (Pin 12)	400mA
PWM Output Peak Energy Discharge 20µ	<b>Joules</b>
Driver Bias Current (Pin 14)	·200mA
Reference Output Current (Pin 16)	-50mA
Slow-Start Sink Current (Pin 8)	20mA
VIN Sense Current (Pin 11)	10mA
Current Limit Inputs (Pins 6 & 7)0.5 to	5.5V ±
Stop Input (Pin 4)0.3 to	o +5.5V

Comparator Inputs
(Pins 1–7, 9–11, 16) Internally clamped at 12V
Power Dissipation at T <sub>A</sub> = 25°C (Note 3) 1000mW
Power Dissipation at T <sub>C</sub> = 25°C (Note 3) 2000mW
Operating Junction Temperature55°C to +150°C
Storage Temperature Range65°C to +150°C
Lead Temperature (Soldering, 10 sec) +300°C
Note 1:All voltages are with respect to ground, Pin 13.
Currents are positive-into, negative-out of the
specified terminal
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Note 2:All pin numbers are referenced to DIL-18 package. Note 3:Consult Packaging Section of Databook for thermal limitations and considerations of package.

## **CONNECTION DIAGRAMS**



PLCC-20, LCC-20	C-20   CC-20 PACKAGE PIN FUNCTIONS				
(TOP VIEW)	200-20				
Q, L PACKAGE	COMP	1			
<u> </u>	START/UV	2			
	OV SENSE	3			
	STOP	4			
/ <del></del>	RESET	5			
3 2 1 20 19	CUR THRESH	7			
<b>4</b> 18	CUR SENSE	8			
5 17	SLOW START	9			
6 16	RT/CT	10			
	RAMP	11			
[ 7 15]	Vin SENSE	12			
8 0 40 44 40 40	PWM OUT	13			
9 10 11 12 13	GROUND	14			
	DRIVE BIAS	15			
	+Vin SUPPLY	17			
	5.0V REF	18			
	INV. INPUT	19			
	N.I. INPUT	20			

**ELECTRICAL CHARACTERISTICS:** Unless otherwise stated, these specifications apply for TA = -55°C to +125°C for the UC1851, -40°C to +85°C for the UC2851, and 0°C to 70°C for the UC3851; ViN = 20V, RT =  $20k\Omega$ , CT = .001 mfd, RR =  $10k\Omega$ , CR = .001mfd. Current Limit Threshold = 200mV, TA = TJ.

		UC1851 / UC2851			UC3851			UNITS
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	
Power Inputs	Power Inputs							
Start-Up Current	VIN = 30V, Pin 2 = 2.5V		4.5	6		4.5	6	mA
Operating Current	VIN = 30V, Pin 2 = 3.5V		15	21		15	21	mA
Supply OV Clamp	VIN = 20mA	33	39	45	33	39	45	V
Reference Section								
Reference Voltage	T <sub>J</sub> = 25°C	4.95	5.0	5.05	4.9	5.0	5.1	V
Line Regulation	VIN = 8 to 30V		10	15		10	20	mV
Load Regulation	IL = 0 to 10mA		10	20		10	30	mV
Total Ref Variation	Over Operating Temperature Range	4.9		5.1	4.85		5.15	V
Short Circuit Current	VREF = 0, TJ = 25°C		-80	-100		-80	-100	mA
Oscillator								
Nominal Frequency	T <sub>J</sub> = 25°C	47	50	53	45	50	55	kHz
Voltage Stability	VIN = 8 to 30V		0.5	1		0.5	1	%
Total Ref Variation	Over Operating Temperature Range	45		55	43		57	kHz
Maximum Frequency	$RT = 2k\Omega$ , $CT = 330pF$	500			500			kHz

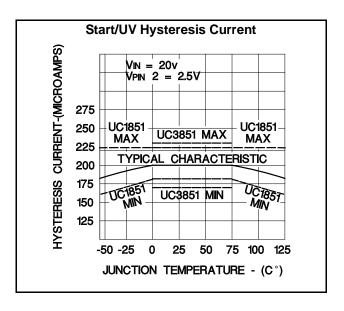
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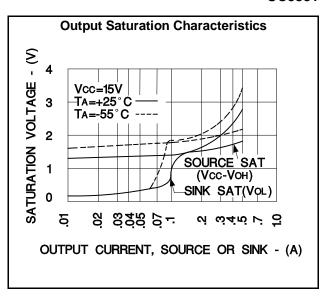
PARAMETER	TEST CONDITIONS	UC1851 / UC2851		UC3851			UNITS	
		MIN	TYP	MAX	MIN	TYP	MAX	
Ramp Generator								
Ramp Current, Minimum	ISENSE = -10μA		-11	-14		-11	-14	μΑ
Ramp Current, Maximum	ISENSE = 1.0mA	-0.9	95		-0.9	95		mA
Ramp Valley		0.3	0.4	0.6	0.3	0.4	0.6	V
Ramp Peak	Clamping Level	3.9	4.2	4.5	3.9	4.2	4.5	V
Error Amplifier		•		•			•	-
Input Offset Voltage	Vcm = 5.0V		0.5	5		2	10	mV
Input Bias Current			0.5	2		1	5	μΑ
Input Offset Current				0.5			0.5	μΑ
Open Loop Gain	$\Delta Vo = 1$ to $3V$	60	66		60	66		dB
Output Swing (Max Output ≤ Ramp Peak - 100mV)	Minimum Total Range	0.3		3.5	0.3		3.5	V
CMRR	VcM = 1.5 to 5.5V	70	80		70	80		dB
PSRR	VIN = 8 to 30V	70	80		70	80		dB
Short Circuit Current	VCOMP = 0V		-4	-10		-4	-10	mA
Gain Bandwidth (Note 1)	T <sub>J</sub> = 25°C, Avol = 0dB	1	2		1	2		MHz
Slew Rate (Note 1)	T <sub>J</sub> = 25°C, A <sub>V</sub> C <sub>L</sub> = 0dB		0.8			0.8		V/μs
PWM Section		•	1					
Continuous Duty Cycle Range (other than zero) (Note 1)	Minimum Total Continuous Range Ramp Peak < 4.2V	2		46	2		46	%
Output High Level	ISOURCE = 20mA	18	18.5		18	18.5		V
	ISOURCE = 200mA	17	18.5		17	18.5		V
Rise Time (Note 1)	T <sub>J</sub> = 25°C, C <sub>L</sub> = 1nF		50	150		50	150	ns
Fall Time (Note 1)	T <sub>J</sub> = 25°C, C <sub>L</sub> = 1nF		50	150		50	150	ns
Output Saturation	IOUT = 20mA		0.2	0.4		0.2	0.4	V
	IOUT = 200mA		1.7	2.2		1.7	2.2	V
Comparator Delay (Note 1)	Pin 8 to Pin 12, TJ = 25°C, RL = $1k\Omega$		300	500		300	500	ns
Sequencing Functions			•					
Comparator Thresholds	Pins 2, 3, 5	2.8	3.0	3.2	2.8	3.0	3.2	V
Input Bias Current	Pins 3, 5 = 0V		-1.0	-4.0		-1.0	-4.0	μΑ
Input Leakage	Pins 3, 5 = 10V		0.1	2.0		0.1	2.0	μΑ
Start/UV Hysteresis Current	Pin 2 = 2.5V	170	200	220	170	200	230	μΑ
Ext. Stop Threshold	Pin 4	0.8	1.6	2.4	0.8	1.6	2.4	V
Error Latch Activate Current	Pin 4 = 0V, Pin 3 > 3V		-120	-200		-120	-200	μΑ
Driver Bias Saturation Voltage, VIN-VOH	IB = -50mA		2	3		2	3	V
Driver Bias Leakage	VB = 0V		-0.1	-10		-0.1	-10	μΑ
Slow-Start Saturation	Is = 10mA		0.2	0.5		0.2	0.5	V
Slow-Start Leakage	Vs = 4.5V		0.1	2.0		0.1	2.0	μΑ
Current Control				·		·		
Current Limit Offset			0	5		0	10	mV
Current Shutdown Offset		370	400	430	360	400	440	mV
Input Bias Current	Pin 7 = 0V		-2	-5		-2	-5	μΑ
Common Mode Range (Note 1)		-0.4		3.0	-0.4		3.0	V
Current Limit Delay (Note 1)	T <sub>J</sub> = 25°C, Pin 7 to 12, R <sub>L</sub> = 1k		200	400		200	400	ns

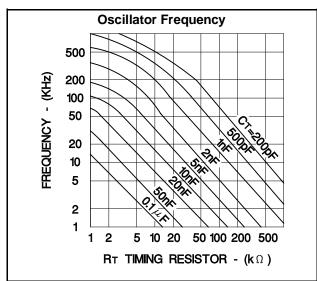
Note 1:Guaranteed by design. Not 100% tested in production.

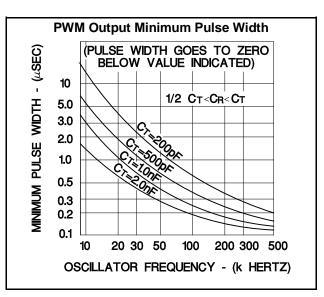
### **FUNCTIONAL DESCRIPTION**

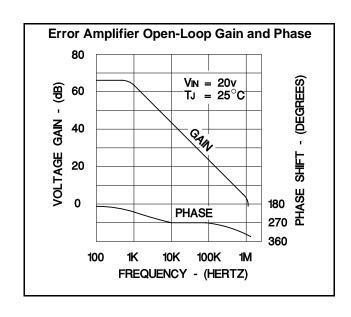
PWM CONTROL	
1. Oscillator	Generates a fixed-frequency internal clock from an external RT and CT.
	Frequency = $\frac{KC}{RTCT}$ where Kc is a first-order correction factor $\approx 0.3 \log (CT \times 10^{12})$ .
2. Ramp Generator:	Develops linear ramp with slope defined externally by $\frac{dV}{dT} = \frac{\text{sense voltage}}{\text{RRCR}}$ .
	CR is normally selected ≤ CT and its value will have some effect upon valley duty cycle. Limiting the minimum value for ISENSE into pin 11 will establish a maximum duty cycle clamp. CR terminal can be used as an input port for current mode control.
3. Error Amplifier	Conventional operational amplifier for closed-loop gain and phase compensation.  Low output impedance; unity-gain stable.  The output is held low by the slow start voltage at turn on in order to minimize overshoot.
4. Reference Generator:	Precision 5.0V for internal and external usage to 50mA.  Tracking 3.0V reference for internal usage only with nominal accuracy of ±2%.  40V clamp zener for chip OV protection, 100mA maximum current.
5. PWM Comparator:	Generates output pulse which starts at termination of clock pulse and ends when the ramp input crosses the lowest of two positive inputs.
6. PWM Latch:	Terminates the PWM output pulse when set by inputs from either the PWM comparator, the pulse-by-pulse comparator, or the error latch. Resets with each internal clock pulse.
7. PWM Output Switch:	Totem pole output stage capable of sourcing and sinking 1 amp peak current. The active "on" state is high.
SEQUENCING FUNCTION	S
1. Start/UV Sense:	With an increasing voltage, this comparator generates a turn-on signal and releases the slow start clamp at a start threshold.  With a decreasing voltage, it generates a turn-off command at a lower level separated by a 200µA
	hysteresis current.
2. Drive Switch:	Disables most of the chip to hold internal current consumption low, and Driver Bias OFF, until input voltage reaches start threshold.
3. Driver Bias:	Supplies drive to external circuitry upon start-up.
4. Slow Start:	Clamps low to hold PWM OFF. Upon release, rises with rate controlled by RsCs for slow increase of output pulse width.  Can also be used as an alternate maximum duty cycle clamp with an external voltage divider.
PROTECTION FUNCTIONS	
1. Error Latch:	When set by momentary input, this latch insures immediate PWM shutdown and hold off until reset. Inputs to Error Latch are: a. OV > 3.2V (Typically 3V) b. Stop > 2.4V (Typically 1.6V) c. Current Sense 400mV over threshold. (Typical). Error Latch resets when slow start voltage falls to 0.4V if Reset Pin < 2.8V. With Pin 5 > 3.2V,
0.00	Error Latch will remain set.
2. Current Limiting:	Differential input comparator terminates individual output pulses each time sense voltage rises above threshold.  When sense voltage rises to 400mV (typical) above threshold, a shutdown signal is sent to Error Latch.
3. External Stop:	A voltage over 2.4 will set the Error Latch and hold the output off. A voltage less than 0.8V will defeat the error latch and prevent shutdown. A capacitor here will slow the action of the error latch for transient protection by providing a Typical Delay of 13ms/μF.

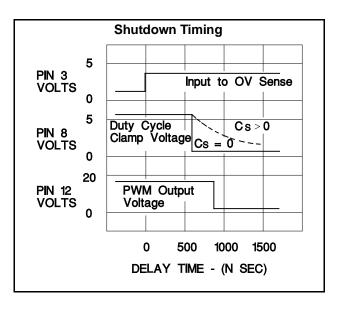




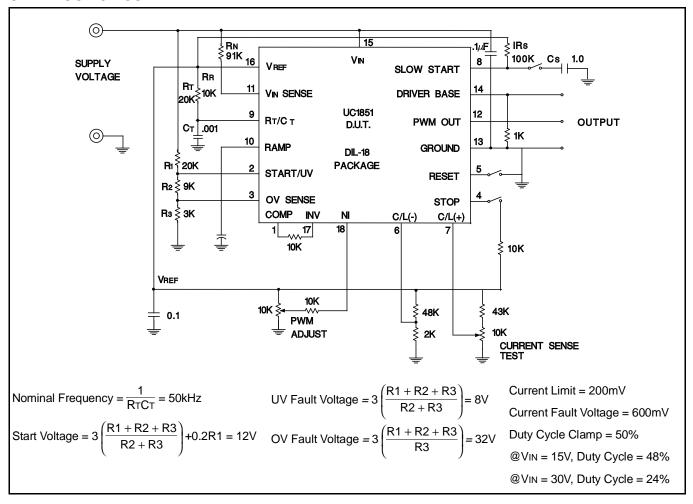




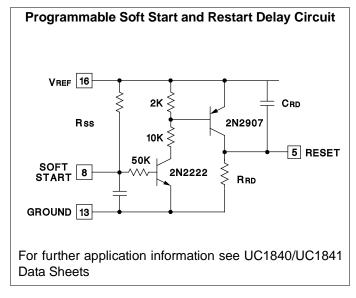


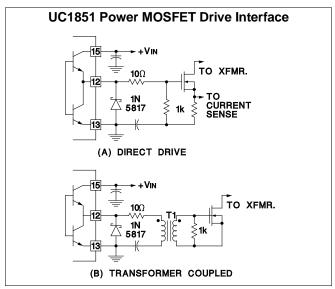


#### **OPEN-LOOP CIRCUIT**



High Peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to pin 13 in a single ground point.





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